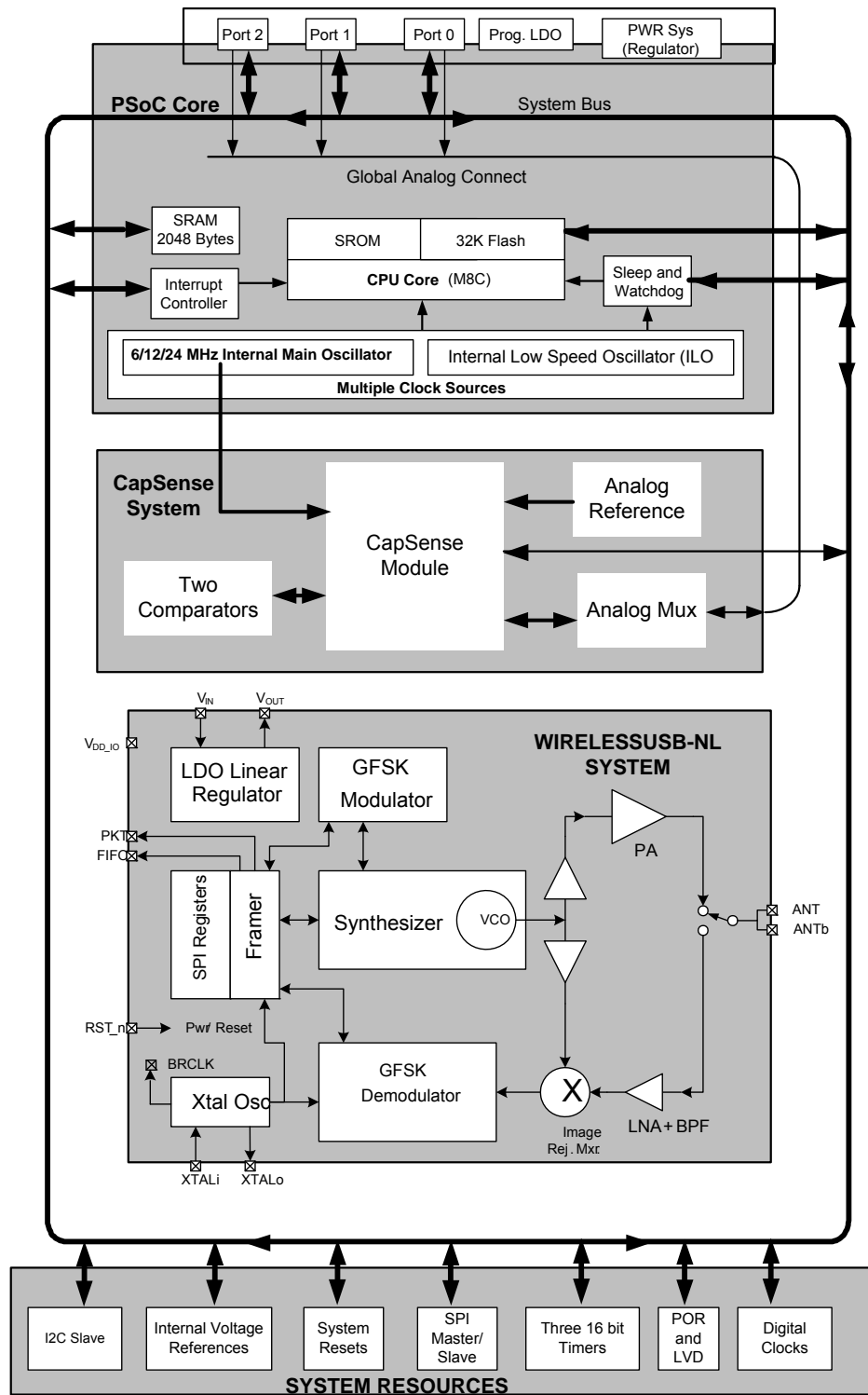


PRoC-CS Features

- **Single Device, Two functions**
 - 8-bit flash based CapSense controller MCU function and 2.4-GHz WirelessUSB™ NL radio transceiver function in a single device
- **Wide operating range: 1.9 V to 3.6 V**
 - Configurable capacitive sensing elements
 - 7 μ A per sensor at 500 ms scan rate
 - Supports SmartSense™ Auto-tuning
 - Supports a combination of CapSense® buttons, sliders, and proximity sensors
 - SmartSense_EMC offers superior noise immunity for applications with challenging conducted and radiated noise conditions
- **RF Attributes**
 - 2.4-GHz WirelessUSB-NL Transceiver function
 - Operates in the 2.4-GHz ISM Band (2.402 GHz - 2.479 GHz)
 - 1-Mbps over-the-air data rate
 - Receive sensitivity typical: -87 dBm
 - 1 μ A typical current consumption in sleep state
 - Closed-loop frequency synthesis
 - Supports frequency-hopping spread spectrum
 - On-chip packet framer with 64-byte first in first out (FIFO) data buffer
 - Built-in auto-retry-acknowledge protocol simplifies usage
 - Built-in cyclic redundancy check (CRC), forward error correction (FEC), data whitening
 - Additional outputs for interrupt request (IRQ) generation
 - Digital readout of received signal strength indication (RSSI)
- **MCU Attributes**
 - Powerful Harvard-architecture processor
 - M8C CPU – Up to 4 MIPS with 24 MHz Internal clock, external crystal resonator or clock signal
 - Low power at high speed
- **Temperature range: 0 °C to +70 °C**
- **Flexible on-chip memory**
 - 32 KB Flash/2 KB SRAM
 - 50,000 flash erase/write cycles
 - Partial flash updates
 - Flexible protection modes
- In-system serial programming (ISSP)
- **Precision, programmable clocking**
 - Internal main oscillator (IMO): 6/12/24 MHz \pm 5%
 - Internal low-speed oscillator (ILO) at 32 kHz for watchdog and sleep timers
 - Precision 32 kHz oscillator for optional external crystal
- **Programmable pin configurations**
 - Up to 13 general-purpose I/Os (GPIOs)
 - Dual mode GPIO: All GPIOs support digital I/O and analog inputs
 - 25-mA sink current on each GPIO
 - 120 mA total sink current on all GPIOs
 - Pull-up, high Z, open-drain modes on all GPIOs
 - CMOS drive mode -5 mA source current on ports 0 and 1 and 1 mA on port 2
 - 20 mA total source current on all GPIOs
- **Versatile analog system**
 - Low-dropout voltage regulator for all analog resources
 - Common internal analog bus enabling capacitive sensing on all pins
 - High power supply rejection ratio (PSRR) comparator
 - 8 to 10-bit incremental analog-to-digital converter (ADC)
- **Additional system resources**
 - I²C slave:
 - Selectable to 50 kHz, 100 kHz, or 400 kHz
 - SPI master and slave: Configurable 46.9 kHz to 12 MHz
 - Three 16-bit timers
 - Watchdog and sleep timers
 - Integrated supervisory circuit
 - Emulated E2PROM using flash memory
- **Complete development tools**
 - Free development tool (PSoC Designer™)
 - Full-featured, in-circuit emulator (ICE) and programmer
 - Full-speed emulation
 - Complex breakpoint structure
 - 128 KB trace memory
- **Package option**
 - 40-pin 6 mm \times 6 mm QFN

Logical Block Diagram



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PSoC[®] Functional Overview

The PSoC family consists of on-chip controller devices, which are designed to replace multiple traditional microcontroller unit (MCU)-based components with one, low cost single-chip programmable component. A PSoC device includes configurable analog and digital blocks, and programmable interconnect. This architecture allows the user to create customized peripheral configurations, to match the requirements of each individual application. Additionally, a fast CPU, flash program memory, SRAM data memory, and configurable I/O are included in a range of convenient pinouts.

The architecture for this device family, as shown in the [Logical Block Diagram on page 2](#), consists of three main areas:

- The Core
- CapSense Analog System
- WirelessUSB-NL System
- System Resources.

A common, versatile bus allows connection between I/O and the analog system.

Each CYRF89435 device includes a dedicated CapSense block that provides sensing and scanning control circuitry for capacitive sensing applications. The 13 GPIOs provide access to the MCU and analog mux.

PSoC Core

The PSoC Core is a powerful engine that supports a rich instruction set. It encompasses SRAM for data storage, an interrupt controller, sleep and watchdog timers, and IMO and ILO. The CPU core, called the M8C, is a powerful processor with speeds up to 24 MHz. The M8C is a 4-MIPS, 8-bit Harvard-architecture microprocessor.

CapSense System

The analog system contains the capacitive sensing hardware. Several hardware algorithms are supported. This hardware performs capacitive sensing and scanning without requiring external components. The analog system is composed of the CapSense PSoC block and an internal 1 V or 1.2 V analog reference, which together support capacitive sensing of up to 13 inputs. Capacitive sensing is configurable on each GPIO pin. Scanning of enabled CapSense pins are completed quickly and easily across multiple ports.

SmartSense

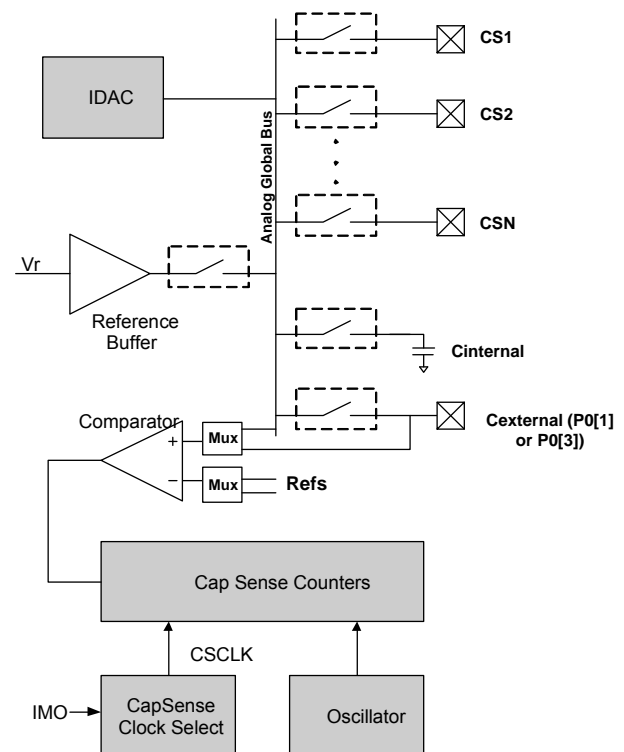
SmartSense is an innovative solution from Cypress that removes manual tuning of CapSense applications. This solution is easy to use and provides a robust noise immunity. It is the only auto-tuning solution that establishes, monitors, and maintains all required tuning parameters. SmartSense allows engineers to go

from prototyping to mass production without re-tuning for manufacturing variations in PCB and/or overlay material properties.

SmartSense_EMCC

In addition to the SmartSense auto-tuning algorithm to remove manual tuning of CapSense applications, SmartSense EMC user module incorporates a unique algorithm to improve robustness of capacitive sensing algorithm/circuit against high frequency conducted and radiated noise. Every electronic device must comply with specific limits for radiated and conducted external noise and these limits are specified by regulatory bodies (for example, FCC, CE, U/L and so on). A very good PCB layout design, power supply design and system design is a mandatory for a product to pass the conducted and radiated noise tests. An ideal PCB layout, power supply design or system design is not often possible because of cost and form factor limitations of the product. SmartSense EMC with superior noise immunity is well suited and handy for such applications to pass radiated and conducted noise test.

Figure 1. CapSense System Block Diagram



Analog Multiplexer System

The Analog Mux Bus can connect to every GPIO pin. Pins are connected to the bus individually or in any combination. The bus also connects to the analog system for analysis with the CapSense block comparator.

Switch control logic enables selected pins to precharge continuously under hardware control. This enables capacitive measurement for applications such as touch sensing. Other multiplexer applications include:

- Complex capacitive sensing interfaces, such as sliders and touchpads.
- Chip-wide mux that allows analog input from any I/O pin.
- Crosspoint connection between any I/O pin combinations.

WirelessUSB-NL System

WirelessUSB-NL, optimized to operate in the 2.4-GHz ISM band, is Cypress's third generation of 2.4-GHz low-power RF technology. WirelessUSB-NL implements a Gaussian frequency-shift keying (GFSK) radio using a differentiated single-mixer, closed-loop modulation design that optimizes power efficiency and interference immunity. Closed-loop modulation effectively eliminates the problem of frequency drift, enabling WirelessUSB-NL to transmit up to 255-byte payloads without repeatedly having to pay power penalties for re-locking the phase-locked loop (PLL) as in open-loop designs

Among the advantages of WirelessUSB-NL are its fast lock times and channel switching, along with the ability to transmit larger payloads. Use of longer payload packets, compared to multiple short payload packets, can reduce overhead, improve overall power efficiency, and help alleviate spectrum crowding.

Combined with Cypress's Capacitive touch sense controllers, WirelessUSB-NL also provides the lowest bill of materials (BOM) cost solution for sophisticated PC peripheral applications such as wireless keyboards and mice, as well as best-in-class wireless performance in other demanding applications. such as toys, remote controls, fitness, automation, presenter tools, and gaming.

With PProC-CS, the WirelessUSB-NL transceiver can add wireless capability to a wide variety of CapSense applications.

The WirelessUSB-NL is a fully-integrated CMOS RF transceiver, GFSK data modem, and packet framer, optimized for use in the 2.4-GHz ISM band. It contains transmit, receive, RF synthesizer, and digital modem functions, with few external components. The transmitter supports digital power control. The receiver uses extensive digital processing for excellent overall performance, even in the presence of interference and transmitter impairments.

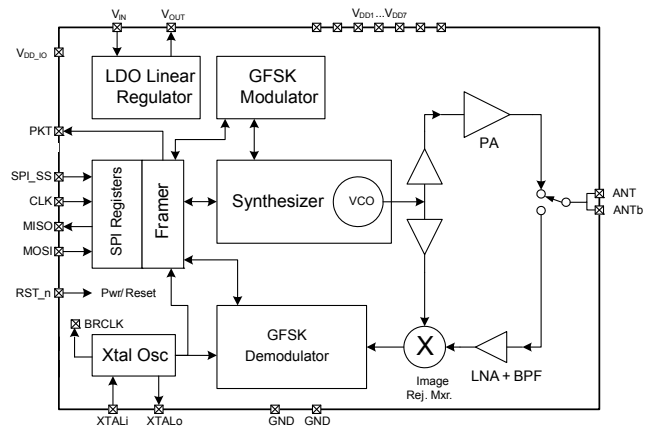
The product transmits GFSK data at approximately 0-dBm output power. Sigma-Delta PLL delivers high-quality DC-coupled transmit data path.

The low-IF receiver architecture produces good selectivity and image rejection, with typical sensitivity of -87 dBm or better on most channels. Sensitivity on channels that are integer multiples of the crystal reference oscillator frequency (12 MHz) may show approximately 5 dB degradation. Digital RSSI values are available to monitor channel quality.

On-chip transmit and receive FIFO registers are available to buffer the data transfer with MCU. Over-the-air data rate is always 1 Mbps even when connected to a slow, low-cost MCU. Built-in CRC, FEC, data whitening, and automatic retry/acknowledge are all available to simplify and optimize performance for individual applications.

For more details on the radio's implementation details and timing requirements, please go through the WirelessUSB-NL datasheet in www.cypress.com.

Figure 2. WirelessUSB-NL logic Block Diagram



Transmit Power Control

The following table lists recommended settings for register 9 for short-range applications, where reduced transmit RF power is a desirable trade off for lower current.

Table 1. Transmit Power Control

Power Setting Description	Typical Transmit Power (dBm)	Value of Register 9	
		Silicon ID 0x1002	Silicon ID 0x2002
PA0 - Highest power	+1	0x1820	0x7820
PA2 - High power	0	0x1920	0x7920
PA4 - High power	−3	0x1A20	0x7A20
PA8 - Low power	−7.5	0x1C20	0x7C20
PA12 - Lower power	−11.2	0x1E20	0x7E20

Note: Silicon ID can be read from Register 31.

Power-on and Register Initialization Sequence

For proper initialization at power up, V_{IN} must ramp up at the minimum overall ramp rate no slower than shown by T_{VIN} specification in the following figure. During this time, the RST_n line must track the V_{IN} voltage ramp-up profile to within approximately 0.2 V. Since most MCU GPIO pins automatically default to a high-Z condition at power up, it only requires a pull-up resistor. When power is stable and the MCU POR releases, and MCU begins to execute instructions, RST_n must then be pulsed low as shown in [Figure 13 on page 31](#), followed by writing Reg[27 = 0x4200. During or after this SPI transaction, the State Machine status can be read to confirm FRAMER_ST= 1, indicating a proper initialization.

Additional System Resources

System resources provide additional capability, such as configurable I²C slave, SPI master/slave communication interface, three 16-bit programmable timers, and various system resets supported by the M8C.

These system resources provide additional capability useful to complete systems. Additional resources include low voltage detection and power-on reset. The merits of each system resource are listed here:

- The I²C slave/SPI master-slave module provides 50/100/400 kHz communication over two wires. SPI communication over three or four wires runs at speeds of 46.9 kHz to 3 MHz (lower for a slower system clock).
- Low-voltage detection (LVD) interrupts can signal the application of falling voltage levels, while the advanced power-on reset (POR) circuit eliminates the need for a system supervisor.
- An internal reference provides an absolute reference for capacitive sensing.
- A register-controlled bypass mode allows the user to disable the LDO regulator.

Getting Started

The quickest way to understand the P_{RoC}-CS silicon is to read this datasheet and then use the PSoC Designer Integrated Development Environment (IDE). This datasheet is an overview of the PSoC integrated circuit and presents specific pin, register, and electrical specifications.

For in depth information, along with detailed programming details, see the [Technical Reference Manual](#) for the CapSense devices.

For up-to-date ordering, packaging, and electrical specification information, see the latest PSoC device datasheets on the web at www.cypress.com/psoc.

CapSense Design Guides

Design Guides are an excellent introduction to the wide variety of possible CapSense designs. They are located at www.cypress.com/go/CapSenseDesignGuides.

Refer Getting Started with CapSense design guide for information on CapSense design and CY8C20XX6A/H/AS CapSense[®] Design Guide for specific information on P_{RoC}-CS controllers.

CYPros Consultants

Certified PSoC consultants offer everything from technical assistance to completed PSoC designs. To contact or become a PSoC consultant go to the [CYPros Consultants](#) web site.

Solutions Library

Visit our growing [library of solution focused designs](#). Here you can find various application designs that include firmware and hardware design files that enable you to complete your designs quickly.

Technical Support

[Technical support](#) – including a searchable Knowledge Base articles and technical forums – is also available online. If you cannot find an answer to your question, call our Technical Support hotline at 1-800-541-4736.

Development Tools

PSoC Designer™ is the revolutionary integrated design environment (IDE) that you can use to customize PSoC to meet your specific application requirements. PSoC Designer software accelerates system design and time to market. Develop your applications using a library of precharacterized analog and digital peripherals (called user modules) in a drag-and-drop design environment. Then, customize your design by leveraging the dynamically generated application programming interface (API) libraries of code. Finally, debug and test your designs with the integrated debug environment, including in-circuit emulation and standard software debug features. PSoC Designer includes:

- Application editor graphical user interface (GUI) for device and user module configuration and dynamic reconfiguration
- Extensive user module catalog
- Integrated source-code editor (C and assembly)
- Free C compiler with no size restrictions or time limits
- Built-in debugger
- In-circuit emulation
- Built-in support for communication interfaces:
 - Hardware and software I²C slaves and masters
 - SPI master and slave, and wireless

PSoC Designer supports the entire library of PSoC 1 devices and runs on Windows XP, Windows Vista, and Windows 7.

PSoC Designer Software Subsystems

Design Entry

In the chip-level view, choose a base device to work with. Then select different onboard analog and digital components that use the PSoC blocks, which are called user modules. Examples of user modules are analog-to-digital converters (ADCs), digital-to-analog converters (DACs), amplifiers, and filters. Configure the user modules for your chosen application and connect them to each other and to the proper pins. Then generate your project. This prepopulates your project with APIs and libraries that you can use to program your application.

The tool also supports easy development of multiple configurations and dynamic reconfiguration. Dynamic reconfiguration makes it possible to change configurations at run time. In essence, this lets you to use more than 100 percent of PSoC's resources for an application.

Code Generation Tools

The code generation tools work seamlessly within the PSoC Designer interface and have been tested with a full range of debugging tools. You can develop your design in C, assembly, or a combination of the two.

Assemblers. The assemblers allow you to merge assembly code seamlessly with C code. Link libraries automatically use absolute addressing or are compiled in relative mode, and linked with other software modules to get absolute addressing.

C Language Compilers. C language compilers are available that support the PSoC family of devices. The products allow you to create complete C programs for the PSoC family devices. The optimizing C compilers provide all of the features of C, tailored to the PSoC architecture. They come complete with embedded libraries providing port and bus operations, standard keypad and display support, and extended math functionality.

Debugger

PSoC Designer has a debug environment that provides hardware in-circuit emulation, allowing you to test the program in a physical system while providing an internal view of the PSoC device. Debugger commands allow you to read and program and read and write data memory, and read and write I/O registers. You can read and write CPU registers, set and clear breakpoints, and provide program run, halt, and step control. The debugger also lets you to create a trace buffer of registers and memory locations of interest.

Online Help System

The online help system displays online, context-sensitive help. Designed for procedural and quick reference, each functional subsystem has its own context-sensitive help. This system also provides tutorials and links to FAQs and an Online Support Forum to aid the designer.

In-Circuit Emulator

A low-cost, high-functionality in-circuit emulator (ICE) is available for development support. This hardware can program single devices.

The emulator consists of a base unit that connects to the PC using a USB port. The base unit is universal and operates with all PSoC devices. Emulation pods for each device family are available separately. The emulation pod takes the place of the PSoC device in the target board and performs full-speed (24 MHz) operation.

Designing with PSoC Designer

The development process for the PSoC device differs from that of a traditional fixed-function microprocessor. The configurable analog and digital hardware blocks give the PSoC architecture a unique flexibility that pays dividends in managing specification change during development and lowering inventory costs. These configurable resources, called PSoC blocks, have the ability to implement a wide variety of user-selectable functions. The PSoC development process is:

1. Select [user modules](#).
2. Configure user modules.
3. Organize and connect.
4. Generate, verify, and debug.

Select User Modules

PSoC Designer provides a library of prebuilt, pretested hardware peripheral components called “user modules”. User modules make selecting and implementing peripheral devices, both analog and digital, simple.

Configure User Modules

Each user module that you select establishes the basic register settings that implement the selected function. They also provide parameters and properties that allow you to tailor their precise configuration to your particular application. For example, a PWM User Module configures one or more digital PSoC blocks, one for each eight bits of resolution. Using these parameters, you can establish the pulse width and duty cycle. Configure the parameters and properties to correspond to your chosen application. Enter values directly or by selecting values from drop-down menus. All of the user modules are documented in datasheets that may be viewed directly in PSoC Designer or on the Cypress website. These [user module datasheets](#) explain the

internal operation of the user module and provide performance specifications. Each datasheet describes the use of each user module parameter, and other information that you may need to successfully implement your design.

Organize and Connect

Build signal chains at the chip level by interconnecting user modules to each other and the I/O pins. Perform the selection, configuration, and routing so that you have complete control over all on-chip resources.

Generate, Verify, and Debug

When you are ready to test the hardware configuration or move on to developing code for the project, perform the “Generate Configuration Files” step. This causes PSoC Designer to generate source code that automatically configures the device to your specification and provides the software for the system. The generated code provides APIs with high-level functions to control and respond to hardware events at run time, and interrupt service routines that you can adapt as needed.

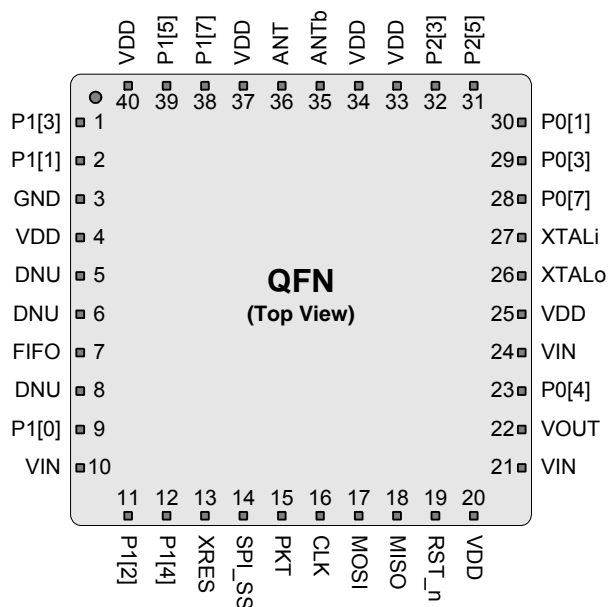
A complete code development environment lets you to develop and customize your applications in C, assembly language, or both.

The last step in the development process takes place inside PSoC Designer's Debugger (accessed by clicking the Connect icon). PSoC Designer downloads the HEX image to the ICE where it runs at full-speed. PSoC Designer debugging capabilities rival those of systems costing many times more. In addition to traditional single-step, run-to-breakpoint, and watch-variable features, the debug interface provides a large trace buffer. The interface lets you to define complex breakpoint events that include monitoring address and data bus values, memory locations, and external signals.

Pinouts

The CYRF89435 P_{RoC}-CS device is available in a 40-pin QFN package, which is illustrated in the following table. Every port pin (labeled with a “P”) is capable of Digital I/O and connection to the common analog bus. However, V_{DD}, and XRES are not capable of Digital I/O.

Figure 3. 40-pin QFN pinout



Pin Definitions

Pin No	Pin name	Pin Description
1	P1[3]/SCLK ^[2]	Digital I/O, Analog I/O, SPI CLK
2	P1[1]/MOSI ^[1]	Digital I/O, Analog I/O, TC CLK, I2C SCL, SPI MOSI
3	GND	Ground connection
4, 20, 25, 33, 34, 37, 40	VDD	Core power supply voltage. Connect all VDD pins to VOUT pin.
5	DNU	Do not use
6	DNU	Do not use
7	FIFO	FIFO status indicator bit
8	DNU	Do not use
9	P1[0] ^[1]	Analog I/O, Digital I/O, TC DATA, I2C SDA
10, 21, 24	VIN	Unregulated input voltage to the on-chip low drop out (LDO) voltage regulator
11	P1[2]	Analog I/O, Digital I/O
12	P1[4]	Analog I/O, Digital I/O, EXT CLK
13	XRES	Active high external reset with internal pull-down
14	SPI_SS	Enable input for SPI, active low. Also used to bring device out of sleep state.
15	PKT	Transmit/receive packet status indicator bit
16	SPI_CLK	Clock input for SPI interface
17	SPI_MOSI	Data input for the SPI bus
18	SPI_MISO	Data output (tristate when not active)
19	RST_n	RST_n Low: Chip shutdown to conserve power. Register values lost RST_n High: Turn on chip, registers restored to default value
22	VOUT	1.8 V output from on-chip LDO. Connect to all VDD pins, do not connect to external loads.
23	P0[4]	Analog I/O, Digital I/O, VREF
26	XTALO	Output of the crystal oscillator gain block
27	XTALI	Input to the crystal oscillator gain block
28	P0[7]	Analog I/O, Digital I/O, SPI CLK
29	P0[3]	Analog I/O, Digital I/O, Integrating input
30	P0[1]	Analog I/O, Digital I/O, Integrating input
31	P2[5]	Analog I/O, Digital I/O, XTAL Out
32	P2[3]	Analog I/O, Digital I/O, XTAL In
35	ANTb	Differential RF input/output. Each of these pins must be DC grounded, 20 kΩ or less
36	ANT	Differential RF input/output. Each of these pins must be DC grounded, 20 kΩ or less
38	P1[7]/SS_N	Digital I/O, Analog I/O, I2C SCL, SPI SS
39	P1[5]/MISO	Digital I/O, Analog I/O, I2C SDA, SPI MISO

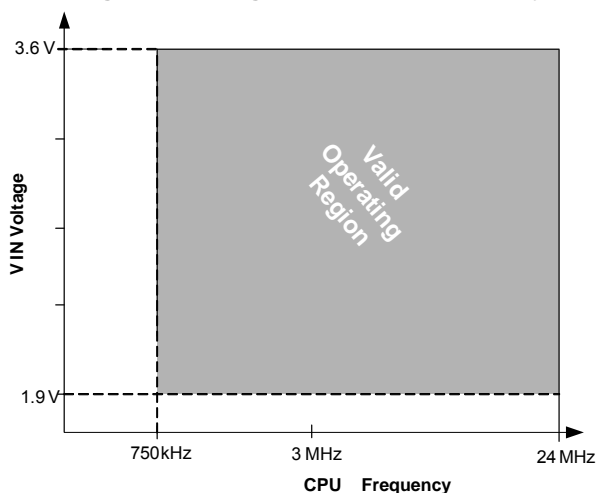
Notes

- On power-up, the SDA(P1[0]) drives a strong high for 256 sleep clock cycles and drives resistive low for the next 256 sleep clock cycles. The SCL(P1[1]) line drives resistive low for 512 sleep clock cycles and both the pins transition to high impedance state. On reset, after XRES de-asserts, the SDA and the SCL lines drive resistive low for 8 sleep clock cycles and transition to high impedance state. Hence, during power-up or reset event, P1[1] and P1[0] may disturb the I2C bus. Use alternate pins if you encounter issues.
- Alternate SPI clock.

Electrical Specifications – PSoC Core

This section presents the DC and AC electrical specifications of the CYRF89435 PSoC devices. For the latest electrical specifications, confirm that you have the most recent datasheet by visiting the web at <http://www.cypress.com/psoc>.

Figure 4. Voltage versus CPU Frequency



Absolute Maximum Ratings

Exceeding maximum ratings may shorten the useful life of the device. User guidelines are not tested.

Table 2. Absolute Maximum Ratings

Symbol	Description	Conditions	Min	Typ	Max	Units
T _{STG}	Storage temperature	Higher storage temperatures reduce data retention time. Recommended Storage Temperature is +25 °C ± 25 °C. Extended duration storage temperatures above 85 °C degrades reliability.	-55	25	125	°C
V _{IN} ^[3]		—	1.9	—	3.63	V
V _{IO}	DC input voltage	—	-0.5	—	V _{IN} + 0.5	V
V _{IOZ} ^[4]	DC voltage applied to tristate	—	-0.5	—	V _{IN} + 0.5	V
I _{MIO}	Maximum current into any port pin	—	-25	—	+50	mA
ESD	Electrostatic discharge voltage	Human body model ESD i) RF pins (ANT, ANTb) ii) Analog pins (XTALi, XTALo) iii) Remaining pins	500 500 2000	—	—	V
LU	Latch-up current	In accordance with JESD78 standard	—	—	140	mA

Operating Temperature

Table 3. Operating Temperature

Symbol	Description	Conditions	Min	Typ	Max	Units
T _A	Ambient temperature	—	0	—	70	°C

Notes

3. Program the device at 3.3 V only. Hence use MiniProg3 only as MiniProg1 does not support programming at 3.3 V.
4. Port1 pins are hot-swap capable with I/O configured in High-Z mode, and pin input voltage above V_{IN}.

DC Chip-Level Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 4. DC Chip-Level Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
$V_{IN}^{[5, 6, 7, 8]}$	Supply voltage	Refer the table DC POR and LVD Specifications on page 17	1.9	–	3.6	V
I_{DD24}	Supply current, IMO = 24 MHz	Conditions are $V_{IN} \leq 3.0$ V, $T_A = 25$ °C, CPU = 24 MHz. CapSense running at 12 MHz, no I/O sourcing current	–	2.88	4.00	mA
I_{DD12}	Supply current, IMO = 12 MHz	Conditions are $V_{IN} \leq 3.0$ V, $T_A = 25$ °C, CPU = 12 MHz. CapSense running at 12 MHz, no I/O sourcing current	–	1.71	2.60	mA
I_{DD6}	Supply current, IMO = 6 MHz	Conditions are $V_{IN} \leq 3.0$ V, $T_A = 25$ °C, CPU = 6 MHz. CapSense running at 6 MHz, no I/O sourcing current	–	1.16	1.80	mA
$I_{DDAVG10}$	Average supply current per sensor	One sensor scanned at 10 ms rate	–	250	–	μA
$I_{DDAVG100}$	Average supply current per sensor	One sensor scanned at 100 ms rate	–	25	–	μA
$I_{DDAVG500}$	Average supply current per sensor	One sensor scanned at 500 ms rate	–	7	–	μA
I_{SB0}	Deep sleep current	$V_{IN} \leq 3.0$ V, $T_A = 25$ °C, I/O regulator turned off	–	0.10	1.05	μA
I_{SB1}	Standby current with POR, LVD and sleep timer	$V_{IN} \leq 3.0$ V, $T_A = 25$ °C, I/O regulator turned off	–	1.07	1.50	μA
I_{SBI2C}	Standby current with I ² C enabled	Conditions are $V_{IN} = 3.3$ V, $T_A = 25$ °C and CPU = 24 MHz	–	1.64	–	μA

Notes

- If powering down in standby sleep mode, to properly detect and recover from a V_{IN} brown out condition any of the following actions must be taken:
 Bring the device out of sleep before powering down.
 Assure that V_{IN} falls below 100 mV before powering back up.
 Set the No Buzz bit in the OSC_CR0 register to keep the voltage monitoring circuit powered during sleep.
 Increase the buzz rate to assure that the falling edge of V_{IN} is captured. The rate is configured through the PSSDC bits in the SLP_CFG register.
 For the referenced registers, refer to the *CY8C20X36 Technical Reference Manual*. In deep sleep mode, additional low power voltage monitoring circuitry allows V_{IN} brown out conditions to be detected for edge rates slower than 1V/ms.
- Always greater than 50 mV above V_{PPOR1} voltage for falling supply.
- Always greater than 50 mV above V_{PPOR2} voltage for falling supply.
- Always greater than 50 mV above V_{PPOR3} voltage for falling supply.

DC GPIO Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 2.4 V to 3.0 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, or 1.9 V to 2.4 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, respectively. Typical parameters apply to 3.3 V at 25°C and are for design guidance only.

Table 5. 2.4 V to 3.0 V DC GPIO Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
R_{PU}	Pull-up resistor	–	4	5.60	8	$k\Omega$
V_{OH1}	High output voltage Port 2 or 3 or 4 pins	$I_{OH} < 10\ \mu\text{A}$, maximum of 10 mA source current in all I/Os	$V_{IN} - 0.20$	–	–	V
V_{OH2}	High output voltage Port 2 or 3 or 4 pins	$I_{OH} = 0.2\ \text{mA}$, maximum of 10 mA source current in all I/Os	$V_{IN} - 0.40$	–	–	V
V_{OH3}	High output voltage Port 0 or 1 pins with LDO regulator Disabled for port 1	$I_{OH} < 10\ \mu\text{A}$, maximum of 10 mA source current in all I/Os	$V_{IN} - 0.20$	–	–	V
V_{OH4}	High output voltage Port 0 or 1 pins with LDO regulator Disabled for Port 1	$I_{OH} = 2\ \text{mA}$, maximum of 10 mA source current in all I/Os	$V_{IN} - 0.50$	–	–	V
V_{OH5A}	High output voltage Port 1 pins with LDO enabled for 1.8 V out	$I_{OH} < 10\ \mu\text{A}$, $V_{IN} > 2.4\ \text{V}$, maximum of 20 mA source current in all I/Os	1.50	1.80	2.10	V
V_{OH6A}	High output voltage Port 1 pins with LDO enabled for 1.8 V out	$I_{OH} = 1\ \text{mA}$, $V_{IN} > 2.4\ \text{V}$, maximum of 20 mA source current in all I/Os	1.20	–	–	V
V_{OL}	Low output voltage	$I_{OL} = 10\ \text{mA}$, maximum of 30 mA sink current on even port pins (for example, P0[2] and P1[4]) and 30 mA sink current on odd port pins (for example, P0[3] and P1[5])	–	–	0.75	V
V_{IL}	Input low voltage	–	–	–	0.72	V
V_{IH}	Input high voltage	–	1.40	–	–	V
V_H	Input hysteresis voltage	–	–	80	–	mV
I_{IL}	Input leakage (absolute value)	–	–	1	1000	nA
C_{PIN}	Capacitive load on pins	Package and pin dependent Temp = 25°C	0.50	1.70	7	pF
$V_{ILLVT2.5}$	Input Low Voltage with low threshold enable set, Enable for Port1	Bit3 of IO_CFG1 set to enable low threshold voltage of Port1 input	0.7	–	–	V
$V_{IHLVT2.5}$	Input High Voltage with low threshold enable set, Enable for Port1	Bit3 of IO_CFG1 set to enable low threshold voltage of Port1 input	1.2	–	–	V

Table 6. 1.9 V to 2.4 V DC GPIO Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
R _{PU}	Pull-up resistor	–	4	5.60	8	kΩ
V _{OH1}	High output voltage Port 2 or 3 or 4 pins	I _{OH} = 10 μA, maximum of 10 mA source current in all I/Os	V _{IN} – 0.20	–	–	V
V _{OH2}	High output voltage Port 2 or 3 or 4 pins	I _{OH} = 0.5 mA, maximum of 10 mA source current in all I/Os	V _{IN} – 0.50	–	–	V
V _{OH3}	High output voltage Port 0 or 1 pins with LDO regulator Disabled for Port 1	I _{OH} = 100 μA, maximum of 10 mA source current in all I/Os	V _{IN} – 0.20	–	–	V
V _{OH4}	High output voltage Port 0 or 1 Pins with LDO Regulator Disabled for Port 1	I _{OH} = 2 mA, maximum of 10 mA source current in all I/Os	V _{IN} – 0.50	–	–	V
V _{OL}	Low output voltage	I _{OL} = 5 mA, maximum of 20 mA sink current on even port pins (for example, P0[2] and P1[4]) and 30 mA sink current on odd port pins (for example, P0[3] and P1[5])	–	–	0.40	V
V _{IL}	Input low voltage	–	–	–	0.30 × V _{IN}	V
V _{IH}	Input high voltage	–	0.65 × V _{IN}	–	–	V
V _H	Input hysteresis voltage	–	–	80	–	mV
I _{IL}	Input leakage (absolute value)	–	–	1	1000	nA
C _{PIN}	Capacitive load on pins	Package and pin dependent temp = 25 °C	0.50	1.70	7	pF

Analog DC Mux Bus Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 7. DC Analog Mux Bus Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
R _{SW}	Switch resistance to common analog bus	–	–	–	800	Ω
R _{GND}	Resistance of initialization switch to GND	–	–	–	800	Ω

The maximum pin voltage for measuring R_{SW} and R_{GND} is 1.8 V

DC Low Power Comparator Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 8. DC Comparator Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V _{LPC}	Low power comparator (LPC) common mode	Maximum voltage limited to V _{IN}	0.0	–	1.8	V
I _{LPC}	LPC supply current	–	–	10	40	μA
V _{OSLPC}	LPC voltage offset	–	–	3	30	mV

Comparator User Module Electrical Specifications

The following table lists the guaranteed maximum and minimum specifications. Unless stated otherwise, the specifications are for the entire device voltage and temperature operating range: $0\text{ }^{\circ}\text{C} \leq T_A \leq 70\text{ }^{\circ}\text{C}$, $1.9\text{ V} \leq V_{IN} \leq 3.6\text{ V}$.

Table 9. Comparator User Module Electrical Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
t_{COMP}	Comparator response time	50 mV overdrive	–	70	100	ns
Offset		Valid from 0.2 V to $V_{IN} - 0.2\text{ V}$	–	2.5	30	mV
Current		Average DC current, 50 mV overdrive	–	20	80	μA
PSRR	Supply voltage > 2 V	Power supply rejection ratio	–	80	–	dB
	Supply voltage < 2 V	Power supply rejection ratio	–	40	–	dB
Input range		–	0		1.5	V

ADC Electrical Specifications
Table 10. ADC User Module Electrical Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
Input						
V_{IN}	Input voltage range	–	0	–	V_{REFADC}	V
C_{IIN}	Input capacitance	–	–	–	5	pF
R_{IN}	Input resistance	Equivalent switched cap input resistance for 8-, 9-, or 10-bit resolution	$1/(500fF \times \text{data clock})$	$1/(400fF \times \text{data clock})$	$1/(300fF \times \text{data clock})$	Ω
Reference						
V_{REFADC}	ADC reference voltage	–	1.14	–	1.26	V
Conversion Rate						
F_{CLK}	Data clock	Source is chip's internal main oscillator. See AC Chip-Level Specifications for accuracy	2.25	–	6	MHz
S8	8-bit sample rate	Data clock set to 6 MHz. Sample rate = $0.001 / (2^{\text{Resolution}} / \text{Data Clock})$	–	23.43	–	ksp/s
S10	10-bit sample rate	Data clock set to 6 MHz. Sample rate = $0.001 / (2^{\text{resolution}} / \text{data clock})$	–	5.85	–	ksp/s
DC Accuracy						
RES	Resolution	Can be set to 8-, 9-, or 10-bit	8	–	10	bits
DNL	Differential nonlinearity	–	–1	–	+2	LSB
INL	Integral nonlinearity	–	–2	–	+2	LSB
E_{OFFSET}	Offset error	8-bit resolution	0	3.20	19.20	LSB
		10-bit resolution	0	12.80	76.80	LSB
E_{GAIN}	Gain error	For any resolution	–5	–	+5	%FSR
Power						
I_{ADC}	Operating current	–	–	2.10	2.60	mA
PSRR	Power supply rejection ratio	PSRR ($V_{IN} > 3.0\text{ V}$)	–	24	–	dB
		PSRR ($V_{IN} < 3.0\text{ V}$)	–	30	–	dB

DC POR and LVD Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 11. DC POR and LVD Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V _{POR1}	2.36 V selected in PSoC Designer	V _{IN} must be greater than or equal to 1.9 V during startup, reset from the XRES pin, or reset from watchdog.	–	2.36	2.41	V
V _{POR2}	2.60 V selected in PSoC Designer		–	2.60	2.66	
V _{POR3}	2.82 V selected in PSoC Designer		–	2.82	2.95	
V _{LVD0}	2.45 V selected in PSoC Designer	–	2.40	2.45	2.51	V
V _{LVD1}	2.71 V selected in PSoC Designer		2.64 ^[9]	2.71	2.78	
V _{LVD2}	2.92 V selected in PSoC Designer		2.85 ^[10]	2.92	2.99	
V _{LVD3}	3.02 V selected in PSoC Designer		2.95 ^[11]	3.02	3.09	
V _{LVD4}	3.13 V selected in PSoC Designer		3.06	3.13	3.20	
V _{LVD5}	1.90 V selected in PSoC Designer		1.84	1.90	2.32	

DC Programming Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 12. DC Programming Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V _{IN}	Supply voltage for flash write operations	–	1.91	–	3.6	V
I _{DDP}	Supply current during programming or verify	–	–	5	25	mA
V _{ILP}	Input low voltage during programming or verify	See the appropriate DC GPIO Specifications on page 13	–	–	V _{IL}	V
V _{IHP}	Input high voltage during programming or verify	See the appropriate DC GPIO Specifications on page 13	V _{IH}	–	–	V
I _{ILP}	Input current when Applying V _{ILP} to P1[0] or P1[1] during programming or verify	Driving internal pull-down resistor	–	–	0.2	mA
I _{IHP}	Input current when applying V _{IHP} to P1[0] or P1[1] during programming or verify	Driving internal pull-down resistor	–	–	1.5	mA
V _{OLP}	Output low voltage during programming or verify		–	–	+ 0.75	V
V _{OHP}	Output high voltage during programming or verify	See appropriate DC GPIO Specifications on page 13 . For V _{IN} > 3 V use V _{OHP} in Table 3 on page 11 .	V _{OH}	–	V _{IN}	V
Flash _{ENPB}	Flash write endurance	Erase/write cycles per block	50,000	–	–	–
Flash _{DR}	Flash data retention	Following maximum Flash write cycles; ambient temperature of 55 °C	20	–	–	Years

Notes

9. Always greater than 50 mV above V_{PPOR1} voltage for falling supply.
10. Always greater than 50 mV above V_{PPOR2} voltage for falling supply.
11. Always greater than 50 mV above V_{PPOR3} voltage for falling supply.

DC I²C Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 3, 2.4 V to 3.0 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, or 1.9 V to 2.4 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, respectively. Typical parameters apply to 3.3 V at 25°C and are for design guidance only.

Table 13. DC I²C Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V _{ILI2C}	Input low level	$3.1\text{ V} \leq V_{IN} \leq 3.6\text{ V}$	–	–	$0.25 \times V_{IN}$	V
		$2.5\text{ V} \leq V_{IN} \leq 3.0\text{ V}$	–	–	$0.3 \times V_{IN}$	V
		$1.9\text{ V} \leq V_{IN} \leq 2.4\text{ V}$	–	–	$0.3 \times V_{IN}$	V
V _{IHI2C}	Input high level	$1.9\text{ V} \leq V_{IN} \leq 3.6\text{ V}$	$0.65 \times V_{IN}$	–	–	V

DC Reference Buffer Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 2.4 V to 3.0 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, or 1.9 V to 2.4 V and $0^{\circ}\text{C} \leq T_A \leq 70^{\circ}\text{C}$, respectively. Typical parameters apply to 3.3 V at 25°C and are for design guidance only.

Table 14. DC Reference Buffer Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
V _{Ref}	Reference buffer output	1.9 V to 3.6 V	1	–	1.05	V
V _{RefHi}	Reference buffer output	1.9 V to 3.6 V	1.2	–	1.25	V

DC IDAC Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 15. DC IDAC Specifications

Symbol	Description	Min	Typ	Max	Units	Notes
IDAC_DNL	Differential nonlinearity	–4.5	–	+4.5	LSB	
IDAC_INL	Integral nonlinearity	–5	–	+5	LSB	
IDAC_Gain (Source)	Range = 0.5x	6.64	–	22.46	μA	DAC setting = 128 dec. Not recommended for CapSense applications.
	Range = 1x	14.5	–	47.8	μA	
	Range = 2x	42.7	–	92.3	μA	
	Range = 4x	91.1	–	170	μA	DAC setting = 128 dec
	Range = 8x	184.5	–	426.9	μA	DAC setting = 128 dec

AC Chip-Level Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 16. AC Chip-Level Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
F _{IMO24}	IMO frequency at 24 MHz Setting	–	22.8	24	25.2	MHz
F _{IMO12}	IMO frequency at 12 MHz setting	–	11.4	12	12.6	MHz
F _{IMO6}	IMO frequency at 6 MHz setting	–	5.7	6.0	6.3	MHz
F _{CPU}	CPU frequency	–	0.75	–	25.20	MHz
F _{32K1}	ILO frequency	–	19	32	50	kHz
F _{32K_U}	ILO untrimmed frequency	–	13	32	82	kHz
DC _{IMO}	Duty cycle of IMO	–	40	50	60	%
DC _{ILO}	ILO duty cycle	–	40	50	60	%
SR _{POWER_UP}	Power supply slew rate	VIN slew rate during power-up	–	–	250	V/ms
t _{XRST}	External reset pulse width at power-up	After supply voltage is valid	1	–	–	ms
t _{XRST2}	External reset pulse width after power-up	Applies after part has booted	10	–	–	μs
t _{OS}	Startup time of ECO	–	–	1	–	s
t _{JIT_IMO}	N = 32	6 MHz IMO cycle-to-cycle jitter (RMS)	–	0.7	6.7	ns
		6 MHz IMO long term N (N = 32) cycle-to-cycle jitter (RMS)	–	4.3	29.3	ns
		6 MHz IMO period jitter (RMS)	–	0.7	3.3	ns
		12 MHz IMO cycle-to-cycle jitter (RMS)	–	0.5	5.2	ns
		12 MHz IMO long term N (N = 32) cycle-to-cycle jitter (RMS)	–	2.3	5.6	ns
		12 MHz IMO period jitter (RMS)	–	0.4	2.6	ns
		24 MHz IMO cycle-to-cycle jitter (RMS)	–	1.0	8.7	ns
		24 MHz IMO long term N (N = 32) cycle-to-cycle jitter (RMS)	–	1.4	6.0	ns
		24 MHz IMO period jitter (RMS)	–	0.6	4.0	ns

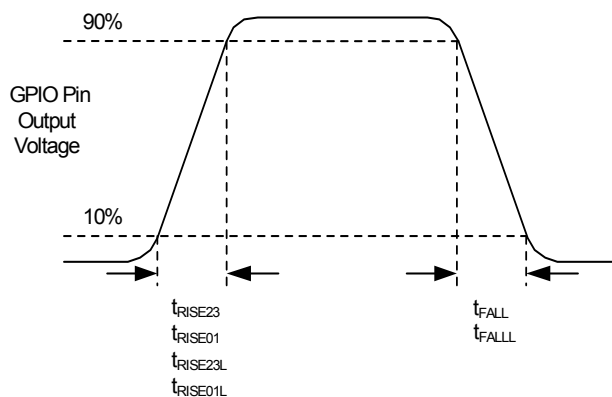
AC GPIO Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 17. AC GPIO Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
F_{GPIO}	GPIO operating frequency	Normal strong mode Port 0, 1	0	—	6 MHz for 1.9 V < VIN < 2.40 V 12 MHz for 2.40 V < VIN < 3.6 V	MHz MHz
t_{RISE23}	Rise time, strong mode, Load = 50 pF Port 2 or 3 or 4 pins	VIN = 3.0 to 3.6 V, 10% to 90%	15	—	80	ns
$t_{RISE23L}$	Rise time, strong mode low supply, Load = 50 pF, Port 2 or 3 or 4 pins	VIN = 1.9 to 3.0 V, 10% to 90%	15	—	80	ns
t_{RISE01}	Rise time, strong mode, Load = 50 pF, Ports 0 or 1	VIN = 3.0 to 3.6 V, 10% to 90%, LDO enabled or disabled	10	—	50	ns
$t_{RISE01L}$	Rise time, strong mode low supply, Load = 50 pF, Ports 0 or 1	VIN = 1.9 to 3.0 V, 10% to 90%, LDO enabled or disabled	10	—	80	ns
t_{FALL}	Fall time, strong mode, Load = 50 pF, all ports	VIN = 3.0 to 3.6 V, 10% to 90%	10	—	50	ns
t_{FALLL}	Fall time, strong mode low supply, Load = 50 pF, all ports	VIN = 1.9 to 3.0 V, 10% to 90%	10	—	70	ns

Figure 5. GPIO Timing Diagram



AC Comparator Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 18. AC Low Power Comparator Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
t_{LPC}	Comparator response time, 50 mV overdrive	50 mV overdrive does not include offset voltage.	–	–	100	ns

AC External Clock Specifications

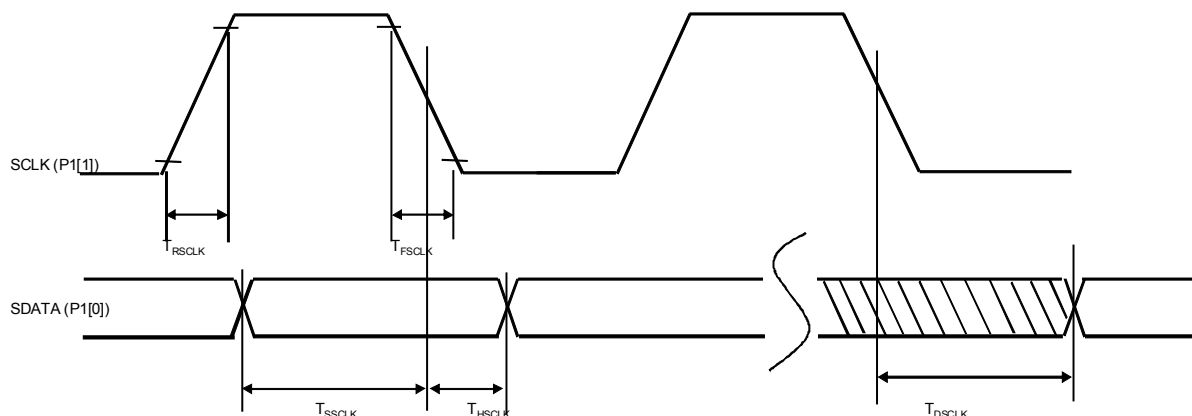
The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 19. AC External Clock Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
F_{OSCEXT}	Frequency (external oscillator frequency)	–	0.75	–	25.20	MHz
	High period	–	20.60	–	5300	ns
	Low period	–	20.60	–	–	ns
	Power-up IMO to switch	–	150	–	–	μ s

AC Programming Specifications

Figure 6. AC Waveform



The following table lists the guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 20. AC Programming Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
t_{RSCLK}	Rise time of SCLK	—	1	—	20	ns
t_{FSCLK}	Fall time of SCLK	—	1	—	20	ns
t_{SSCLK}	Data setup time to falling edge of SCLK	—	40	—	—	ns
t_{HSCLK}	Data hold time from falling edge of SCLK	—	40	—	—	ns
F_{SCLK}	Frequency of SCLK	—	0	—	8	MHz
t_{ERASEB}	Flash erase time (block)	—	—	—	18	ms
t_{WRITE}	Flash block write time	—	—	—	25	ms
t_{DSCLK3}	Data out delay from falling edge of SCLK	$3.0 \leq V_{DD} \leq 3.6$	—	—	85	ns
t_{DSCLK2}	Data out delay from falling edge of SCLK	$1.9 \leq V_{DD} \leq 3.0$	—	—	130	ns
t_{XRST3}	External reset pulse width after power-up	Required to enter programming mode when coming out of sleep	300	—	—	μ s
t_{XRES}	XRES pulse length	—	300	—	—	μ s
$t_{VDDWAIT}$	V_{DD} stable to wait-and-poll hold off	—	0.1	—	1	ms
$t_{VDDXRES}$	V_{DD} stable to XRES assertion delay	—	14.27	—	—	ms
t_{POLL}	SDATA high pulse time	—	0.01	—	200	ms
t_{ACQ}	“Key window” time after a V_{DD} ramp acquire event, based on 256 ILO clocks.	—	3.20	—	19.60	ms
$t_{XRESINI}$	“Key window” time after an XRES event, based on 8 ILO clocks	—	98	—	615	μ s

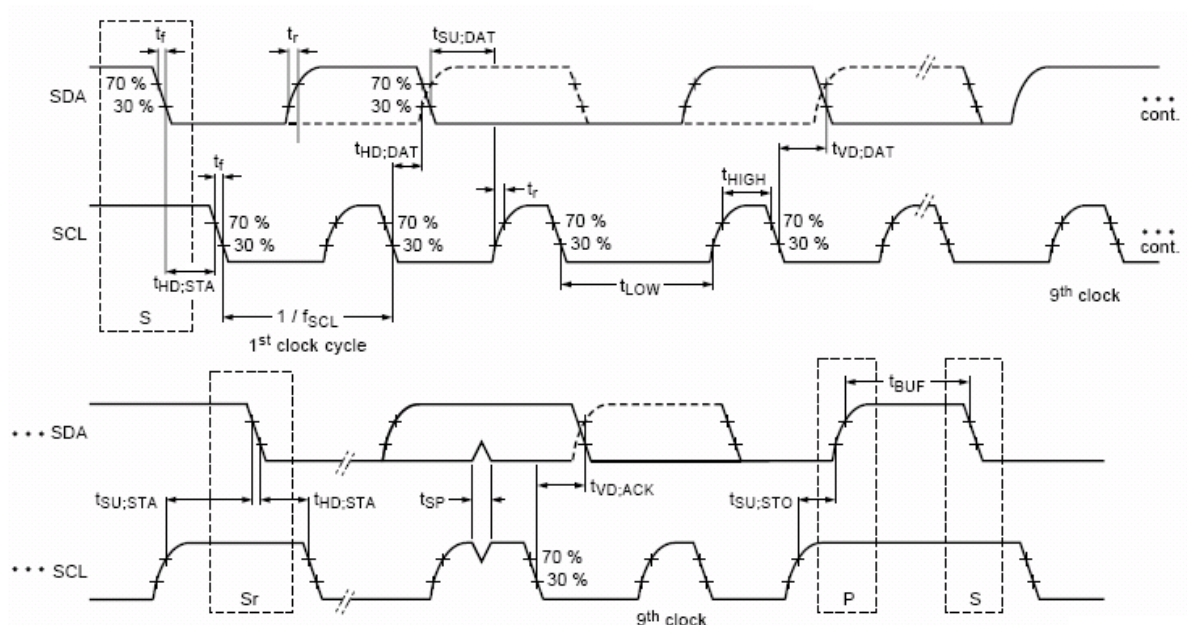
AC I²C Specifications

The following table lists guaranteed maximum and minimum specifications for the entire voltage and temperature ranges.

Table 21. AC Characteristics of the I²C SDA and SCL Pins

Symbol	Description	Standard Mode		Fast Mode		Units
		Min	Max	Min	Max	
f_{SCL}	SCL clock frequency	0	100	0	400	kHz
$t_{HD;STA}$	Hold time (repeated) START condition. After this period, the first clock pulse is generated	4.0	—	0.6	—	μ s
t_{LOW}	LOW period of the SCL clock	4.7	—	1.3	—	μ s
t_{HIGH}	HIGH Period of the SCL clock	4.0	—	0.6	—	μ s
$t_{SU;STA}$	Setup time for a repeated START condition	4.7	—	0.6	—	μ s
$t_{HD;DAT}$	Data hold time	0	3.45	0	0.90	μ s
$t_{SU;DAT}$	Data setup time	250	—	100 ^[12]	—	ns
$t_{SU;STO}$	Setup time for STOP condition	4.0	—	0.6	—	μ s
t_{BUF}	Bus free time between a STOP and START condition	4.7	—	1.3	—	μ s
t_{SP}	Pulse width of spikes are suppressed by the input filter	—	—	0	50	ns

Figure 7. Definition for Timing for Fast/Standard Mode on the I²C Bus



Note

12. A Fast-Mode I²C-bus device can be used in a standard mode I²C-bus system, but the requirement $t_{SU;DAT} \geq 250$ ns must then be met. This automatically be the case if the device does not stretch the LOW period of the SCL signal. If such device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line $t_{max} + t_{SU;DAT} = 1000 + 250 = 1250$ ns (according to the Standard-Mode I²C-bus specification) before the SCL line is released.

SPI Master AC Specifications

Table 22. SPI Master AC Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
F_{SCLK}	SCLK clock frequency	$V_{IN} \geq 2.4\text{ V}$ $V_{IN} < 2.4\text{ V}$	— —	— —	6 3	MHz MHz
DC	SCLK duty cycle	—	—	50	—	%
t_{SETUP}	MISO to SCLK setup time	$V_{IN} \geq 2.4\text{ V}$ $V_{IN} < 2.4\text{ V}$	60 100	— —	— —	ns ns
t_{HOLD}	SCLK to MISO hold time	—	40	—	—	ns
t_{OUT_VAL}	SCLK to MOSI valid time	—	—	—	40	ns
t_{OUT_HIGH}	MOSI high time	—	40	—	—	ns

Figure 8. SPI Master Mode 0 and 2

SPI Master, modes 0 and 2

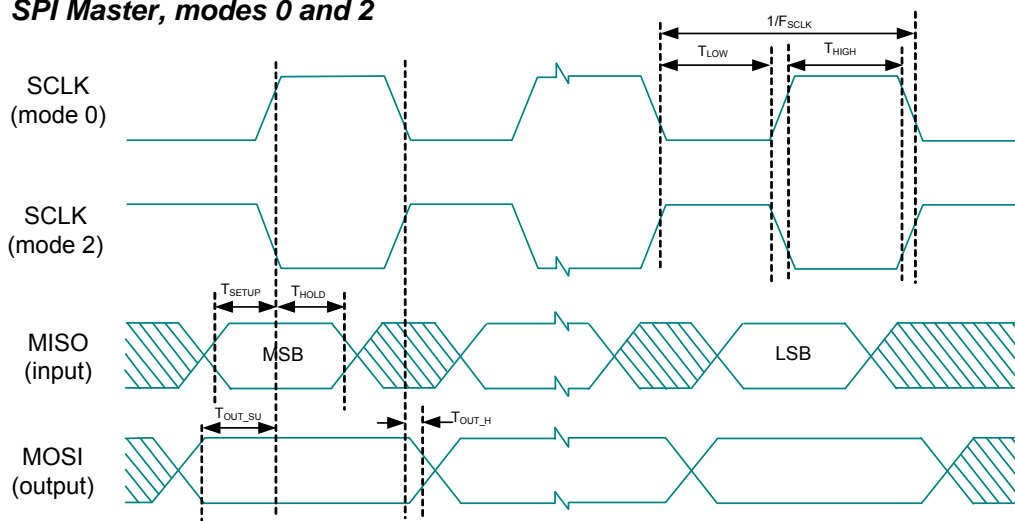
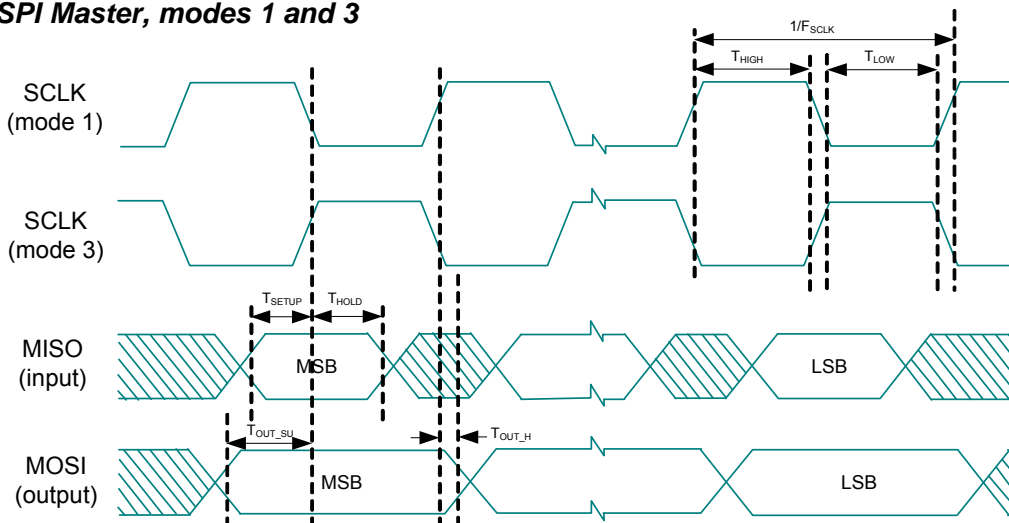


Figure 9. SPI Master Mode 1 and 3

SPI Master, modes 1 and 3



SPI Slave AC Specifications

Table 23. SPI Slave AC Specifications

Symbol	Description	Conditions	Min	Typ	Max	Units
F_{SCLK}	SCLK clock frequency	—	—	—	4	MHz
t_{LOW}	SCLK low time	—	42	—	—	ns
t_{HIGH}	SCLK high time	—	42	—	—	ns
t_{SETUP}	MOSI to SCLK setup time	—	30	—	—	ns
t_{HOLD}	SCLK to MOSI hold time	—	50	—	—	ns
t_{SS_MISO}	SS high to MISO valid	—	—	—	153	ns
t_{SCLK_MISO}	SCLK to MISO valid	—	—	—	125	ns
t_{SS_HIGH}	SS high time	—	50	—	—	ns
t_{SS_CLK}	Time from SS low to first SCLK	—	$2/SCLK$	—	—	ns
t_{CLK_SS}	Time from last SCLK to SS high	—	$2/SCLK$	—	—	ns

Figure 10. SPI Slave Mode 0 and 2

SPI Slave, modes 0 and 2

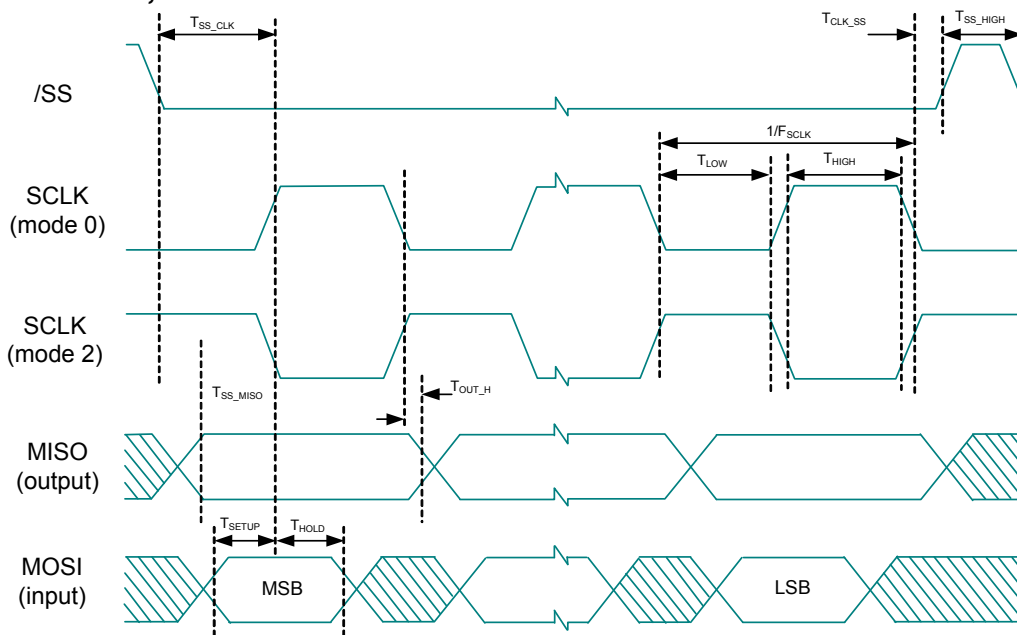
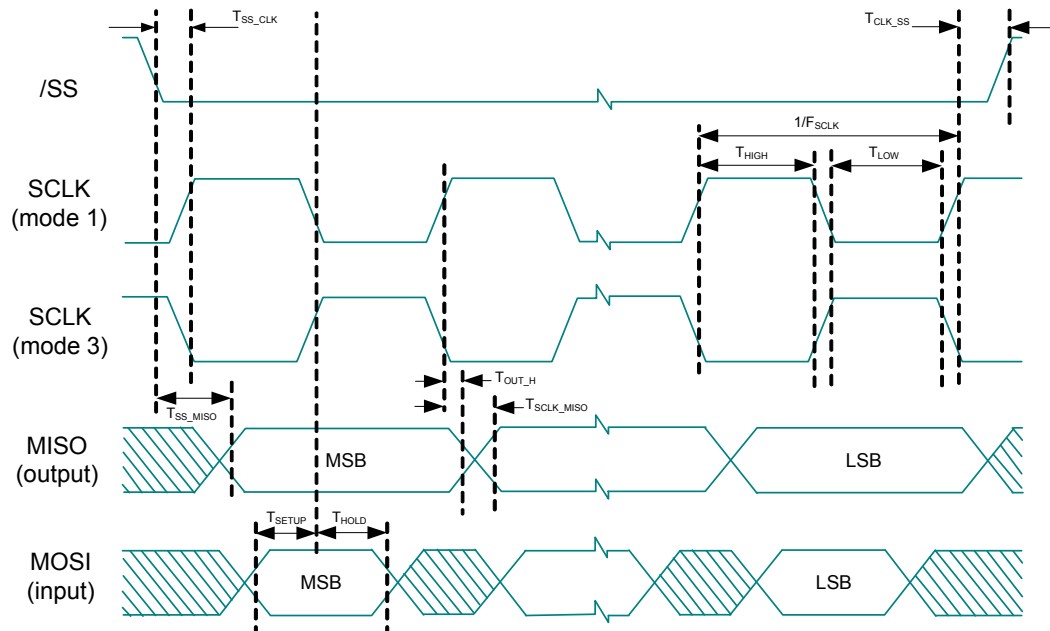


Figure 11. SPI Slave Mode 1 and 3

SPI Slave, modes 1 and 3



Electrical Specifications – RF Section

Symbol	Description	Min	Typ	Max	Units	Test Condition and Notes
	Supply voltage					
V _{IN}	DC power supply voltage range	1.9	–	3.6	VDC	Input to V _{IN} pins
	Current consumption					
I _{DD_TX2}	Current consumption – Tx	–	18.5	–	mA	Transmit power PA2.
I _{DD_TX12}		–	13.7	–	mA	Transmit power PA12.
I _{DD_RX}	Current consumption – Rx	–	18	–	mA	
I _{DD_IDLE1}	Current consumption – idle	–	1.1	–	mA	
I _{DD_SLPx}	Current consumption – sleep	–	1	–	μA	Temperature = +25 °C. Using firmware sleep patch. Register 27 = 0x1200, for V _{IN} ≥ 3.00 VDC only
I _{DD_SLP_r}		–	8	–	μA	Temperature = +25 °C; using firmware sleep patch Register 27 = 0x4200.
I _{DD_SLP_h}		–	38	–	μA	Temperature = +70 °C 'C' grade part; using firmware sleep patch Register 27 = 0x4200
V _{IH}	Logic input high	0.8 V _{IN}	–	1.2 V _{IN}	V	
V _{IL}	Logic input low	0	–	0.8	V	
I _{_LEAK_IN}	Input leakage current	–	–	10	μA	
V _{OH}	Logic output high	0.8 V _{IN}	–	–	V	I _{OH} = 100 μA source
V _{OL}	Logic output low	–	–	0.4	V	I _{OL} = 100 μA sink
I _{_LEAK_OUT}	Output leakage current	–	–	10	μA	MISO in tristate
T _{_RISE_OUT}	Rise/fall time (SPI MISO)	–	8	25	ns	7 pF cap. load
T _{_RISE_IN}	Rise/fall time (SPI MOSI)	–	–	25	ns	
T _{r_spi}	CLK rise, fall time (SPI)	–	–	25	ns	Requirement for error-free register reading, writing.
F _{OP}	Operating frequency range	2400	–	2482	MHz	Usage on-the-air is subject to local regulatory agency restrictions regarding operating frequency.
V _{SWR_I}	Antenna port mismatch (Z ₀ = 50 Ω)	–	<2:1	–	VSWR	Receive mode. Measured using LC matching circuit
VSWR _{_O}		–	<2:1	–	VSWR	Transmit mode. Measured using LC matching circuit
Receive section						Measured using LC matching circuit for BER ≤ 0.1%

Electrical Specifications – RF Section *(continued)*

Symbol	Description	Min	Typ	Max	Units	Test Condition and Notes
RxS _{base}	Receiver sensitivity (FEC off)	–	–87	–	dBm	Room temperature only 0-ppm crystal frequency error.
RxS _{temp}		–	–84	–	dBm	Over temperature; 0-ppm crystal frequency error.
RxS _{ppm}		–	–84	–	dBm	Room temperature only 80-ppm total frequency error (± 40-ppm crystal frequency error, each end of RF link)
RxS _{temp+ppm}		–	–80	–	dBm	Over temperature; 80-ppm total frequency error (± 40-ppm crystal frequency error, each end of RF link)
R _{xmax-sig}	Maximum usable signal	–20	0	–	dBm	Room temperature only
T _s	Data (Symbol) rate	–	1	–	μs	
Minimum Carrier/Interference ratio						For BER ≤ 0.1%. Room temperature only.
CI _{_cochannel}	Co-channel interference	–	+9	–	dB	–60-dBm desired signal
CI _{_1}	Adjacent channel interference, 1-MHz offset	–	+6	–	dB	–60-dBm desired signal
CI _{_2}	Adjacent channel interference, 2-MHz offset	–	–12	–	dB	–60-dBm desired signal
CI _{_3}	Adjacent channel interference, 3-MHz offset	–	–24	–	dB	–67-dBm desired signal
OBB	Out-of-band blocking	–	≥ –27	–	dBm	30 MHz to 12.75 GHz Measured with ACX BF2520 ceramic filter on ant. pin. –67-dBm desired signal, BER ≤ 0.1%. Room temperature only.
Transmit section						Measured using a LC matching circuit
P _{AVH}	RF output power	–	+1	–	dBm	PA0 (PA_GN = 0, Reg9 = 0x1820). Room temperature only
P _{AVL}		–	–11.2	–	dBm	PA12 (PA_GN = 12, Reg9 = 0x1E20). Room temperature only.
TxP _{fx2}	Second harmonic	–	–45	–	dBm	Measured using a LC matching circuit. Room temperature only.
TxP _{fx3}	Third and higher harmonics	–	≤ –45	–	dBm	Measured using a LC matching circuit. Room temperature only.
Modulation characteristics						
Df1 _{avg}		–	263	–	kHz	Modulation pattern: 11110000...
Df2 _{avg}		–	255	–	kHz	Modulation pattern: 10101010...
In-band spurious emission						
IBS _{_2}	2-MHz offset	–	–	–20	dBm	
IBS _{_3}	3-MHz offset	–	–	–30	dBm	
IBS _{_4}	≥ 4-MHz offset	–	≤ –30	–	dBm	

Electrical Specifications – RF Section *(continued)*

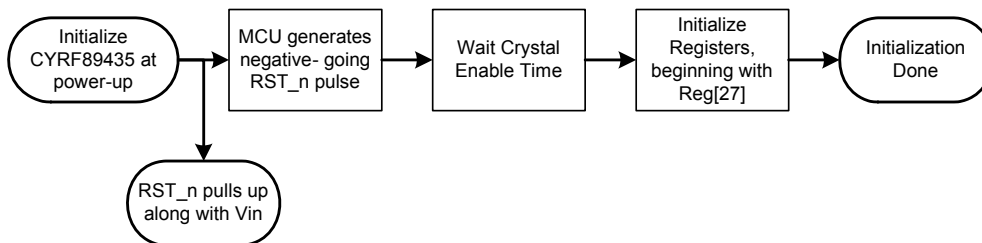
Symbol	Description	Min	Typ	Max	Units	Test Condition and Notes
RF VCO and PLL section						
F _{step}	Channel (Step) size		1	–	MHz	
L _{100k}	SSB phase noise		–75	–	dBc/Hz	100-kHz offset
L _{1M}			–105	–	dBc/Hz	1-MHz offset
dF _{X0}	Crystal oscillator frequency error	–40	–	+40	ppm	Relative to 12-MHz crystal reference frequency
T _{HOP}	RF PLL settling time	–	100	150	μs	Settle to within 30 kHz of final value. AutoCAL off.
T _{HOP_AC}		–	250	350	μs	Settle to within 30 kHz of final value. AutoCAL on.
LDO voltage regulator section						
V _{DO}	Dropout voltage	–	0.17	0.3	V	Measured during receive state

Initialization Timing Requirements

Table 24. Initialization Timing Requirements

Timing Parameter	Min	Max	Unit	Notes
T_{RSU}	–	30 / 150	ms	30 ms Reset setup time necessary to ensure complete Reset for $V_{IN} = 6.5\text{mV/s}$, 150 ms Reset setup time necessary to ensure complete Reset for $V_{IN} = 2\text{mV/s}$
T_{RPW}	1	10	μs	Reset pulse width necessary to ensure complete reset
T_{CMIN}	3	–	ms	Minimum recommended crystal oscillator and APLL settling time
T_{VIN}	–	6.5 / 2	mV/s	Maximum ramp time for V_{IN} , measured from 0 to 100% of final voltage. For example, if $V_{IN} = 3.3\text{ V}$, the max ramp time is $6.5 \times 3.3 = 21.45\text{ ms}$. If $V_{IN} = 1.9\text{ V}$, the max ramp time = $6.5 \times 1.9 = 12.35\text{ ms}$. Reset setup time necessary to ensure complete Reset for $V_{IN} = 6.5\text{ mV/s}$ Reset setup time necessary to ensure complete Reset for $V_{IN} = 6.5\text{ mV/s}$ Reset setup time necessary to ensure complete Reset for $V_{IN}=6.5\text{ mV/s}$

Figure 12. Initialization Flowchart

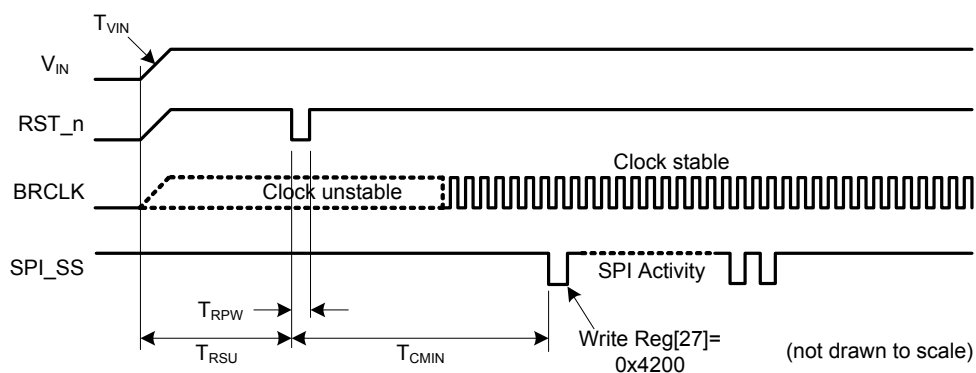


SPI Timing Requirements

Table 25. SPI Timing Requirements

Timing Parameter	Min	Max	Unit	Notes
T_{SSS}	20	–	ns	Setup time from assertion of SPI_SS to CLK edge
T_{SSH}	200	–	ns	Hold time required deassertion of SPI_SS
T_{SCKH}	40	–	ns	CLK minimum high time
T_{SCKL}	40	–	ns	CLK minimum low time
T_{SCK}	83	–	ns	Maximum CLK clock is 12 MHz
T_{SSU}	30	–	ns	MOSI setup time
T_{SHD}	10	–	ns	MOSI hold time
T_{SS_SU}	10	–	ns	Before SPI_SS enable, CLK hold low time requirement
T_{SS_HD}	200	–	ns	Minimum SPI inactive time
T_{SDO}	–	35	ns	MISO setup time, ready to read
T_{SDO1}	–	5	ns	If MISO is configured as tristate, MISO assertion time
T_{SDO2}	–	250	ns	If MISO is configured as tristate, MISO deassertion time
$T1 \text{ Min_R50}$	350	–	ns	When reading register 50 (FIFO)
$T1 \text{ Min}$	83	–	ns	When writing Register 50 (FIFO), or reading/writing any registers other than register 50.

Figure 13. Power-on and Register Programming Sequence



- After RST_n transitions from 0 to 1, BRCLK begins running at 12-MHz clock.
- After register initialization, CYRF89435 is ready to transmit or receive.

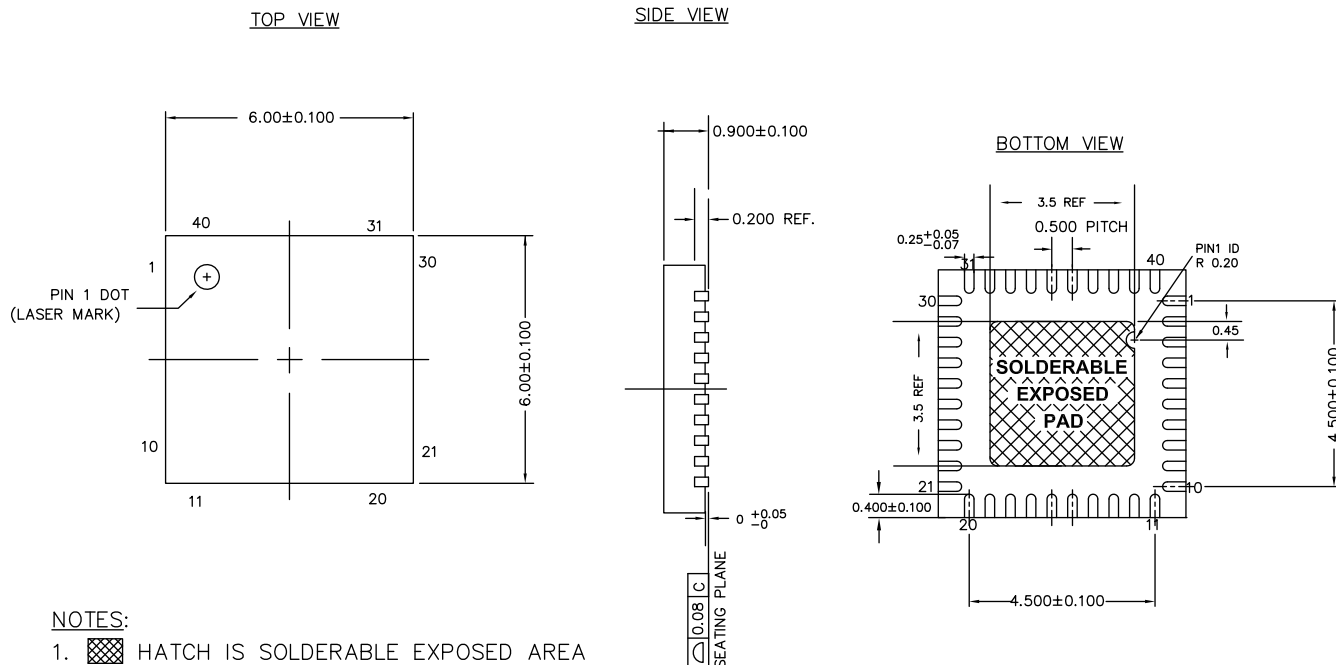
Packaging Information

This section illustrates the packaging specifications for the CY7C89435 PSoC device, along with the thermal impedances for each package.

Important Note

Emulation tools may require a larger area on the target PCB than the chip's footprint. For a detailed description of the emulation tools' dimensions, refer to the document titled *PSoC Emulator Pod Dimensions* at <http://www.cypress.com/design/MR10161>.

Figure 14. 40-pin QFN (6 × 6 × 1.0 mm) LT40B 3.5 × 3.5 mm E-Pad (Sawn) Package Outline, 001-13190



001-13190 *H

Important Notes

- For information on the preferred dimensions for mounting QFN packages, see the following Application Note at http://www.amkor.com/products/notes_papers/MLFAppNote.pdf.
- Pinned vias for thermal conduction are not required for the low power PSoC device.

Thermal Impedances

Table 26. Thermal Impedances per Package

Package	Typical θ_{JA} ^[13]	Typical θ_{JC}
40-pin QFN ^[14]	27°C/W	34°C/W

Capacitance on Crystal Pins

Table 27. Typical Package Capacitance on Crystal Pins

Package	Package Capacitance
40-pin QFN	36 pF

Solder Reflow Specifications

Table 28 shows the solder reflow temperature limits that must not be exceeded.

Table 28. Solder Reflow Specifications

Package	Minimum Peak Temperature	Maximum Peak Temperature
40-pin QFN	260 °C	265 °C

Notes

13. $T_J = T_A + \text{Power} \times \theta_{JA}$.

14. To achieve the thermal impedance specified for the QFN package, the center thermal pad must be soldered to the PCB ground plane.

Development Tool Selection

Software

PSoC Designer™

At the core of the PSoC development software suite is PSoC Designer. Utilized by thousands of PSoC developers, this robust software has been facilitating PSoC designs for over half a decade. PSoC Designer is available free of charge at <http://www.cypress.com>.

PSoC Programmer

Flexible enough to be used on the bench in development, yet suitable for factory programming, PSoC Programmer works either as a standalone programming application or it can operate directly from PSoC Designer. PSoC Programmer software is compatible with both PSoC ICE-Cube In-Circuit Emulator and PSoC MiniProg. PSoC Programmer is available free of charge at <http://www.cypress.com>.

Development Kits

All development kits are sold at the Cypress Online Store.

CY3215-DK Basic Development Kit

The CY3215-DK is for prototyping and development with PSoC Designer. This kit supports in-circuit emulation and the software interface enables users to run, halt, and single step the processor and view the content of specific memory locations. PSoC Designer supports the advance emulation features also. The kit includes:

- PSoC Designer Software CD
- ICE-Cube In-Circuit Emulator
- ICE Flex-Pod for CY8C29X66A Family
- Cat-5 Adapter
- Mini-Eval Programming Board
- 110 ~ 240 V Power Supply, Euro-Plug Adapter
- iMAGEcraft C Compiler (Registration Required)
- ISSP Cable
- 2 CY8C29466A-24PXI 28-pin PDIP Chip Samples

Device Programmers

Firmware needs to be downloaded to PSoC CS device only at 3.3 V using MiniProg3 Programmer. This Programmer kit can be purchased from Cypress Store using part# 'CY8CKIT-002 - MiniProg3'. It is a small, compact programmer which connects PC via a USB 2.0 cable (provided along with CY8cKIT-002).

Note: MiniProg1 Programmer should not be used as it does not support programming at 3.3 V.

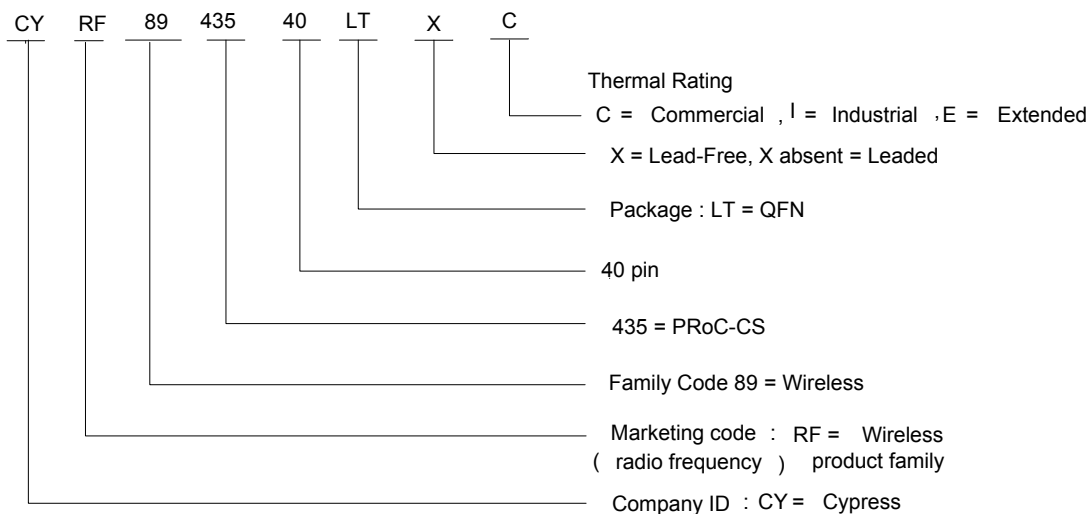
Ordering Information

The following table lists the CY7C89435 PSoC devices' key package features and ordering codes.

Table 29. PSoC Device Key Features and Ordering Information

Package	Ordering Code	Flash (Bytes)	SRAM (Bytes)	CapSense Blocks	Digital I/O Pins	Analog Inputs	XRES Pin	ADC
40-pin (6 × 6 × 1.0 mm) QFN	CYRF89435-40LTXC	32 K	2 K	1	13	13	Yes	Yes

Ordering Code Definitions



Acronyms

Table 30. Acronyms Used in this Document

Acronym	Description
AC	Alternating Current
ADC	Analog-to-Digital Converter
API	Application Programming Interface
CMOS	Complementary Metal Oxide Semiconductor
CPU	Central Processing Unit
DAC	Digital-to-Analog Converter
DC	Direct Current
EOP	End Of Packet
FSR	Full Scale Range
GPIO	General Purpose Input/Output
GUI	Graphical User Interface
I ² C	Inter-Integrated Circuit
ICE	In-Circuit Emulator
IDAC	Digital Analog Converter Current
ILO	Internal Low Speed Oscillator
IMO	Internal Main Oscillator
I/O	Input/Output
ISSP	In-System Serial Programming
LCD	Liquid Crystal Display
LDO	Low Dropout (regulator)
LSB	Least-Significant Bit
LVD	Low Voltage Detect
MCU	Micro-Controller Unit
MIPS	Mega Instructions Per Second
MISO	Master In Slave Out
MOSI	Master Out Slave In
MSB	Most-Significant Bit
OCD	On-Chip Debugger
POR	Power On Reset
PPOR	Precision Power On Reset
PSRR	Power Supply Rejection Ratio
PWRSYS	Power System
PSoC [®]	Programmable System-on-Chip
SLIMO	slow internal main oscillator
SRAM	Static Random Access Memory
SNR	Signal to Noise Ratio
QFN	Quad Flat No-lead
SCL	Serial I2C Clock
SDA	Serial I2C Data
SDATA	Serial ISSP Data
SPI	Serial Peripheral Interface
SS	Slave Select
SSOP	Shrink Small Outline Package
TC	Test Controller
USB	Universal Serial Bus
USB D+	USB Data+
USB D-	USB Data-
WLCSP	Wafer Level Chip Scale Package
XTAL	Crystal

Reference Documents

- *Technical reference manual for CY8C20xx6 devices*
- *In-system Serial Programming (ISSP) protocol for 20xx6 (AN2026C)*
- *Host Sourced Serial Programming for 20xx6 devices (AN59389)*

Document Conventions

Units of Measure

Table 31. Units of Measure

Symbol	Unit of Measure
°C	degree Celsius
dB	decibels
fF	femtofarad
g	gram
Hz	hertz
KB	1024 bytes
Kbit	1024 bits
KHz	kilohertz
Ksps	kilo samples per second
kΩ	kilohm
MHz	megahertz
MΩ	megaohm
μA	microampere
μF	microfarad
μH	microhenry
μs	microsecond
μW	microwatt
mA	milliampere
ms	millisecond
mV	millivolt
nA	nanoampere
nF	nanofarad
ns	nanosecond
nV	nanovolt
W	ohm
pA	picoampere
pF	picofarad
pp	peak-to-peak
ppm	parts per million
ps	picosecond
sps	samples per second
s	sigma: one standard deviation
V	volt
W	watt

Numeric Naming

Hexadecimal numbers are represented with all letters in uppercase with an appended lowercase 'h' (for example, '14h' or '3Ah'). Hexadecimal numbers may also be represented by a '0x' prefix, the C coding convention. Binary numbers have an appended lowercase 'b' (for example, '01010100b' or '01000011b'). Numbers not indicated by an 'h', 'b', or 0x are decimal.

Glossary

Crosspoint connection	Connection between any GPIO combination via analog multiplexer bus.
Differential non-linearity	Ideally, any two adjacent digital codes correspond to output analog voltages that are exactly one LSB apart. Differential non-linearity is a measure of the worst case deviation from the ideal 1 LSB step.
Hold time	Hold time is the time following a clock event during which the data input to a latch or flip-flop must remain stable in order to guarantee that the latched data is correct.
I ² C	It is a serial multi-master bus used to connect low speed peripherals to MCU.
Integral nonlinearity	It is a term describing the maximum deviation between the ideal output of a DAC/ADC and the actual output level.
Latch-up current	Current at which the latch-up test is conducted according to JESD78 standard (at 125 degree Celsius)
Power supply rejection ratio (PSRR)	The PSRR is defined as the ratio of the change in supply voltage to the corresponding change in output voltage of the device.
Scan	The conversion of all sensor capacitances to digital values.
Setup time	Period required to prepare a device, machine, process, or system for it to be ready to function.
Signal-to-noise ratio	The ratio between a capacitive finger signal and system noise.
SPI	Serial peripheral interface is a synchronous serial data link standard.

Document History Page

Document Title: CYRF89435, P _{RoC} ™ - CapSense® Document Number: 001-76581				
Revision	ECN	Orig. of Change	Submission Date	Description of Change
**	3545779	ANTG	03/13/2012	New silicon document
*A	3591949	ANTG	05/14/2012	Modified title. Updated status "Company Confidential" of the datasheet. Changed "P _{RoC} NL - CapSense" to "P _{RoC} -CS" everywhere in the datasheet. Updated the Electrical Specifications. Updated the RF specifications.
*B	3714928	AKHL	08/16/2012	Major text update. Updated the pinout (Figure 3).
*C	3747532	AKHL	09/25/2012	Removed "Company Confidential" tag in the header. Replaced package diagram spec with 001-13190.
*D	3784571	AKHL	10/18/2012	Updated PSoC® Functional Overview (Added Transmit Power Control). Updated Electrical Specifications – RF Section (Replaced CYRF8935 with CYRF89435 in Figure 12 and also in the last bullet point below Figure 13). Updated Development Tool Selection (Updated Evaluation Tools (Removed "CY8CKIT-002 - MiniProg 3"), updated Device Programmers (Removed "CY3207ISSP In-System Serial Programmer (ISSP)", added the content from the removed section "CY8CKIT-002 - MiniProg 3" with slight modification). Updated in new template.
*E	3982770	AKHL	05/15/2013	Updated P_{RoC}-CS Features . Updated Logical Block Diagram . Updated PSoC® Functional Overview : Updated WirelessUSB-NL System (Updated Figure 2). Updated Transmit Power Control (Updated Table 1). Removed "Development Kits". Removed "Training". Updated Electrical Specifications – PSoC Core : Updated Absolute Maximum Ratings (Updated Table 2). Updated Operating Temperature (Updated Table 3). Updated Electrical Specifications – RF Section : Updated SPI Timing Requirements (Updated Table 25). Updated Packaging Information : No change in Package Diagram revision. Updated Capacitance on Crystal Pins (Updated Table 27). Updated Solder Reflow Specifications (Updated Table 28). Updated Development Tool Selection : Removed "Evaluation Tools". Removed "Accessories (Emulation and Programming)". Removed "Third Party Tools". Updated Ordering Information : No Change in part numbers. Added Ordering Code Definitions .

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