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## FDP8870

### N-Channel PowerTrench® MOSFET 30V, 156A, 4.1mΩ

#### General Description

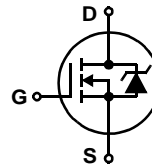
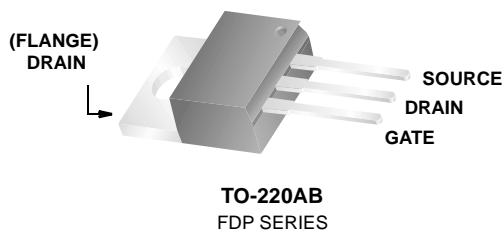
This N-Channel MOSFET has been designed specifically to improve the overall efficiency of DC/DC converters using either synchronous or conventional switching PWM controllers. It has been optimized for low gate charge, low  $r_{DS(ON)}$  and fast switching speed.

#### Features

- $r_{DS(ON)} = 4.1m\Omega$ ,  $V_{GS} = 10V$ ,  $I_D = 35A$
- $r_{DS(ON)} = 4.6m\Omega$ ,  $V_{GS} = 4.5V$ ,  $I_D = 35A$
- High performance trench technology for extremely low  $r_{DS(ON)}$
- Low gate charge
- High power and current handling capability
- RoHS Compliant

#### Applications

- DC/DC converters



#### MOSFET Maximum Ratings $T_C = 25^\circ C$ unless otherwise noted

Symbol	Parameter	Ratings	Units
$V_{DSS}$	Drain to Source Voltage	30	V
$V_{GS}$	Gate to Source Voltage	$\pm 20$	V
$I_D$	Drain Current		
	Continuous ( $T_C = 25^\circ C$ , $V_{GS} = 10V$ ) (Note 1)	156	A
	Continuous ( $T_C = 25^\circ C$ , $V_{GS} = 4.5V$ ) (Note 1)	147	A
	Continuous ( $T_{amb} = 25^\circ C$ , $V_{GS} = 10V$ , with $R_{\theta JA} = 62^\circ C/W$ )	19	A
	Pulsed	Figure 4	A
$E_{AS}$	Single Pulse Avalanche Energy (Note 2)	300	mJ
$P_D$	Power dissipation	160	W
	Derate above $25^\circ C$	1.07	$W/^\circ C$
$T_J, T_{STG}$	Operating and Storage Temperature	-55 to 175	$^\circ C$

#### Thermal Characteristics

$R_{\theta JC}$	Thermal Resistance Junction to Case TO-220	0.94	$^\circ C/W$
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-220 ( Note 3)	62	$^\circ C/W$

#### Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDP8870	FDP8870	TO-220AB	Tube	N/A	50 units

**Electrical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Typ	Max	Units
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**Off Characteristics**

$B_{VDSS}$	Drain to Source Breakdown Voltage	$I_D = 250\mu\text{A}, V_{GS} = 0\text{V}$	30	-	-	V
$I_{DSS}$	Zero Gate Voltage Drain Current	$V_{DS} = 24\text{V}$ $V_{GS} = 0\text{V}$	-	-	1	$\mu\text{A}$
		$T_C = 150^\circ\text{C}$	-	-	250	
$I_{GSS}$	Gate to Source Leakage Current	$V_{GS} = \pm 20\text{V}$	-	-	$\pm 100$	nA

**On Characteristics**

$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}, I_D = 250\mu\text{A}$	1.2	-	2.5	V
$r_{DS(ON)}$	Drain to Source On Resistance	$I_D = 35\text{A}, V_{GS} = 10\text{V}$	-	0.0034	0.0041	$\Omega$
		$I_D = 35\text{A}, V_{GS} = 4.5\text{V}$	-	0.0040	0.0046	
		$I_D = 35\text{A}, V_{GS} = 10\text{V}, T_J = 175^\circ\text{C}$	-	0.0051	0.0065	

**Dynamic Characteristics**

$C_{ISS}$	Input Capacitance	$V_{DS} = 15\text{V}, V_{GS} = 0\text{V}, f = 1\text{MHz}$	-	5200	-	pF	
$C_{OSS}$	Output Capacitance		-	970	-	pF	
$C_{RSS}$	Reverse Transfer Capacitance		-	570	-	pF	
$R_G$	Gate Resistance	$V_{GS} = 0.5\text{V}, f = 1\text{MHz}$	-	2.1	-	$\Omega$	
$Q_{g(TOT)}$	Total Gate Charge at 10V	$V_{GS} = 0\text{V to } 10\text{V}$	$V_{DD} = 15\text{V}$ $I_D = 35\text{A}$ $I_g = 1.0\text{mA}$	-	106	132	nC
$Q_{g(5)}$	Total Gate Charge at 5V	$V_{GS} = 0\text{V to } 5\text{V}$		-	56	69	nC
$Q_{g(TH)}$	Threshold Gate Charge	$V_{GS} = 0\text{V to } 1\text{V}$		-	5.0	6.5	nC
$Q_{gs}$	Gate to Source Gate Charge			-	15	-	nC
$Q_{gs2}$	Gate Charge Threshold to Plateau			-	10	-	nC
$Q_{gd}$	Gate to Drain "Miller" Charge			-	23	-	nC

**Switching Characteristics** ( $V_{GS} = 10\text{V}$ )

$t_{ON}$	Turn-On Time	$V_{DD} = 15\text{V}, I_D = 35\text{A}$ $V_{GS} = 4.5\text{V}, R_{GS} = 3.3\Omega$	-	-	168	ns
$t_{d(ON)}$	Turn-On Delay Time		-	11	-	ns
$t_r$	Rise Time		-	105	-	ns
$t_{d(OFF)}$	Turn-Off Delay Time		-	70	-	ns
$t_f$	Fall Time		-	46	-	ns
$t_{OFF}$	Turn-Off Time		-	-	173	ns

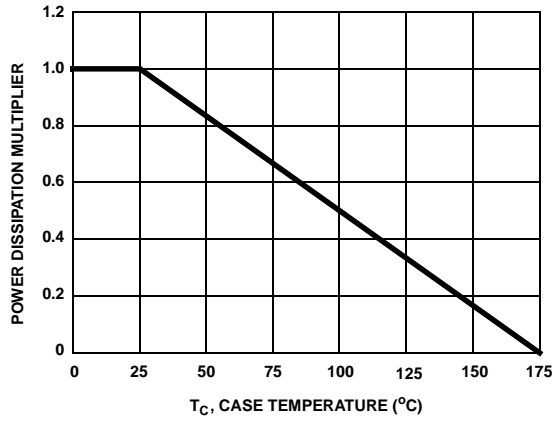
**Drain-Source Diode Characteristics**

$V_{SD}$	Source to Drain Diode Voltage	$I_{SD} = 35\text{A}$	-	-	1.25	V
		$I_{SD} = 15\text{A}$	-	-	1.0	V
$t_{rr}$	Reverse Recovery Time	$I_{SD} = 35\text{A}, di_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	37	ns
$Q_{RR}$	Reverse Recovered Charge	$I_{SD} = 35\text{A}, di_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	21	nC

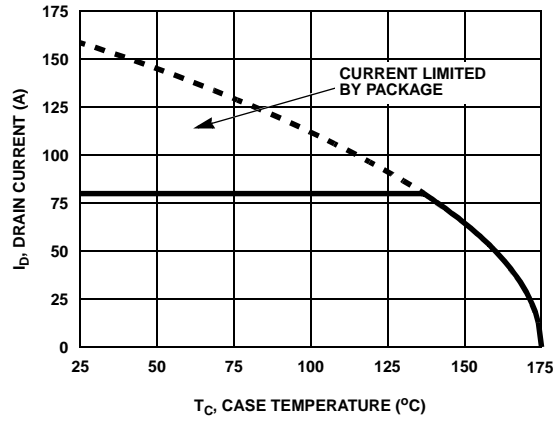
**Notes:**

- 1: Package current limitation is 80A.
- 2: Starting  $T_J = 25^\circ\text{C}$ ,  $L = 0.15\text{mH}$ ,  $I_{AS} = 64\text{A}$ ,  $V_{DD} = 27\text{V}$ ,  $V_{GS} = 10\text{V}$ .
- 3: Pulse width = 100s.

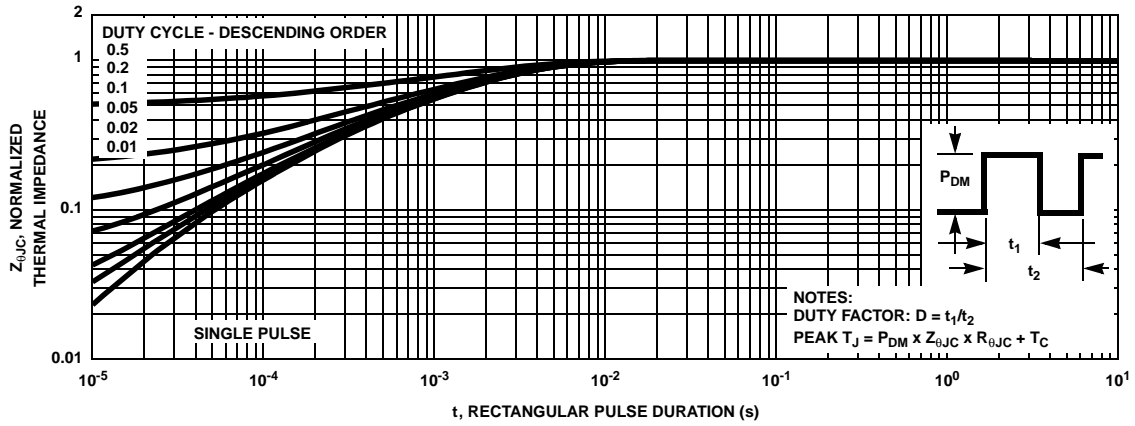
**Typical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted



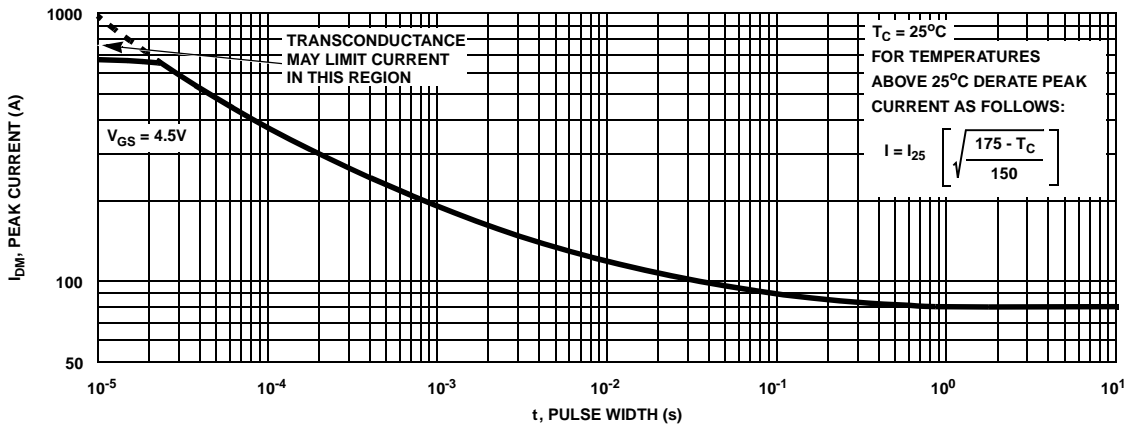
**Figure 1. Normalized Power Dissipation vs Case Temperature**



**Figure 2. Maximum Continuous Drain Current vs Case Temperature**

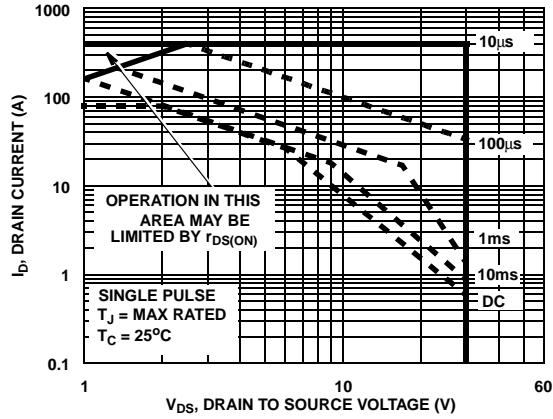


**Figure 3. Normalized Maximum Transient Thermal Impedance**

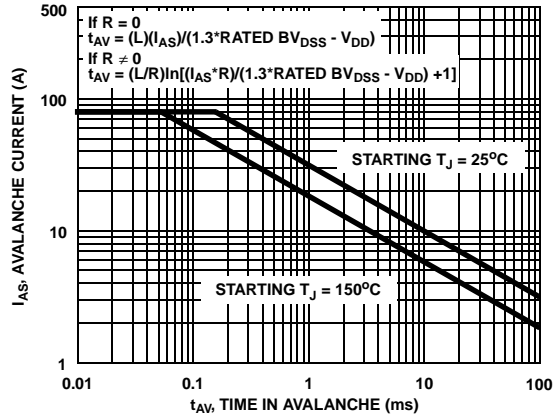


**Figure 4. Peak Current Capability**

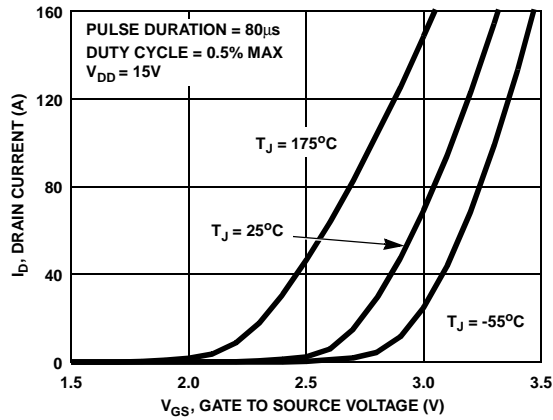
**Typical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted



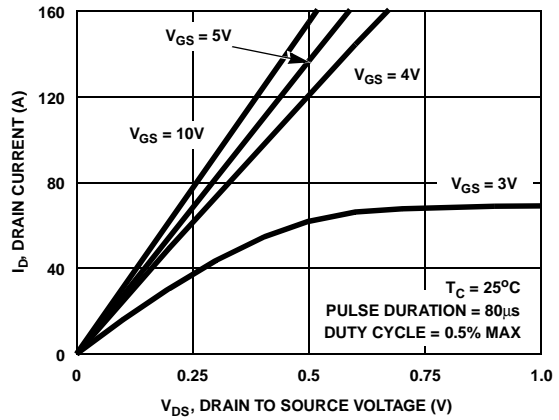
**Figure 5. Forward Bias Safe Operating Area**



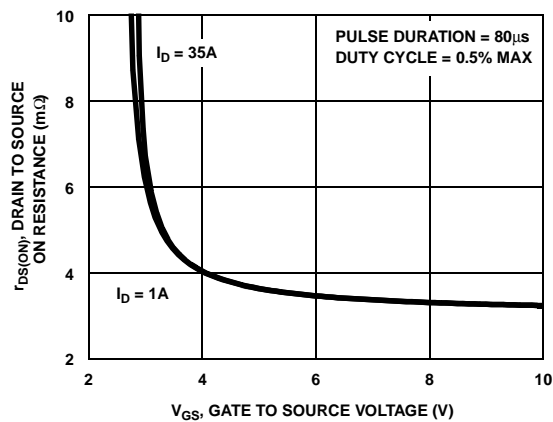
**Figure 6. Unclamped Inductive Switching Capability**  
 NOTE: Refer to Fairchild Application Notes AN7514 and AN7515



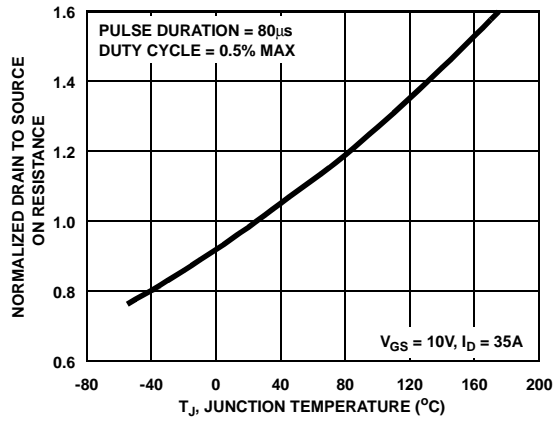
**Figure 7. Transfer Characteristics**



**Figure 8. Saturation Characteristics**

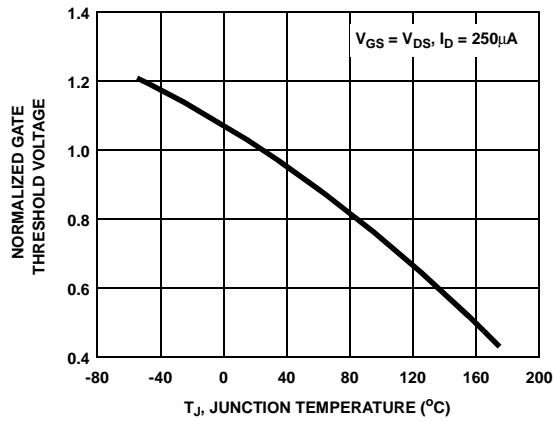


**Figure 9. Drain to Source On Resistance vs Gate Voltage and Drain Current**

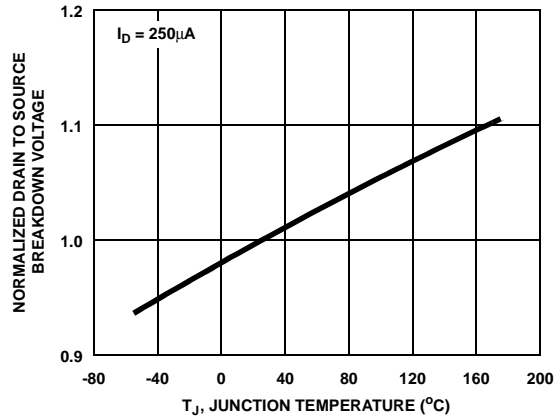


**Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature**

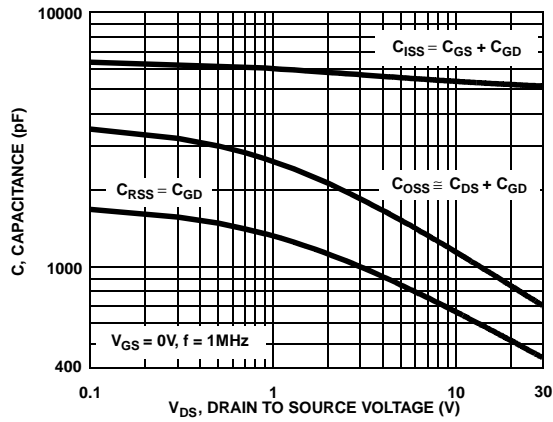
**Typical Characteristics**  $T_C = 25^\circ\text{C}$  unless otherwise noted



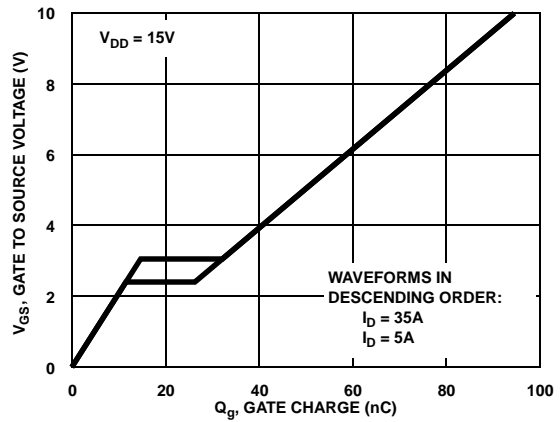
**Figure 11. Normalized Gate Threshold Voltage vs Junction Temperature**



**Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature**



**Figure 13. Capacitance vs Drain to Source Voltage**



**Figure 14. Gate Charge Waveforms for Constant Gate Current**

Test Circuits and Waveforms

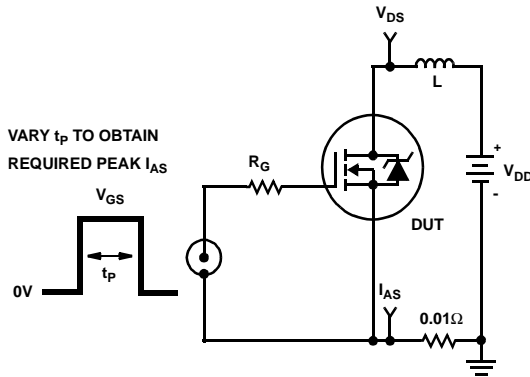


Figure 15. Unclamped Energy Test Circuit

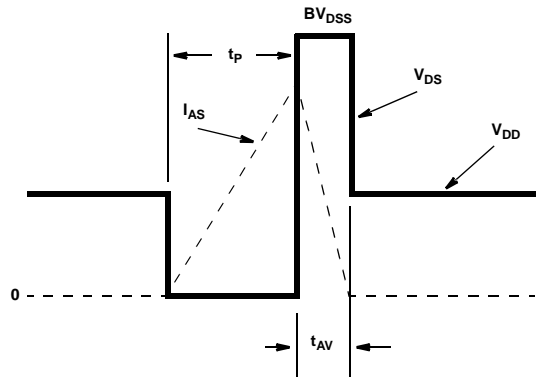


Figure 16. Unclamped Energy Waveforms

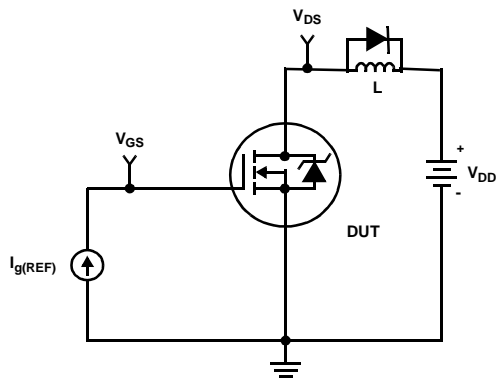


Figure 17. Gate Charge Test Circuit

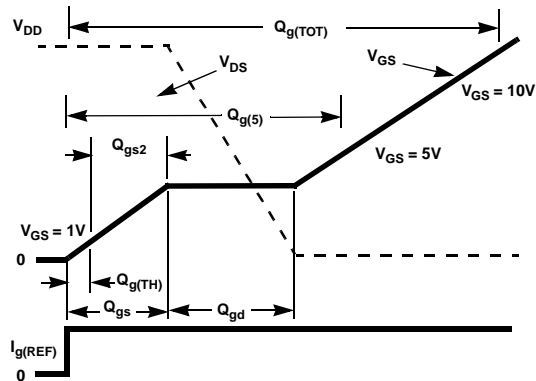


Figure 18. Gate Charge Waveforms

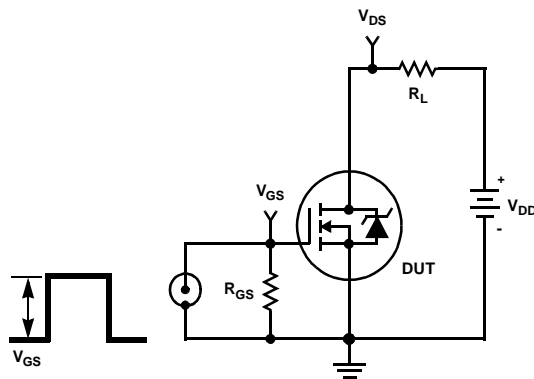


Figure 19. Switching Time Test Circuit

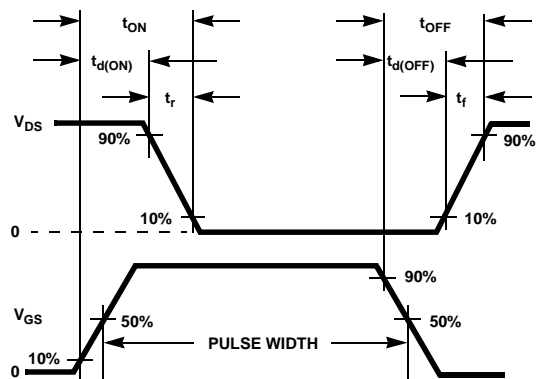


Figure 20. Switching Time Waveforms

## PSPICE Electrical Model

.SUBCKT FDP8870 2 1 3 ; rev December 2003

Ca 12 8 4.5e-9

Cb 15 14 4.5e-9

Cin 6 8 4.7e-9

Dbody 7 5 DbodyMOD  
 Dbreak 5 11 DbreakMOD  
 Dplcap 10 5 DplcapMOD

Ebreak 11 7 17 18 33.45  
 Eds 14 8 5 8 1  
 Egs 13 8 6 8 1  
 Esg 6 10 6 8 1  
 Evthres 6 21 19 8 1  
 Evtemp 20 6 18 22 1

It 8 17 1

Lgate 1 9 3.6e-9  
 Ldrain 2 5 1.0e-9  
 Lsource 3 7 3.3e-9

RLgate 1 9 36  
 RLdrain 2 5 10  
 RLsource 3 7 33

Mmed 16 6 8 8 MmedMOD  
 Mstro 16 6 8 8 MstroMOD  
 Mweak 16 21 8 8 MweakMOD

Rbreak 17 18 RbreakMOD 1  
 Rdrain 50 16 RdrainMOD 2.15e-3  
 Rgate 9 20 2.1  
 RSLC1 5 51 RSLCMOD 1e-6  
 RSLC2 5 50 1e3  
 Rsource 8 7 RsourceMOD 9e-4  
 Rvthres 22 8 RvthresMOD 1  
 Rvtemp 18 19 RvtempMOD 1  
 S1a 6 12 13 8 S1AMOD  
 S1b 13 12 13 8 S1BMOD  
 S2a 6 15 14 13 S2AMOD  
 S2b 13 15 14 13 S2BMOD

Vbat 22 19 DC 1

ESLC 51 50 VALUE={{(V(5,51)/ABS(V(5,51)))\*(PWR(V(5,51))/(1e-6\*500),10))}}

.MODEL DbodyMOD D (IS=7.5E-12 IKF=17 N=1.01 RS=2.1e-3 TRS1=2e-3 TRS2=2e-7  
 + CJO=1.9e-9 M=0.57 TT=9e-11 XT1=2.6)

.MODEL DbreakMOD D (RS=8e-2 TRS1=1e-3 TRS2=-8.9e-6)

.MODEL DplcapMOD D (CJO=1.75e-9 IS=1e-30 N=10 M=0.4)

.MODEL MmedMOD NMOS (VTO=2.1 KP=30 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=2.1 T\_ABS=25)

.MODEL MstroMOD NMOS (VTO=2.51 KP=650 IS=1e-30 N=10 TOX=1 L=1u W=1u T\_ABS=25)

.MODEL MweakMOD NMOS (VTO=1.67 KP=0.1 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=21 RS=0.1 T\_ABS=25)

.MODEL RbreakMOD RES (TC1=8.3e-4 TC2=-9e-7)

.MODEL RdrainMOD RES (TC1=2.3e-3 TC2=5e-6)

.MODEL RSLCMOD RES (TC1=1e-4 TC2=1e-6)

.MODEL RsourceMOD RES (TC1=8e-3 TC2=1e-6)

.MODEL RvthresMOD RES (TC1=-2.3e-3 TC2=-9e-6)

.MODEL RvtempMOD RES (TC1=-3e-3 TC2=2e-7)

.MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-4 VOFF=-2)

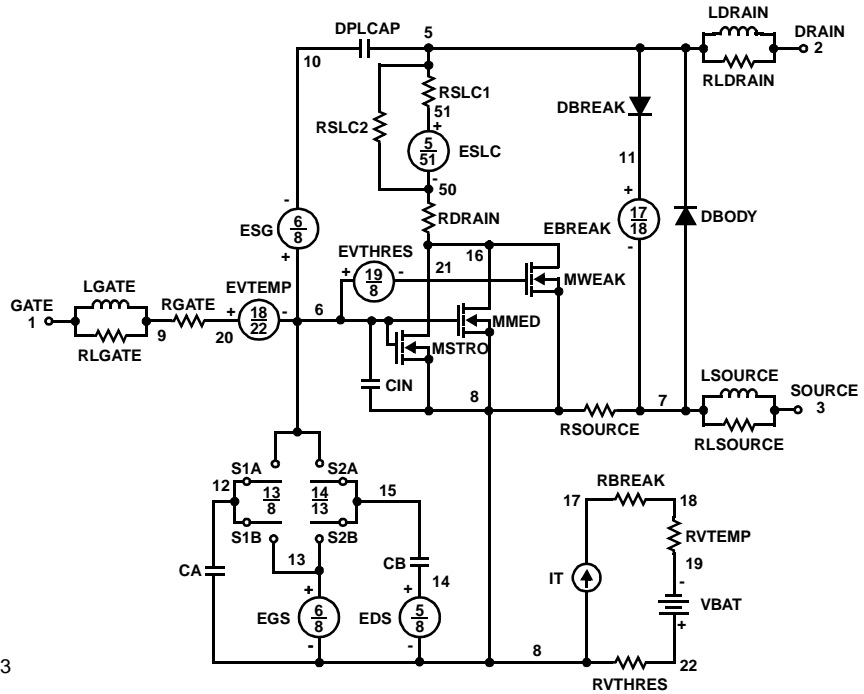
.MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-2 VOFF=-4)

.MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1 VOFF=-0.5)

.MODEL S2BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-0.5 VOFF=-1)

.ENDS

Note: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.







### PSPICE Thermal Model

REV 23 December 2003

FDP8870T

CTHERM1 TH 6 1e-3  
 CHERM2 6 5 2e-3  
 CHERM3 5 4 3e-3  
 CHERM4 4 3 9e-3  
 CHERM5 3 2 1e-2  
 CHERM6 2 TL 2e-2

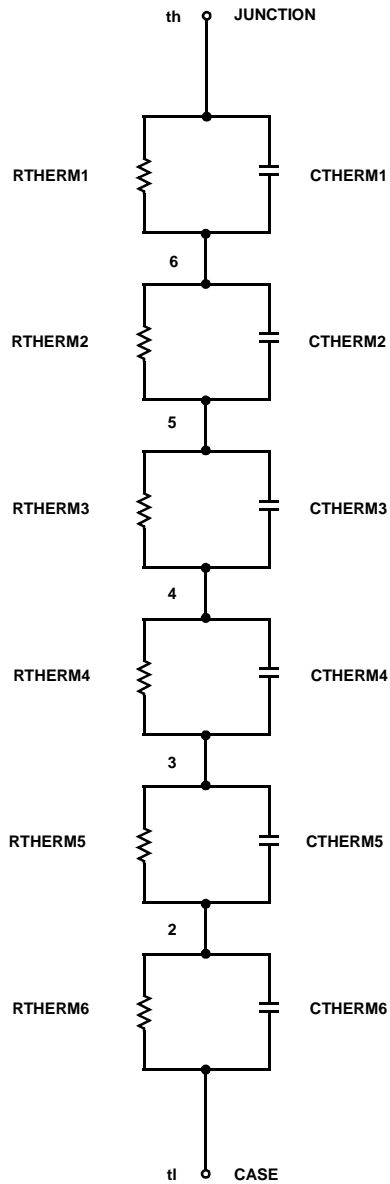
RHERM1 TH 6 3e-2  
 RHERM2 6 5 8e-2  
 RHERM3 5 4 1.1e-1  
 RHERM4 4 3 1.6e-1  
 RHERM5 3 2 1.72e-1  
 RHERM6 2 TL 2e-1

### SABER Thermal Model

SABER thermal model FDP8870T  
 template thermal\_model th tl

```
thermal_c th, tl
{
    ctherm.ctherm1 th 6 =1e-3
    ctherm.ctherm2 6 5 =2e-3
    ctherm.ctherm3 5 4 =3e-3
    ctherm.ctherm4 4 3 =9e-3
    ctherm.ctherm5 3 2 =1e-2
    ctherm.ctherm6 2 tl =2e-2
```





```
rtherm.rtherm1 th 6 =3e-2
rtherm.rtherm2 6 5 =8e-2
rtherm.rtherm3 5 4 =1.1e-1
rtherm.rtherm4 4 3 =1.6e-1
rtherm.rtherm5 3 2 =1.72e-1
rtherm.rtherm6 2 tl =2e-1
}
```





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| Build it Now™   | F-PFS™  | Power-SPM™  | <b>power</b><br>the franchise   |
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| CorePOWER™  | Global Power ResourceSM   | Programmable Active Droop™  | TinyBuck™   |
| CROSSVOLT™  | Green FPS™  | QFET®   | TinyLogic®  |
| CTL™  | Green FPS™ e-Series™  | QS™   | TINYOPTO™   |
| Current Transfer Logic™   | GTO™  | Quiet Series™   | TinyPower™  |
| EcoSPARK®   | IntelliMAX™   | RapidConfigure™   | TinyPWM™  |
| EfficientMax™   | ISOPLANAR™  | Saving our world 1mW at a time™   | TinyWire™   |
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