

Application Note No. 051

SIEGET45 - Low Noise Amplifier with BFP520
Transistor at 1.9 GHz

RF & Protection Devices



Never stop thinking

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SIEGET45 - Low Noise Amplifier with BFP520 Transistor at 1.9 GHz

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Previous Version:

Page	Subjects (major changes since last revision)
All	Document layout change

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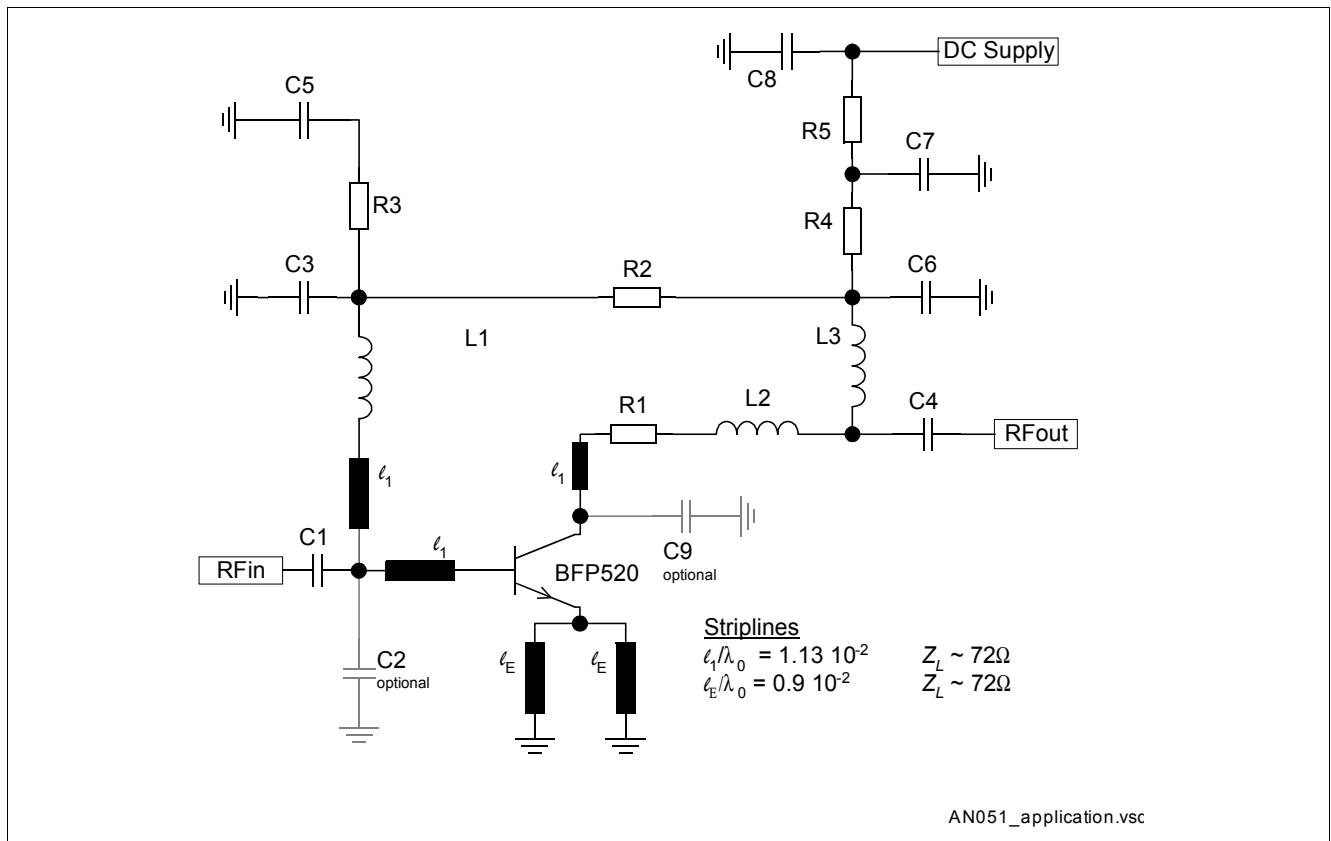
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1 SIEGET[®]45 - Low Noise Amplifier with BFP520 Transistor at 1.9 GHz

This application note describes an example of a Low Noise Amplifier with the BFP520 (5th Generation, SOT343) SIEGET[®]45-line. Improved performance in Gain (16 dB) and OIP_3 (15 dBm) are the outstanding features at low current $I_C = 7 \text{ mA}$ (@ $V_{CC} = 2 \text{ V}$). Consider while designing, the device shows potential instability in the frequency range around 1 GHz. Due to the hard stability matching conditions the 2nd example is still stable over the whole frequency range. The 1st example is unstable at 1 GHz nevertheless the gain exceeds 16.5 dB, noise figure $\sim 1.4 \text{ dB}$.

Table 1 Simulation and Measurement results

Data @ 1.9 GHz	Simulation result	Measurement result
Gain (S21)	16 dB	16.5 dB
$R_{LIN} S11 $	13 dB	> 11 dB
$R_{Lout} S22 $	17 dB	> 20 dB
IP_3	-	15 dBm
NF	1.5 dB	1.5 dB


Figure 1 Schematic of the BFP520 Application @ 1.95 GHz

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1.1 PCB Layout and Component Placement

Fixed components: J1: connector / J2: switch / C8: 4.7 uF (Elko)

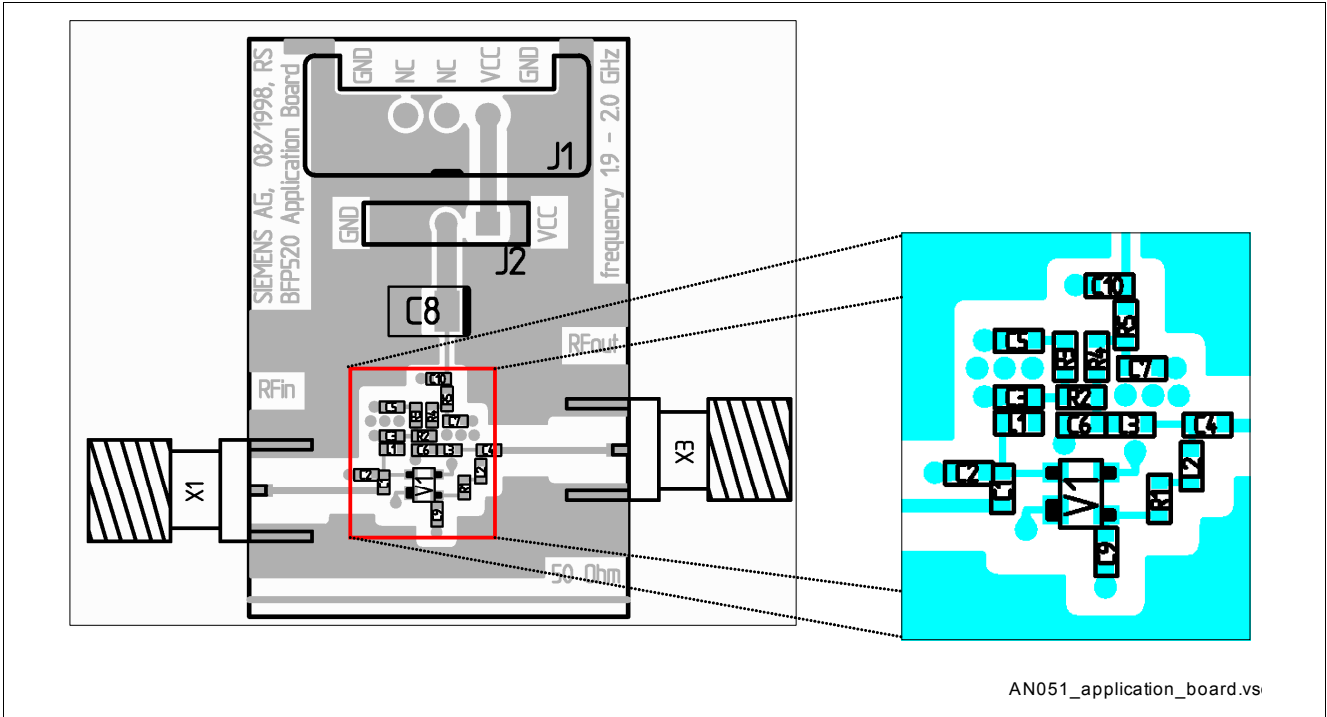


Figure 2 Application board (scale 3:1) - original size 23 x 35 mm

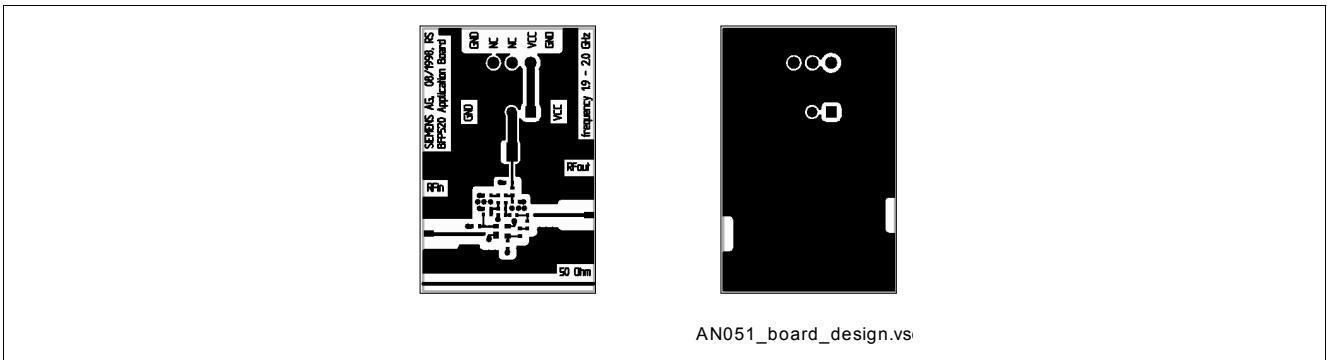


Figure 3 Board design - Top Layer and Layer-2 - scale 1:1

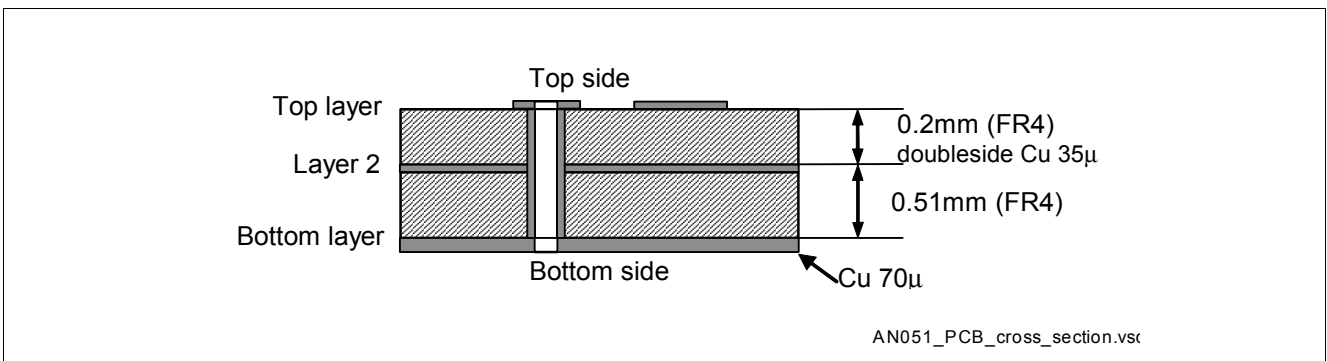


Figure 4 PCB cross section

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1.2 Circuit modeling and simulation

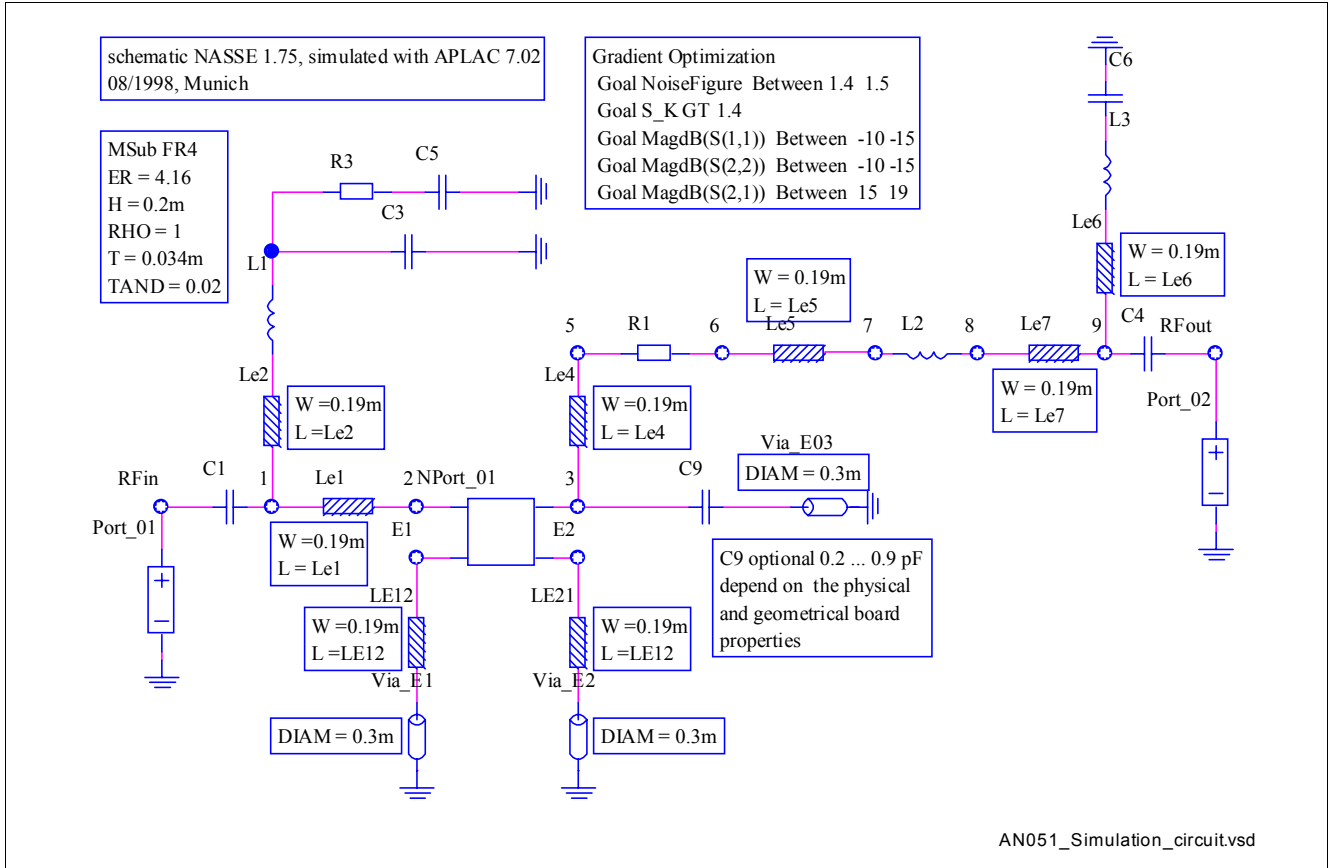


Figure 5 RF Simulation Circuit of the BFP520 Application

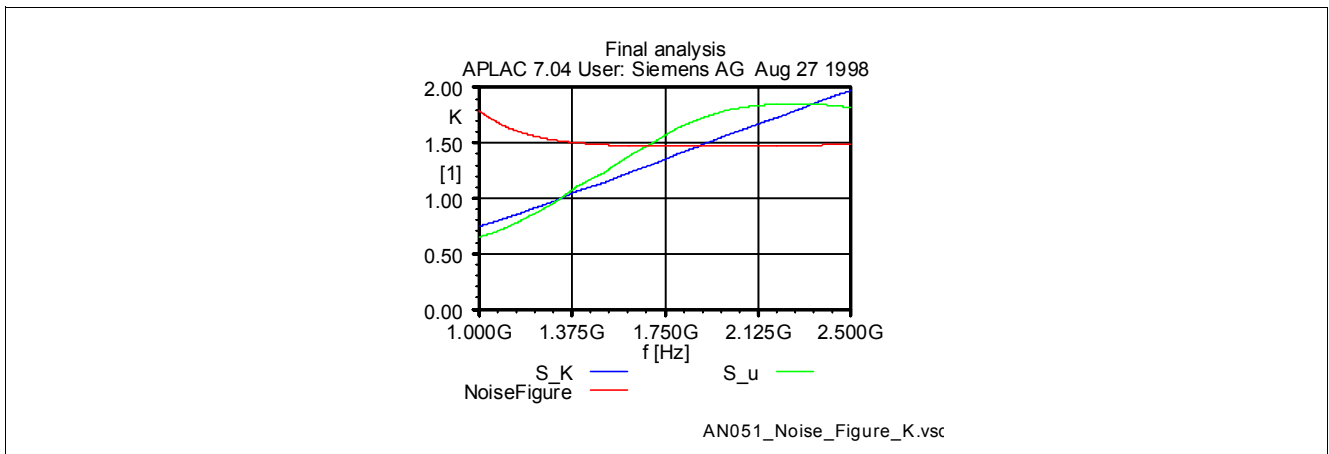


Figure 6 Noise Figure, Stability K (S_k) μ (S_u)

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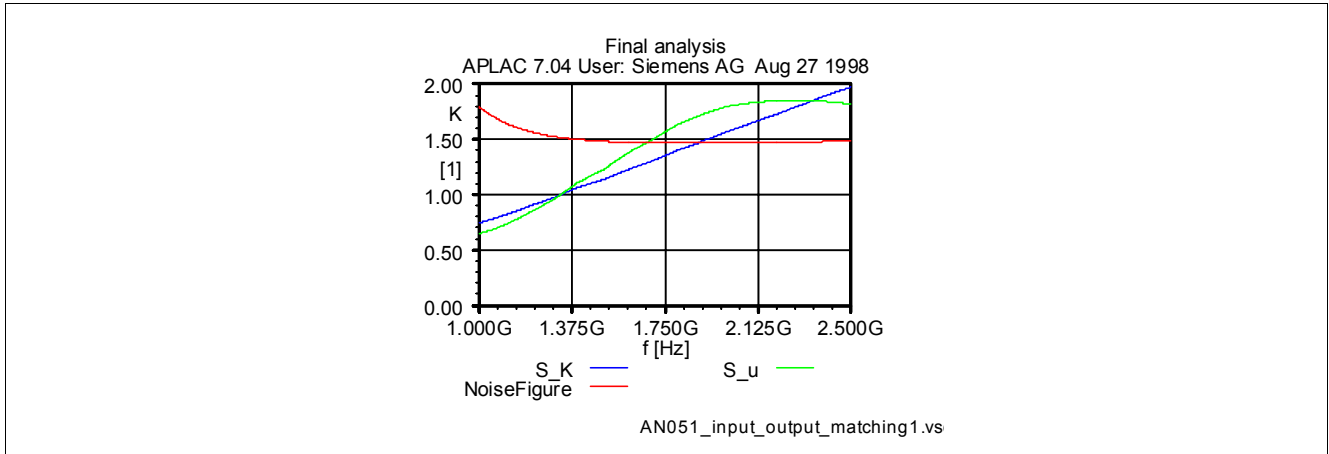


Figure 7 Input and output matching

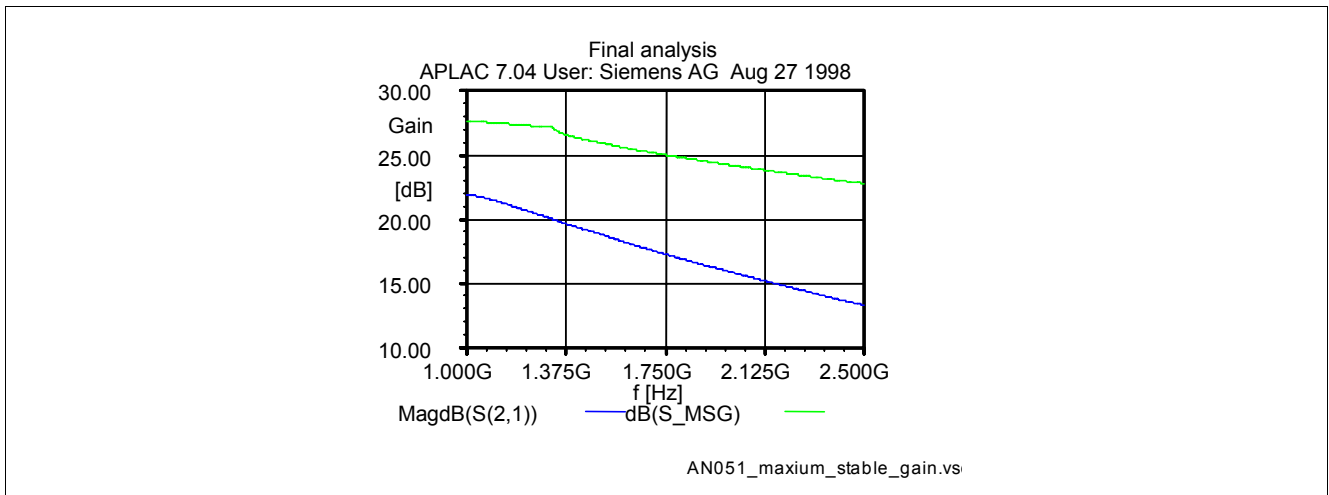


Figure 8 Maximum stable gain and transducer gain

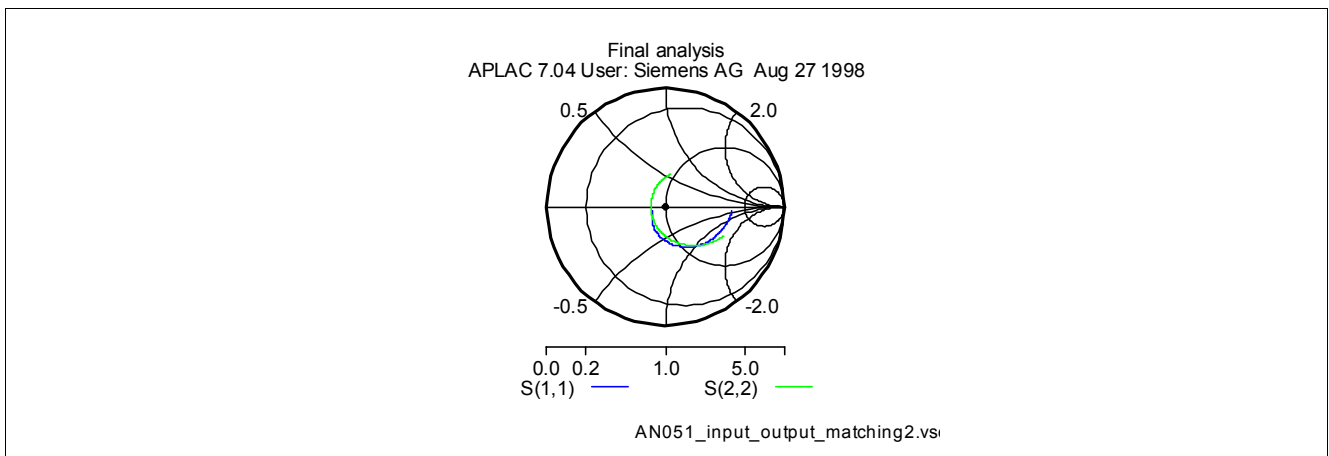


Figure 9 Input and output matching

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1.3 1st application circuit measurement results

Stable under proved condition (50 Ω); best noise and performance. The application shows potential instability at 1 GHz.

Table 2 Components of 1st application

Component	Value	Comment	Component	Value	Comment
C1	15 pF ¹⁾	Input match	L1	12 nH ²⁾	Input match
C2		Improve stability	L2	2.2 nH ²⁾	Output match
C3	15 pF ¹⁾		L3	15 nH ²⁾	Output match
C4	100 pF ¹⁾	Output match			
C4	100 pF ¹⁾		R1	27 Ω ³⁾	Improve stability
C6	15 pF ¹⁾		R2	18 Ω ³⁾	Bias; $I_C = 6$ mA
C7	68 pF ¹⁾		R3	33 Ω ³⁾	
C8	1 nF ¹⁾		R4	33 Ω ³⁾	
C9		Improve stability	R5	0 Ω ³⁾	$V_{CC} = 2$ V

1) Murata

2) Toko 0402

3) S & M 0402

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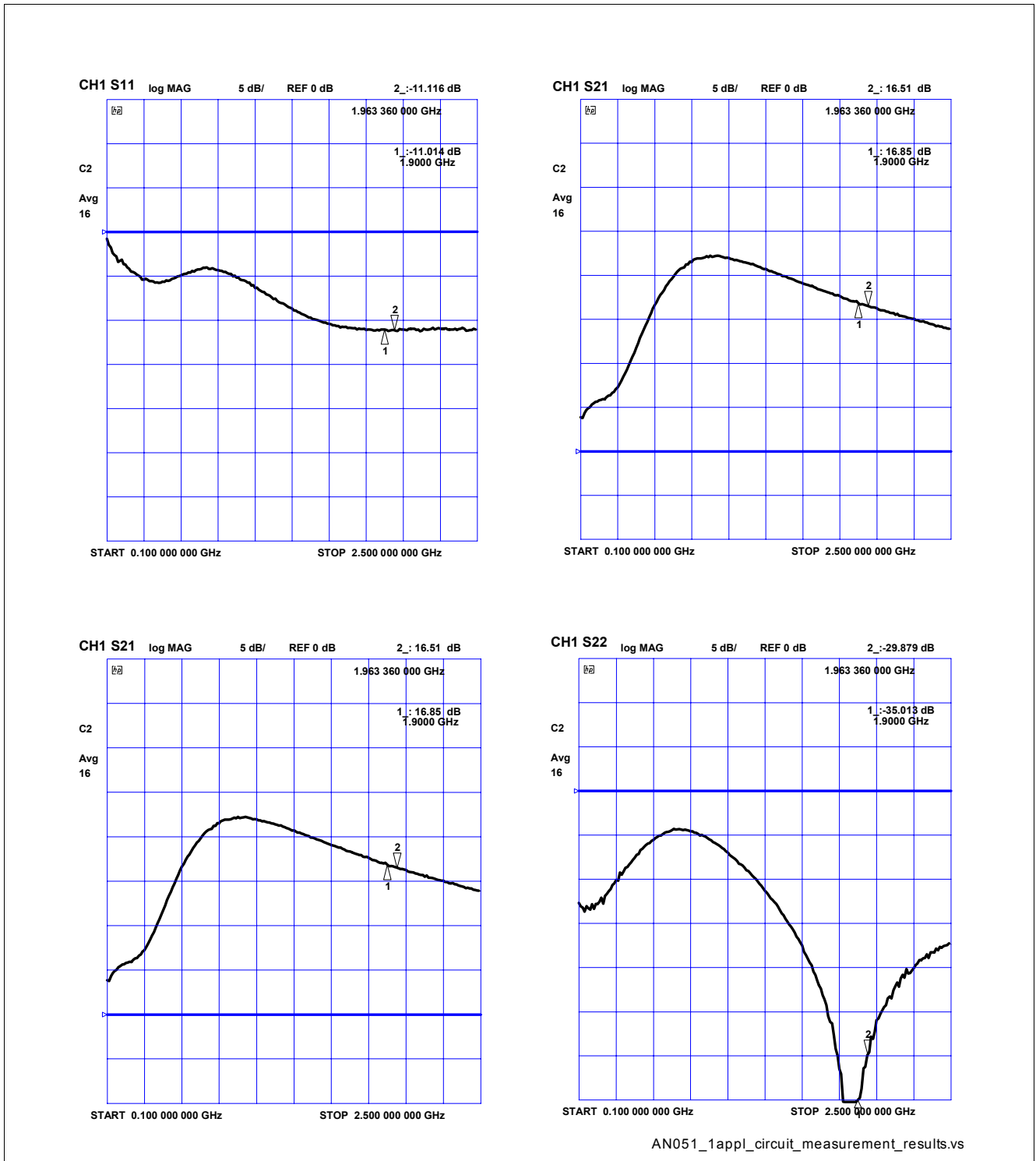
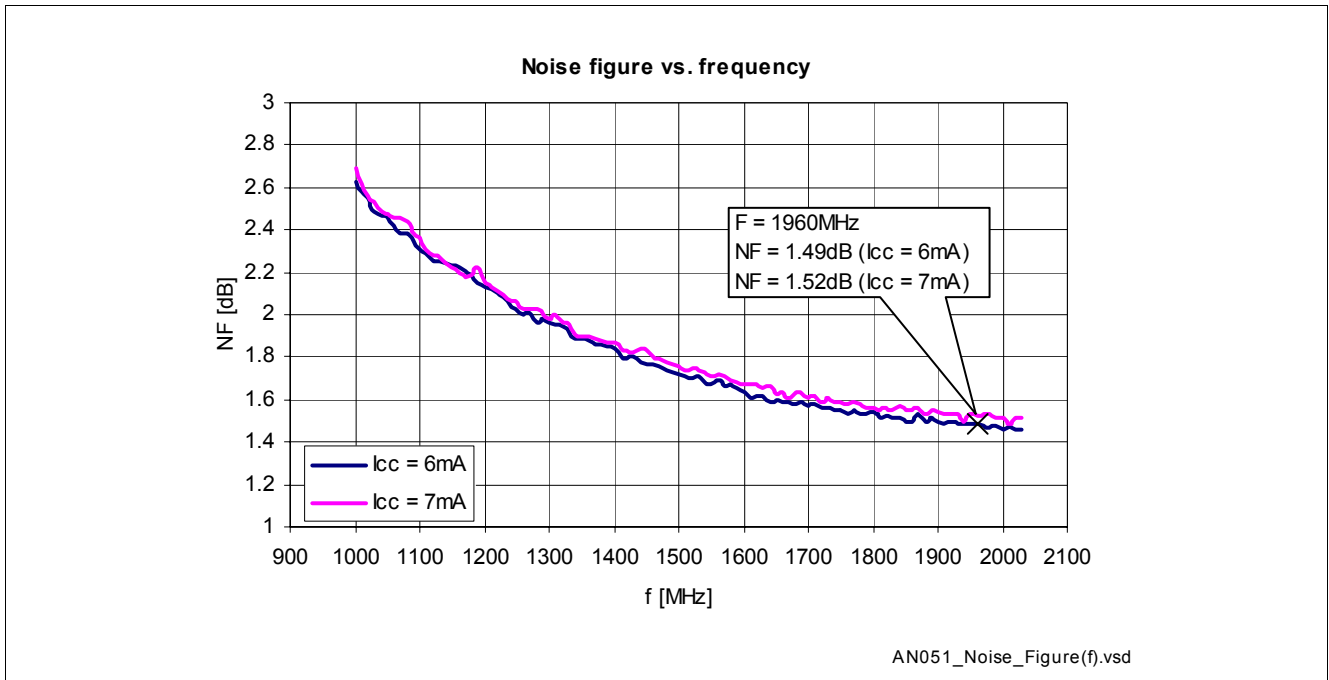


Figure 10 Measurement results of 1st application

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Figure 11 Noise Figure ($V_{CC} = 2\text{ V}$, $T_C = 25\text{ °C}$)

1.4 2nd application circuit measurement results

Absolutely stable over the whole frequency range ($k > 1$, 0.1... 2.5 GHz, see stability circles in [Figure 12](#) and [Figure 13](#)). Less performance in noise figure (1.5 dB) and gain (15 dB) than the application example above.

Table 3 Components of 1st application

Component	Value	Comment	Component	Value	Comment
C1	15 pF ¹⁾	Input match	L1	6.8 nH ²⁾	Input match
C2		Improve stability	L2	1.2 nH ²⁾	Output match
C3	15 pF ¹⁾		L3	5.6 nH ²⁾	Output match
C4	100 pF ¹⁾	Output match			
C4	100 pF ¹⁾		R1	22 Ω ³⁾	Improve stability
C6	15 pF ¹⁾		R2	18 Ω ³⁾	Bias; $I_C = 6\text{ mA}$
C7	68 pF ¹⁾		R3	33 Ω ³⁾	
C8	1 nF ¹⁾		R4	33 Ω ³⁾	
C9	0.5 pF ¹⁾	Improve stability	R5	0 Ω ³⁾	$V_{CC} = 2\text{ V}$

1) Murata

2) Toko 0402

3) S & M 0402

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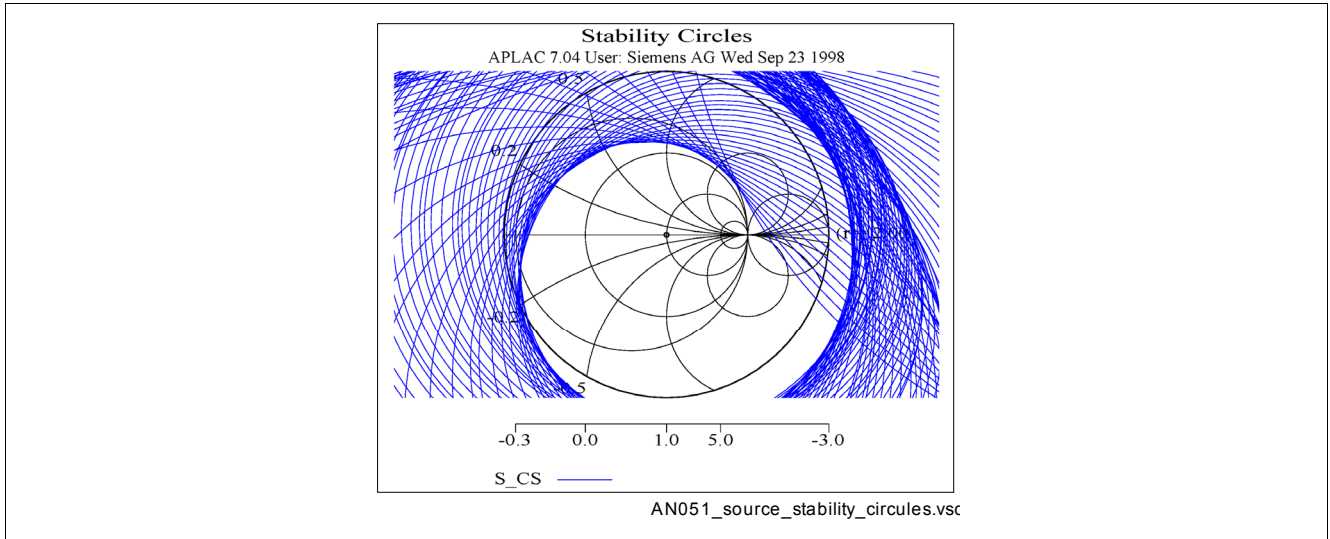


Figure 12 Source stability circles

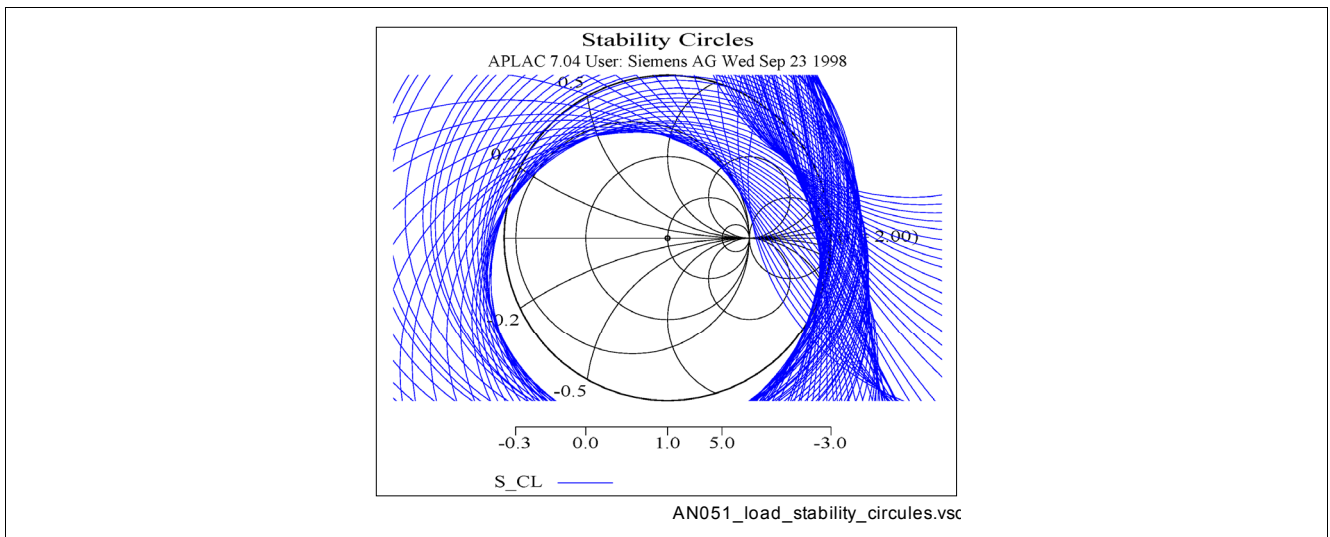
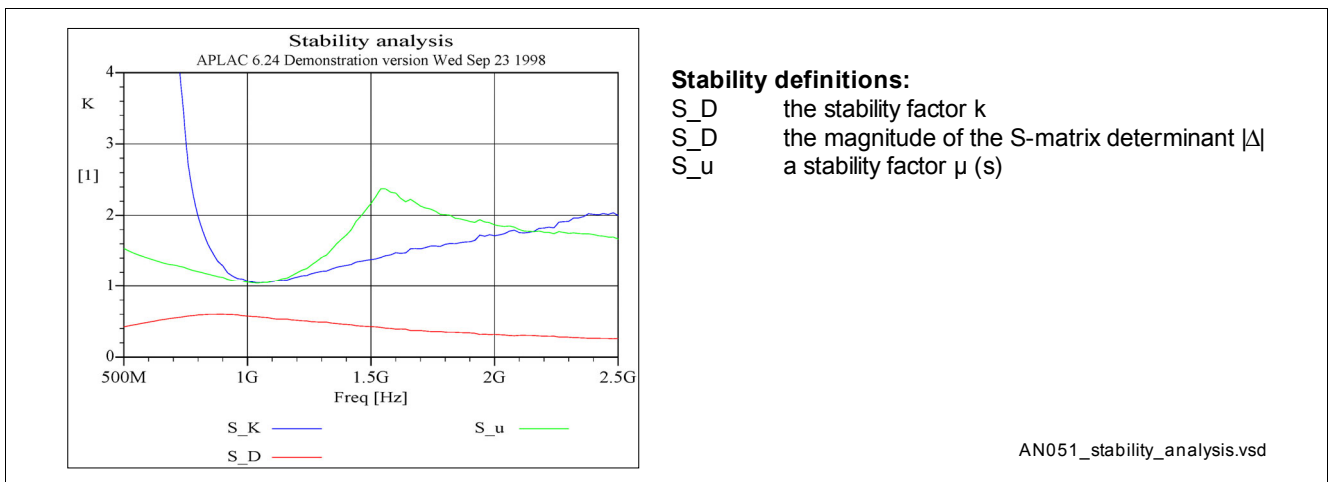


Figure 13 Load stability circles



Stability definitions:

- S_D the stability factor k
- S_D the magnitude of the S-matrix determinant $|\Delta|$
- S_u a stability factor μ (s)

Figure 14 Stability analysis

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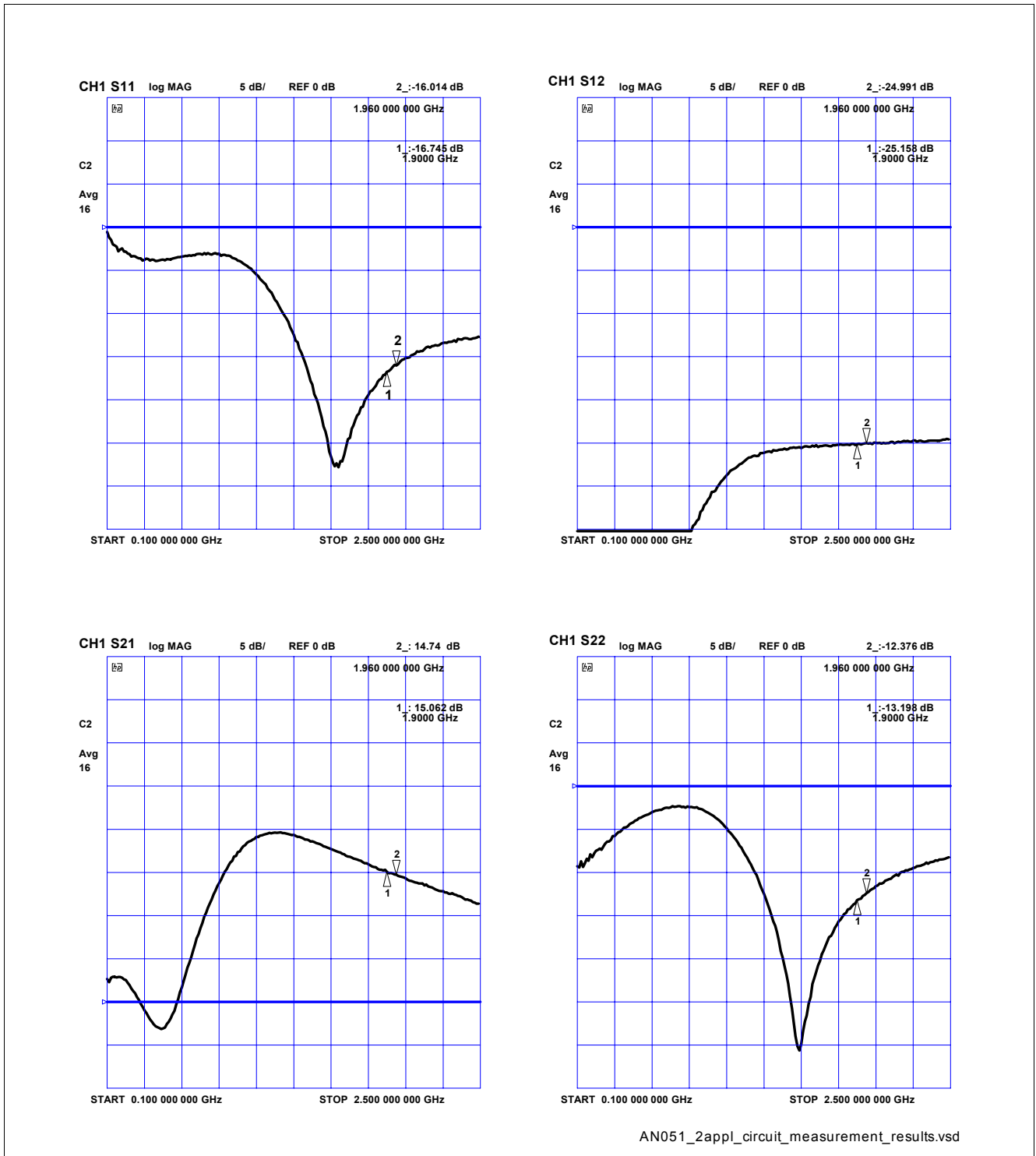


Figure 15 Measurements results of 2nd application