

**MM53200 Encoder/Decoder**

The MM53200 Encoder/Decoder is an MOS/LSI Digital Code Transmitter — Receiver system.

**Features**

- A single chip contains both the Encoder and Decoder.
- Oscillator stability is non-critical, 5% components may be used.
- Cross interference of receivers in close proximity is virtually eliminated by circuitry which requires 4 valid words to be received, each within 64 ms of the other.

**Operation**

In the transmit mode the twelve inputs are scanned sequentially producing the output pattern shown in Figure

1. This code is generated at the rate of 0.96 ms/bit, or 11.52 ms/word with 11.52 ms reset pulse between words.

In the receiver mode, the incoming signal is compared to the local code in a sequential manner; if there is an error, the system is reset and begins its comparison on the next pulse. If all twelve bits are received correctly, a "valid" signal will be generated. This signal clears a 64 ms counter and clocks a 3 stage counter. The 3 stage counter counts the "valid" pulses and when 4 pulses have been received, the transmit/receive output goes low. After the transmit/receive output is enabled, the next "valid" must be received within 128 ms, giving a one valid in 6 requirement to keep the transmit/receive output low.

Connection diagrams for the device in the Receive and Transmit modes are shown in Figures 2 and 3.

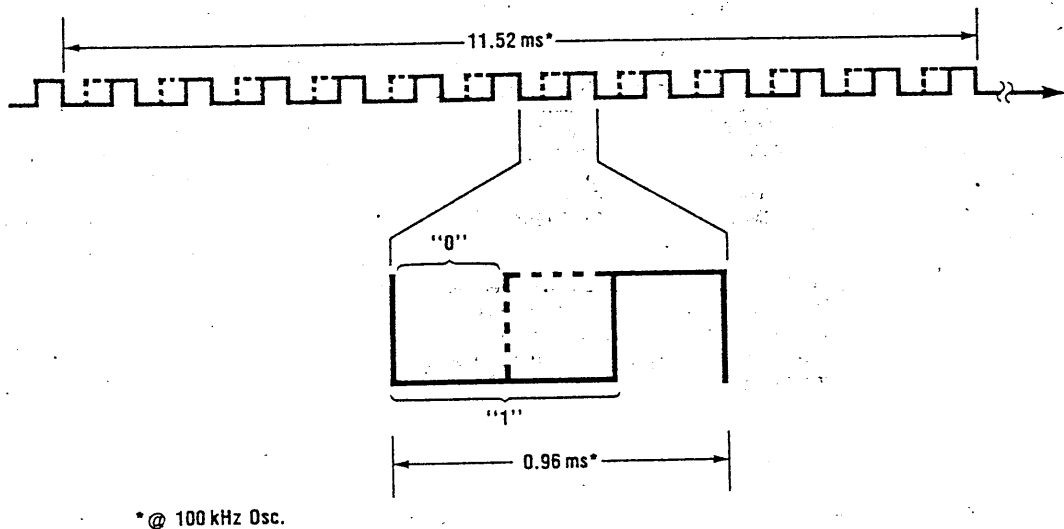


Figure 1. Output Waveform

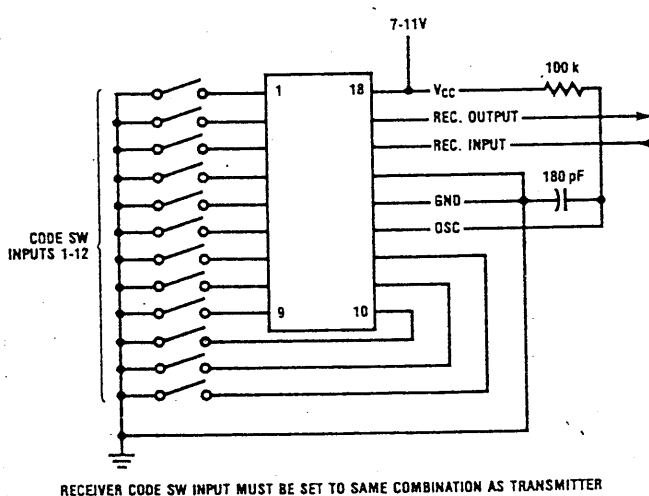


Figure 2. Pin Connections for Receiver Mode

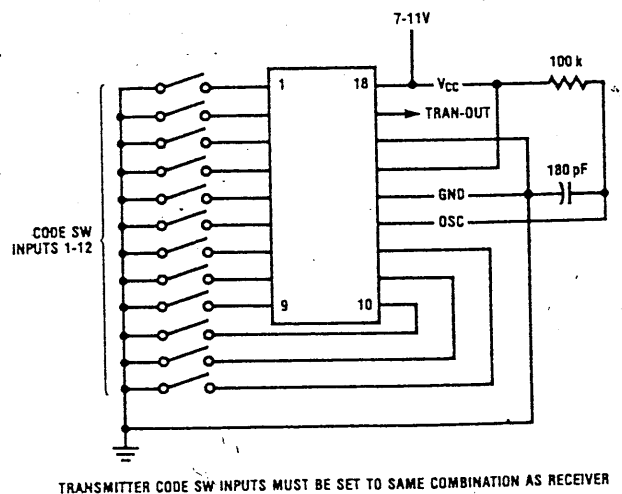


Figure 3. Pin Connections for Transmitter Mode

# Application of the MM53200 Encoder/Decoder

National Semiconductor  
 Application Note AN-290  
 Thomas B. Mills  
 Central Applications Engineering  
 January 1982



Application of the MM53200 Encoder/Decoder

## Introduction

The MM53200 is an easy-to-use MOS-LSI encoder-decoder designed for simple and reliable on-off signaling applications, such as garage door openers, electronic key, and alarm systems.

Application of the MM53200 requires only a resistor and capacitor to function as an encoder or decoder. In the encoder mode, a 12-bit pulse width data stream is generated according to the state of the data select inputs. Input pullup resistors require only that a single pole closure to ground be made to change a particular bit.

In the receive mode, the data stream is compared bit by bit to the data inputs of the decoder chip. If no errors are found, a "valid" signal is generated which clears a 64ms counter and clocks a 3-stage counter. Valid pulses are counted by the 3-stage counter so that when four valid data streams have been received, the decoder output goes low, and remains low as long as one in six data streams are valid. This arrangement guarantees the decoder will not accept a "valid" transmission on noise, yet have enough hysteresis not to "chatter" on noisy signals.

A simple signaling system for two wire applications is shown in Figure 1. This system uses base band signaling and has been tested successfully to simulated distances of over 1,000 ft. of No. 22 twisted-pair telephone wire. Output current limits the capability of the part to drive substantially more wire capacitance, however, the addition of a buffer amplifier will substantially increase the distance signaling that may be achieved.

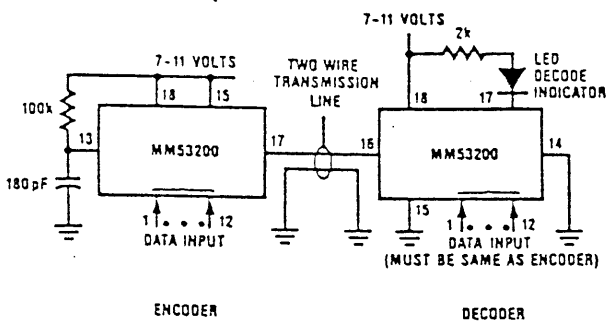


Figure 1. Simple Two-Wire Signaling System or Test Circuit.

In most applications, the data stream will be used to modulate a carrier frequency that is optimum for the medium over which the command signal must be sent.

These carriers will generally fall in the following categories:

- a. Ultrasonic: 30-60kHz
- b. Carrier Current: 50-300kHz
- c. Unlicensed RF Transmission: 49 or 300MHz
- d. Light: Red or Infrared

In the normal mode of operation, with a 100kHz clock frequency, the MM53200 generates 0.3ms pulses (or a baud rate of 3kHz). In order to pass these pulses, the baseband channel bandwidth must be greater than 6kHz. Since a data word lasts 11.52ms, with a 11.52ms "dead" time between words, the low frequency response of the channel must be 20 times less than the pulse duration for less than 10% base line shift, or approximately 2Hz.

## Signal-to-Noise Performance

In any practical communications system, noise is the limiting factor in communications distance. For a given transmitted power, whether limited by FCC or practical constraints, the received signal can only be amplified to the point that receiver noise masks the signal and makes decoding impossible.

Figure 2 shows a typical communications link. The input to the receiver consists of both signal and noise. From this point on, the only thing that can be done to improve the signal-to-noise ratio is reducing the bandwidth of the filter. However, this can only be done to the point where there is no loss of signal—in fact, an optimum point exists which is approximately:  $B\tau=0.7$  for a multistage RC filter, where B=bandwidth in Hz;  $\tau$ =pulse width in seconds.

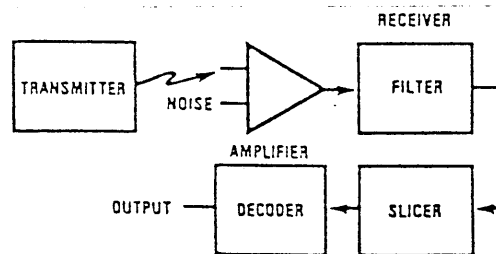


Figure 2. Typical Communications Link.

This factor can be determined experimentally for any digital decoder: by using circuitry similar to that of Figure 3. Signal amplitude is adjusted for a convenient level at the test point, while the slicing level is set to be halfway between the peak-to-peak signal level. Noise is increased until decoding occurs 50% of the time and the ratio of peak-to-peak signal vs. rms noise as measured at the test point is noted. If this ratio is plotted vs. normalized filter bandwidth, the curve of Figure 4 results.

### Ultrasonic Remote Control

The clock of MM53200 is set to 12.5kHz, with pulses of 3.2ms (100k, C=1500pF).

The LF357, a fast operational amplifier, is connected as a square waves oscillator, which is triggered by the transistor. It will deliver bursts of the carrier frequency (38kHz depending on the transducer used). The optimum value is adjusted with the 100kHz potentiometer, according to the transducer specifications.

The ultrasonic decoder uses a dual BIFET op amp, LF353. The first stage is simply an amplifier with a gain of 41dB. The signal is then demodulated by the diode, DC level is isolated by the 3.3μF capacitor, and signal is

again amplified by the second half of the LF358. The signal is then formed through a third op amp (one-half LM358) which acts as a comparator and drives the receiving MM53200. The second half of LM358 is used as a buffer to drive the load.

This circuitry has been tested up to 30 ft. The range could be increased with other special transducers.

### UHF Remote Control Link

A UHF transmitter/receiver application for the MM53200 is described. The circuits are suitable for general purpose remote control over distances of 100-1,000 ft. The circuits shown are suitable for FCC approval under Part 15, Subpart D for the transmitter and Subpart C for the receiver. They operate in the 300-400MHz baud.

The transmitter is a grounded-base Colpitts-type tuned-collector oscillator, similar to that used for television and FM local oscillators. Output pulses from the MM53200 are applied to the base of the oscillator transistor turning it on and off. Frequency of operation is determined by L<sub>1</sub>, C<sub>2</sub>, and stray capacitance. Feedback is between collector and emitter via stray capacitance in Q<sub>1</sub>.

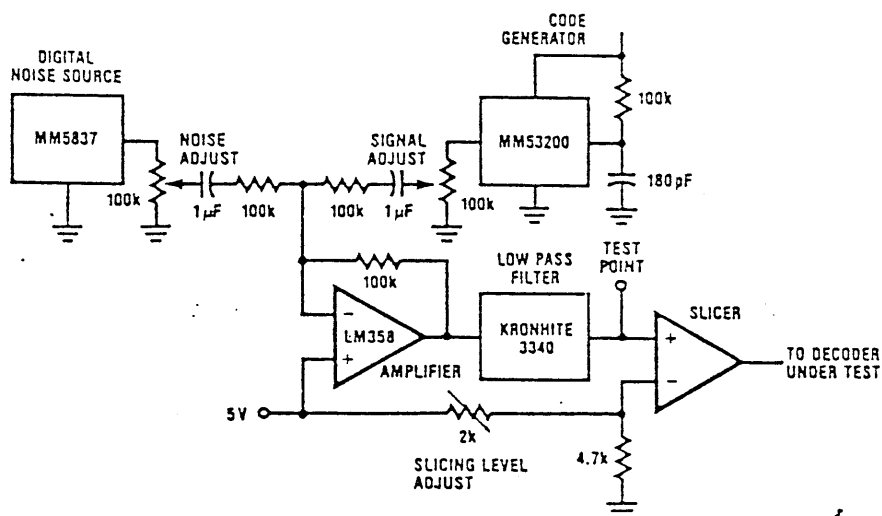


Figure 3. Signal-to-Noise Test Circuit.

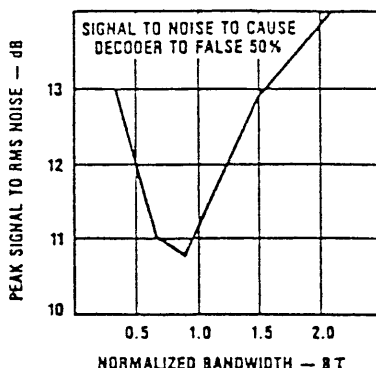


Figure 4. Signal-to-Noise vs. Normalized Bandwidth.

The receiver is a superregenerative type with a grounded-base RF amplifier ( $Q_1$ ) to increase sensitivity and reduce detector radiation.  $Q_2$  functions as the detector which is essentially a UHF oscillator that continues to turn itself on and off at a 200kHz rate. The detected signal is amplified by a dual operational amplifier, one half of which is used as a linear small signal amplifier while the second is used as a comparator to drive the decoder IC.

With a  $4\mu\text{V}$  peak RF input signal, approximately  $0.5\text{mV}$  of signal is available from the detector, and  $100\text{mV}_{\text{p-p}}$  is available at the input of  $A_2$ . At this level, the peak signal to rms noise at the output of  $A_1$  is approximately 12dB and satisfactory decoding should result. Receiver center frequency may be varied by changing  $C_8$  with

little effect on sensitivity,  $A_2$  produces logic level pulses for the MM53200 decoder, whose input data stream agrees with the preset code.

A voltage regulator is required since the detector circuit has no power supply rejection and small variations in supply voltage due to ripple and load variation will cause loss of data.

A properly operating system will have very narrow pulses of 6V peak at a 200-400kHz rate across  $R_9$ . Detector operation may be checked by using Pin 1 of  $A_1$  as a test point. Here, with no input signal, there should be approximately  $0.2\text{V}_{\text{p-p}}$  noise. This point may be used to tune receivers and transmitters together for maximum response.

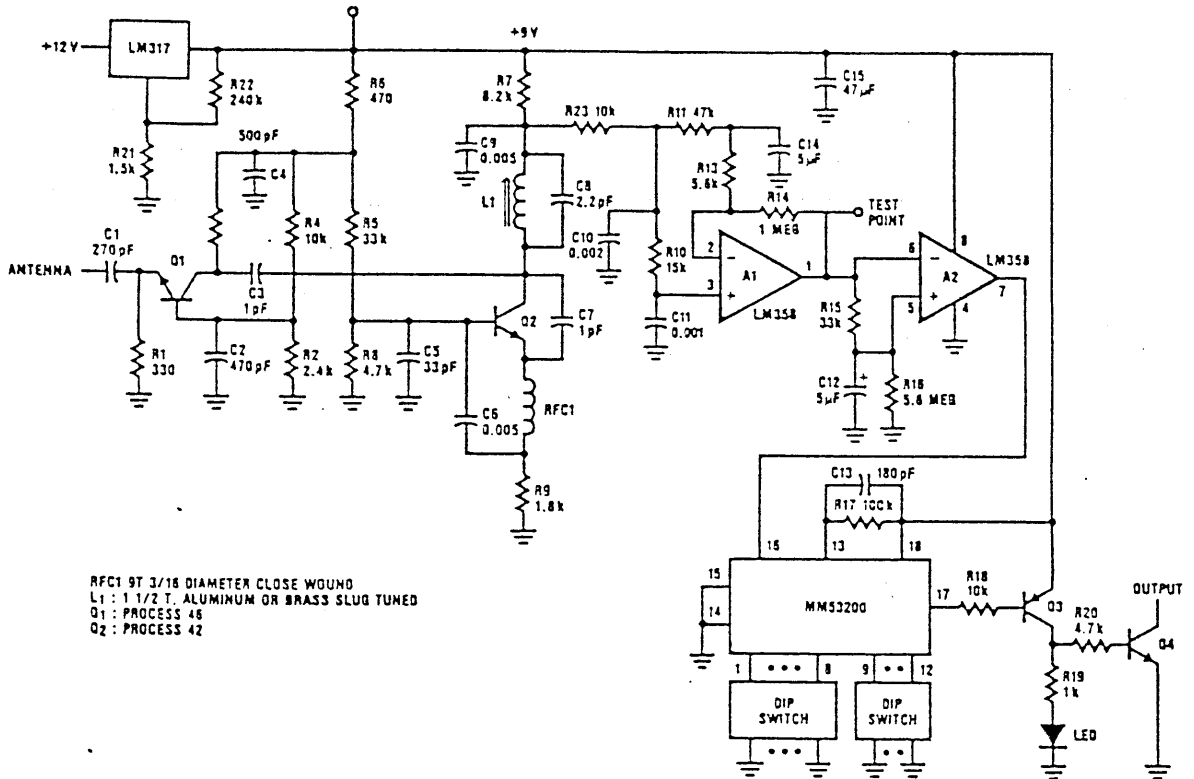


Figure 8. Remote Control Receiver.

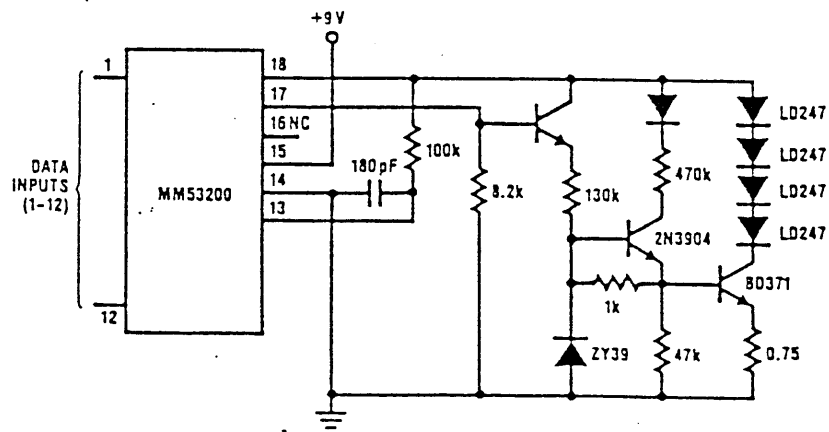


Figure 9. Infrared Transmitter.



## Infrared Link

For Infrared transmission, the encoded signal has to be amplified to drive the emitting diode. We are using a transistor amplifier, with a constant current driving capability. The signal is transmitted by four Infrared emitting diodes (here LD241). One LED is connected in series in the collector of one of the transistors to "show" transmission.

The receiver uses two Infrared sensitive diodes (BPW61). The first diode is actually the receiver, which receives both the signal and the ambient light (noise). The second diode receives only ambient light. A difference amplifier gives the resulting signal with low distortion, which is again amplified and fed into Pin 16 of MM53200. The amplifier is a LM324, quad op amp. Two small signal diodes (1N4148) are used to limit input signal amplitude, which might be used when the receiver is very close to the transmitter.

We show here three receivers connected in parallel after one filter, each of which is programmed on a different code and switches for a different transmitted code.

In a typical application, control has been obtained over 20 ft.

## Adding a Latch Function

Figure 11 shows an application where the system latches itself as soon as a message has been received.

In the Idle mode, output 17 is high and Q<sub>1</sub> is on. The RC oscillator output pin is free running, and MM53200 is waiting for incoming code.

If a valid signal is received on Pin 16, output 17 goes low. Q<sub>1</sub> turns off, the RC network stops oscillating, and the MM53200 internal clock stops working. The MM53200 stays in the same position, i.e., output low. In our example, Q<sub>2</sub> will stay on and the LED will be on. The system is reset by resetting the clock by connecting the base of C<sub>1</sub> to V+ through a push-button oscillator starts again, and MM53200 waits for another call.

This latch function costs the user one npn transistor and one resistor. For example, it could be used to power on another remote-controlled function running from the same output decoder after receiving the correct "address" code.

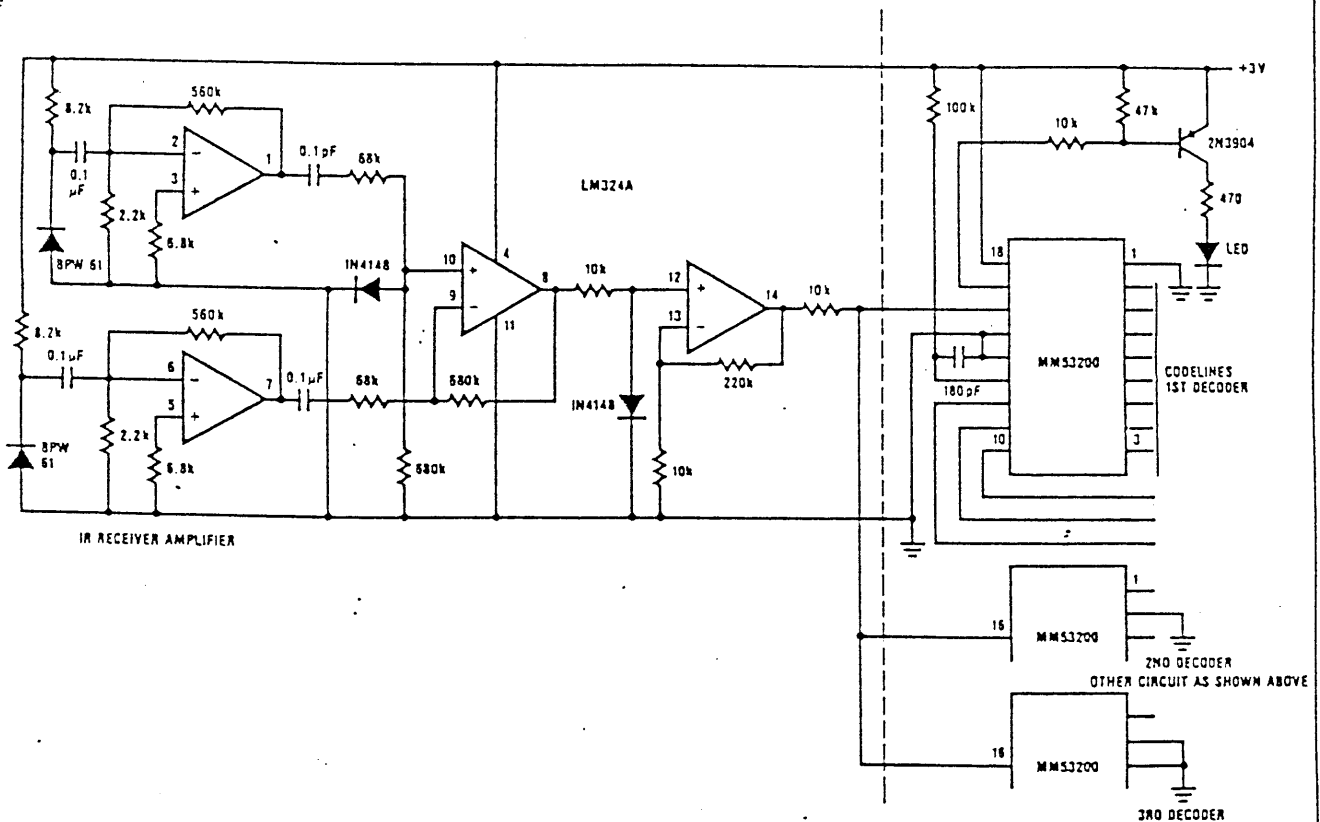


Figure 10. IR Receiver Application.

### A Latched On-Off Function

With the addition of a JK flip-flop, a 2-state sequential application is possible. Figure 12 shows how either of two outputs may be enabled with a single decoder. One output of the 74C73 flip-flop will force a change in data to the receiver, so, in order to switch back to the original state, the original code must be present.

method has to be chosen according to the application and the desired range. For short distances—up to 30 ft.—infrared or ultrasonic or visible light may be chosen for both indoor or outdoor applications. For longer distances, up to 500 ft., radio transmissions of either 49 or 300MHz may be used.

### Conclusion

MM53200 is a versatile encoder-decoder that finds its application in numerous cases. The type of transmitting

This is a very inexpensive and easy way to control functions, and the over 4,000 different combinations make it a very safe encoding system.

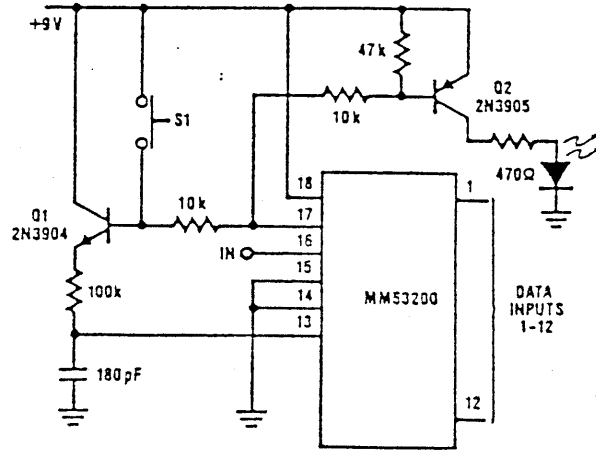


Figure 11. Latch Function.

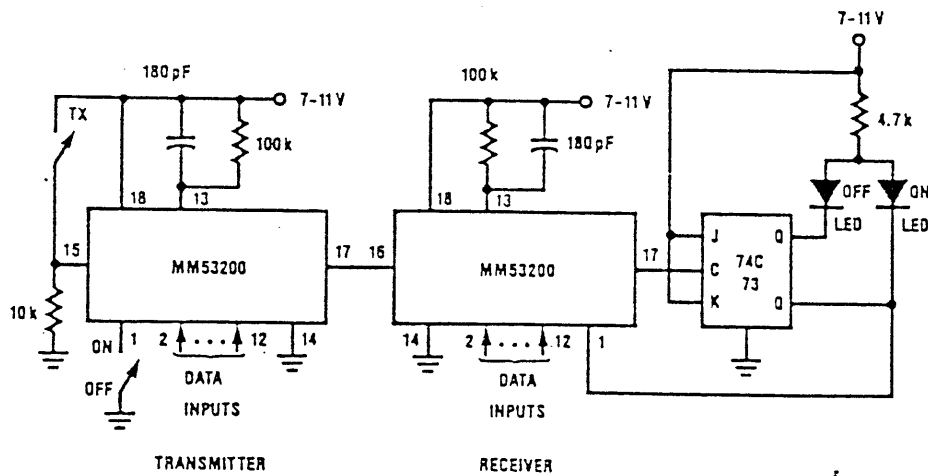


Figure 12. MM53200 Latched ON/OFF Applications.

### LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform, when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



National Semiconductor Corporation  
2900 Semiconductor Drive  
Santa Clara, California 95051  
Tel.: (408) 737-5000  
TWX: (910) 339-9240

National Semiconductor GmbH  
Fürstenriederstraße Nr. 51  
8 München 21  
West Germany  
Tel.: (089) 58 30 81  
Telex: 522772

NS Japan K.K.  
POB 4152,  
Shinjuku Center Building,  
1-25-1 Nishishinjuku,  
Shinjuku-ku,  
Tokyo 160, Japan  
Tel.: (03) 349-0811  
TWX: 232-2015 NSC J-J

National Semiconductor  
Hong Kong, Ltd.  
1st Floor,  
Cheung Kong Electronic Bldg.  
4 Hing Yip Street  
Kwun Tong  
Kowloon, Hong Kong  
Tel.: 3-899235  
Telex: 43866 MSEHK HX  
Cable: NATSEMI HX

NS Electronics De Brasil  
Avda. Brigadeiro Faria Lima 830  
8 Andar  
01452 Sao Paulo, Brasil  
Tel.:  
1121008 CABINE SAO PAULO  
113193 INSBR BR

NS Electronics Pty. Ltd.  
Cnr. Stud. Rd. & Min. Highway  
Bayswater, Victoria 3153  
Australia  
Tel.: 03-729-8333  
Telex: AA32098