



#### DESCRIPTION

The MP6545 is a three-phase power stage IC designed to drive brushless DC (BLDC) motors or other loads.

The MP6545 operates across a 4.5V to 45V input voltage ( $V_{IN}$ ) range. It can deliver an output current ( $I_{OUT}$ ) up to 2.5A per phase, depending on the ambient temperature ( $T_A$ ) and PCB layout.

Internal safety and diagnostic features include over-current protection (OCP), input over-voltage protection (OVP), input under-voltage lockout (UVLO), and thermal shutdown.

The MP6545 has separate high-side (HS) and low-side (LS) input pins for each output pin. The MP6545 is available in QFN-28 (4mmx5mm) and TSSOP-28EP packages.

#### **FEATURES**

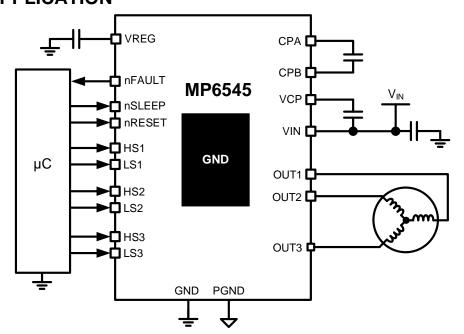
- 4.5V to 45V Operating Input Voltage (V<sub>IN</sub>) Range
- 2.5A Maximum Output Current (I<sub>OUT MAX</sub>)
- Three-Channel Half-Bridge (HB) Driver
- Low 150mΩ On Resistance (R<sub>DS(ON)</sub>) per MOSFET
- Protection Functions Include:
  - Over-Current Protection (OCP)
  - Over-Voltage Protection (OVP)
  - Under-Voltage Lockout (UVLO)
  - Over-Temperature (OT) Shutdown
  - Fault Indication Output
- Available in QFN-28 (4mmx5mm) and TSSOP-28EP Packages

### **APPLICATIONS**

- Brushless DC (BLDC) Motors
- · Laser Printers and Copiers
- Textile Machines

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## TYPICAL APPLICATION





#### ORDERING INFORMATION

Part Number*	Package	Top Marking	MSL Rating
MP6545GV	QFN-28 (4mmx5mm)	See Below	2
MP6545GF	TSSOP-28EP	See Below	2a

<sup>\*</sup> For Tape & Reel, add suffix -Z (e.g. MP6545GV-Z).

## **TOP MARKING (MP6545GF)**

MPSYYWW MP6545

LLLLLLLL

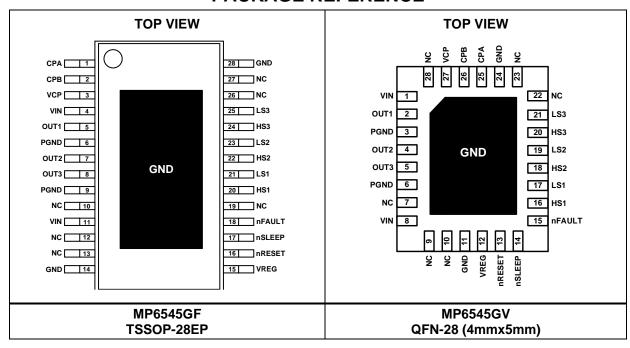
**TOP MARKING (MP6545GV)** 

MPSYWW MP6545

LLLLLL

MPS: MPS prefix YY: Year code WW: Week code MP6545: Part number LLLLLLLL: Lot number MPS: MPS prefix YY: Year code WW: Week code MP6545: Part number LLLLL: Lot number

### **PACKAGE REFERENCE**



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<sup>\*</sup> For Tape & Reel, add suffix -Z (e.g. MP6545GF-Z).



## **PIN FUNCTIONS**

Pin # (QFN)	Pin # (TSSOP)	Name	Description
1, 8	4, 11	VIN	<b>Input supply voltage.</b> Decouple the VIN pin to ground using a minimum 100nF ceramic capacitor. Additional bulk capacitance may be required.
2	5	OUT1	Output terminal 1.
3, 6	6, 9	PGND	Power ground for half H-bridge outputs.
4	7	OUT2	Output terminal 2.
5	8	OUT3	Output terminal 3.
11, 24	14, 28	GND	Signal ground.
12	15	VREG	Internal regulator. Connect a 1µF, 16V ceramic capacitor (X7R) to ground.
13	16	nRESET	<b>Reset input.</b> Pull the nRESET pin active low to reset the protection circuits and disable the outputs. This pin has an internal pull-down resistor.
14	17	nSLEEP	<b>Sleep mode input.</b> Pull the nSLEEP pin logic low to enter low-power sleep mode. This pin has an internal pull-down resistor.
15	18	nFAULT	<b>Fault indication.</b> The nFAULT pin is an open-drain output. If used, nFAULT requires an external pull-up resistor. nFAULT is in logic low under fault conditions.
16	20	HS1	<b>High input turns on the OUT1 high-side MOSFET (HS FET).</b> The HS1 pin has an internal pull-down resistor.
17	21	LS1	<b>High input turns on the OUT1 low-side MOSFET (LS FET).</b> The LS1 pin has an internal pull-down resistor.
18	22	HS2	<b>High input turns on the OUT2 HS-FET.</b> The HS2 pin has an internal pull-down resistor.
19	23	LS2	<b>High input turns on the OUT2 LS-FET.</b> The LS2 pin has an internal pull-down resistor.
20	24	HS3	<b>High input turns on the OUT3 HS-FET.</b> The HS3 pin has an internal pull-down resistor.
21	25	LS3	<b>High input turns on the OUT3 LS-FET.</b> The LS3 pin has an internal pull-down resistor.
25	1	CPA	Charge pump capacitor. Connect a 100nF ceramic capacitor that is rated for
26	2	CPB	V <sub>IN</sub> between the CPA and CPB pins.
27	3	VCP	<b>Charge pump output.</b> The VCP pin requires a 1µF, 16V ceramic capacitor to VIN.

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## **ABSOLUTE MAXIMUM RATINGS (1)** Supply voltage (V<sub>IN</sub>) .....-0.3V to +48V OUTx voltage (V<sub>OUT1</sub>, V<sub>OUT2</sub>, V<sub>OUT3</sub>)..... .....-0.7V to +48V VCP, CPB ......V<sub>IN</sub> to V<sub>IN</sub> + 6.5V PGND to GND .....-0.3V to +0.3V All other pins to GND .....-0.3V to +6.5V Continuous power dissipation (T<sub>A</sub> = 25°C) (2) TSSOP.......3.9W Storage temperature ...... -55°C to +150°C Junction temperature ......150°C Lead temperature (solder) ......260°C ESD Ratings Human body model (HBM) ..... ±2kV Charged device model (CDM) ..... ±2kV Recommended Operating Conditions (3) Supply voltage (V<sub>IN</sub>) ......4.5V to 45V PGND to GND .....-0.2V to +0.2V Operating junction temp (T<sub>J</sub>).... -40°C to +125°C

Thermal Resistance (4)	$oldsymbol{ heta}$ JA	$\boldsymbol{\theta}$ JC
QFN-28 (4mmx5mm)	40	9°C/W
TSSOP-28EP		

#### Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature,  $T_J$  (MAX), the junction-to-ambient thermal resistance,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . The maximum allowable continuous power dissipation at any ambient temperature is calculated by  $P_D$  (MAX) =  $(T_J$  (MAX)  $T_A)$  /  $\theta_{JA}$ . Exceeding the maximum allowable power dissipation can produce an excessive die temperature, which may cause the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, a 4-layer PCB.



## **ELECTRICAL CHARACTERISTICS**

 $V_{IN} = 24V$ ,  $T_A = 25$ °C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Power Supply						
Input supply voltage	V <sub>IN</sub>		4.5		45	V
Quiescent current	I <sub>INQ</sub>	nSLEEP = 1, with no load		2.8		mA
Quiescent current	I <sub>INSLEEP</sub>	nSLEEP = 0		0.9	10	μA
Internal MOSFETs						
Output on resistance	R <sub>DS(ON)_HS</sub>	$I_{OUT} = 1A, T_J = 25^{\circ}C$		135	170	mΩ
Output on resistance	R <sub>DS(ON)_LS</sub>	$I_{OUT} = 1A, T_J = 25^{\circ}C$		150	185	mΩ
Body diode forward voltage	VF	I <sub>OUT</sub> = 1A			1.1	V
Control Logic Inputs						
Logic-low input threshold	V <sub>IL</sub>				0.8	V
Logic-high input threshold	ViH		2			V
Logic input current	I <sub>IN_H</sub>	V <sub>IN</sub> = 5V	-100		+100	μA
Logic input current	I <sub>IN_L</sub>	$V_{IN} = 0V$	-20		+20	μA
Internal pull-down resistance	R <sub>PD</sub>	To GND		100		kΩ
nFAULT Output						
Output low voltage	Vol	I <sub>OUT</sub> = 5mA			0.5	V
Output high leakage current	Іон	Vout = 5V			1	μA
Protection Circuits						
V <sub>IN</sub> under-voltage lockout (UVLO) rising threshold	Vuvlo				4.5	V
V <sub>IN</sub> UVLO hysteresis	$\Delta V_{UVLO}$			300		mV
V <sub>IN</sub> over-voltage protection (OVP) threshold	V <sub>OVP</sub>		45		48	V
Over-current (OC) trip level	I <sub>OCP1</sub>	Sinking	3	4.5		Α
Over-current (OC) trip level	locp2	Sourcing	3	4.5		Α
OC deglitch time	tocp			1		μs
Thermal shutdown	T <sub>TSD</sub>			165		°C
Thermal shutdown hysteresis	$\DeltaT_TSD$			15		°C



# TYPICAL TIMIMG CHARACTERISTICS

 $V_{IN} = 24V$ ,  $T_A = 25$ °C, unless otherwise noted.

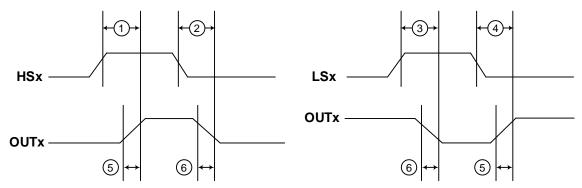


Figure 1: Timing Diagram

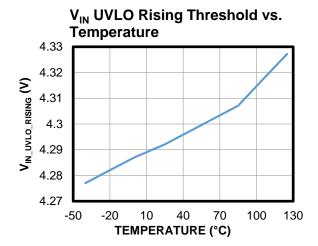
**Table 1: Timing Characteristics** 

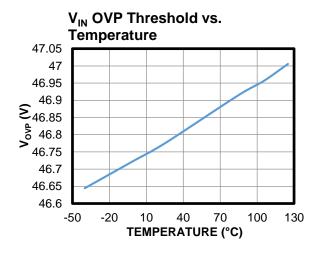
Parameter	Symbol	Condition	Min	Тур	Max	Units
HSx high to OUTx high delay time	t <sub>1</sub>	10mA load connected	40		360	ns
HSx low to OUTx low delay time	t <sub>2</sub>	from OUTx to GND	40		360	ns
LSx high to OUTx low delay time	<b>t</b> 3	10mA load connected	40		360	ns
LSx low to OUTx high delay time	t <sub>4</sub>	from OUTx to VIN	40		360	ns
Output rising time	<b>t</b> 5		1		55	ns
Output falling time	t <sub>6</sub>		1		165	ns
Dead time					80	ns

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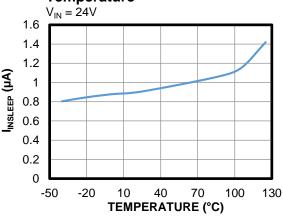


## TYPICAL CHARACTERISTICS

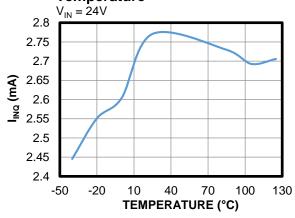




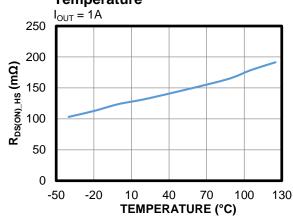




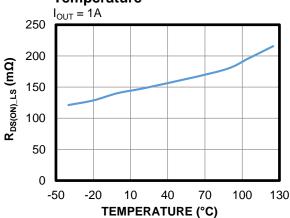




# HS-FET On Resistance vs. Temperature

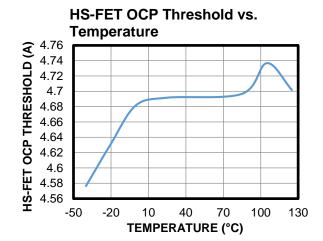


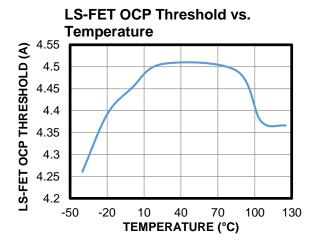
# LS-FET On Resistance vs. Temperature





## TYPICAL CHARACTERISTICS (continued)



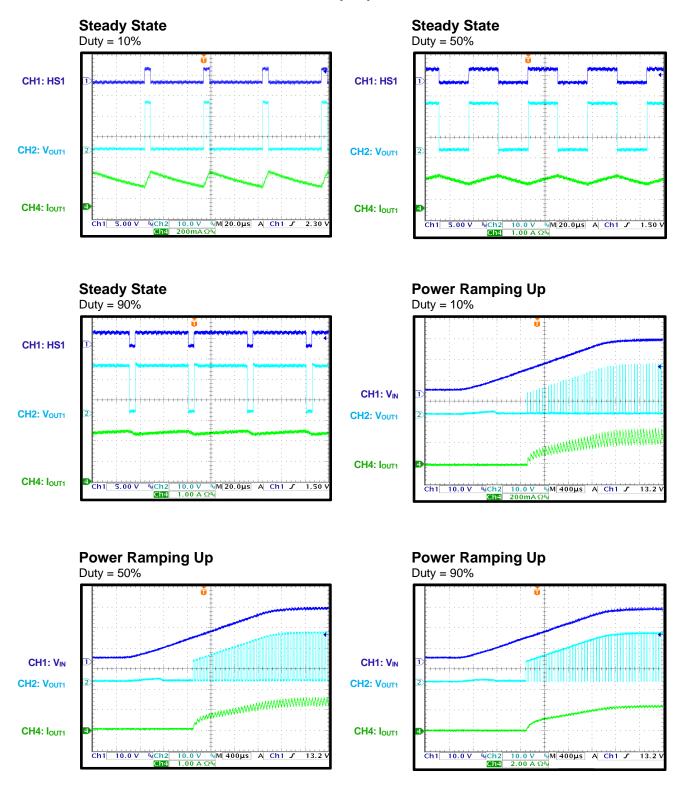


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### TYPICAL PERFORMANCE CHARACTERISTICS

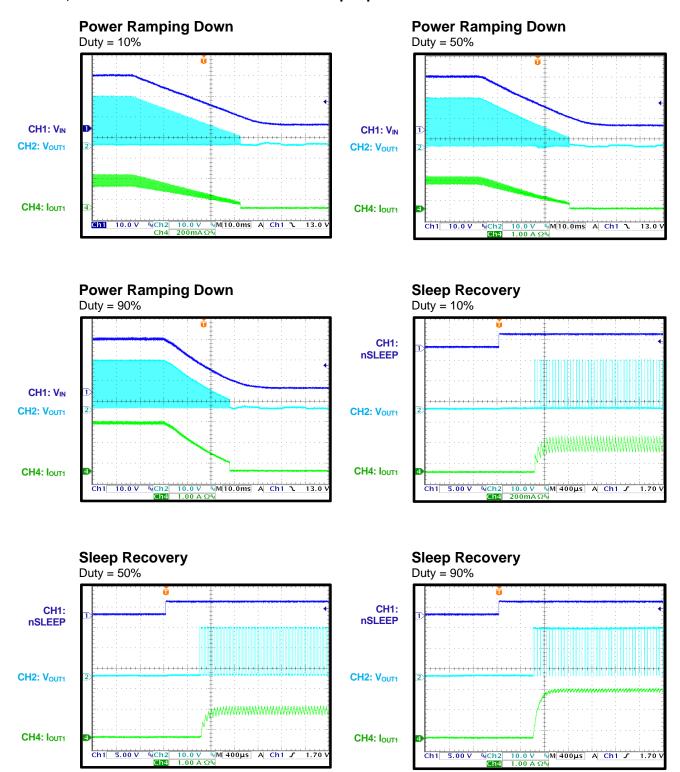
 $V_{IN}$  = 24V, phase 1: switching with 20kHz frequency, phase 2: LS-FET on, phase 3: disabled,  $T_A$  = 25°C, resistor + inductor load:  $4\Omega$  + 0.34mH per phase with star connection.





## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

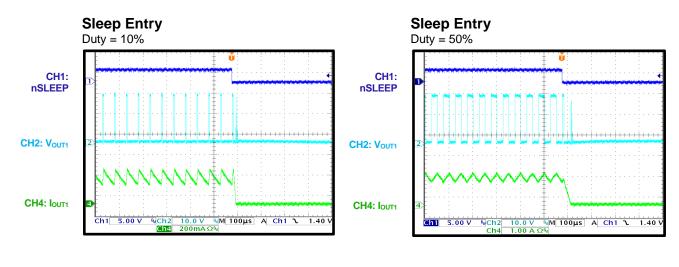
 $V_{IN}$  = 24V, phase 1: switching with 20kHz frequency, phase 2: LS-FET on, phase 3: disabled,  $T_A$  = 25°C, resistor + inductor load:  $4\Omega$  + 0.34mH per phase with star connection.

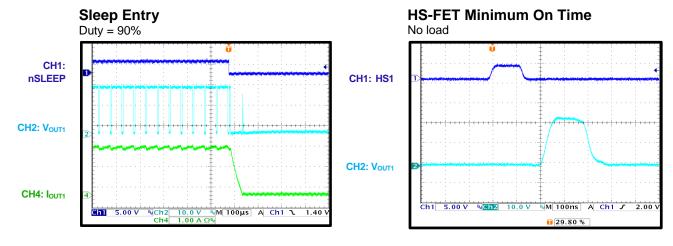


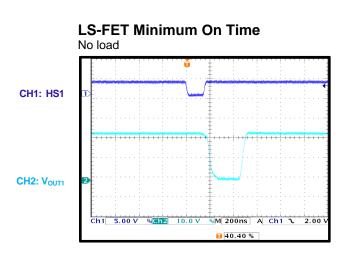


## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{IN}$  = 24V, phase 1: switching with 20kHz frequency, phase 2: LS-FET on, phase 3: disabled,  $T_A$  = 25°C, resistor + inductor load:  $4\Omega$  + 0.34mH per phase with star connection.









## **FUNCTIONAL BLOCK DIAGRAM**

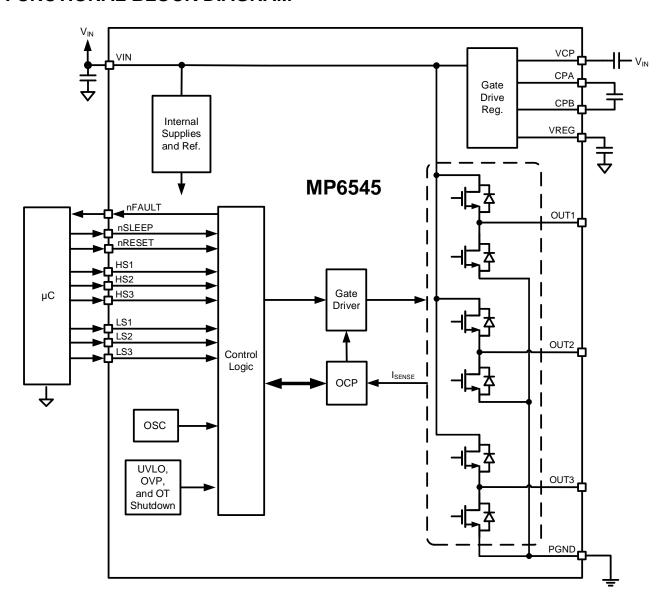


Figure 2: Functional Block Diagram



#### **OPERATION**

The MP6545 is a three-phase power stage designed to drive brushless DC (BLDC) motors or other loads. It integrates six N-channel power MOSFETs connected as three half H-bridges, with a 2.5A current capability. The device operates across a wide 4.5V to 45V supply input voltage ( $V_{\text{IN}}$ ) range.

#### nSLEEP and nRESET

Pull the nSLEEP pin low to force the MP6545 to enter a low-power sleep mode. In this mode, the gate driver charge pump stops, and all the internal circuits and outputs are disabled. All inputs are ignored when nSLEEP is active low.

When the MP6545 exits sleep mode, 600µs must pass before driving the motor. This allows the internal circuitry to stabilize. nSLEEP has an internal pull-down resistor.

Pull the nRESET pin low to reset the latched protection features, including over-current protection (OCP) and over-voltage protection (OVP), as well as to disable the outputs to a high-impedance (Hi-Z) state.

#### Input Interface

The MP6545 includes three half-bridges that operate independently. The device has high-side (HS) and low-side (LS) inputs for each of the output pins. Table 2 shows the HSx and LSx input logic.

Table 2: HSx and LSx Input Logic

HSx	LSx	OUTx
0	0	Hi-Z
0	1	L
1	0	Н
1	1	Hi-Z

Note that all logic inputs have weak, internal pull-down resistors.

#### **Automatic Synchronous Rectification**

If the output high-side MOSFET (HS-FET) and low-side MOSFET (LS-FET) are turned off, then the recirculation current must continue to flow when driving current through an inductive load. This current is typically passed through the MOSFET body diodes. To prevent excess power dissipation in the body diodes, the MP6545 implements automatic synchronous rectification.

If both the HS-FET and LS-FET are turned off and the voltage on an OUTx pin  $(V_{OUTx})$  is pulled below ground, then the LS-FET turns on until the current flowing through it approaches 0A, or until the HS-FET is commanded to turn on. Similarly, if  $V_{OUTx}$  exceeds  $V_{IN}$ , then the HS-FET turns on until the current approaches 0A, or until the LS-FET turns on.

#### Internal Supplies (V<sub>REG</sub> and V<sub>CP</sub>)

The internal regulators generate a 5V supply  $(V_{REG})$  for the LS gate drive, and a supply exceeding  $V_{IN}$  by 5V  $(V_{CP})$  for the HS gate drive. These supplies require external capacitors.

The VREG pin requires a  $1\mu F$  ceramic capacitor to ground. Meanwhile, the VCP pin requires a  $1\mu F$  ceramic capacitor to VIN. Both capacitors should be X7R ceramic capacitors, and rated for at least 16V.

Connect the charge pump flying capacitor between the CPA and CPB pins using a 100nF ceramic capacitor (X7R) rated for at least the maximum  $V_{\rm IN}$ .

#### **nFAULT**

The MP6545 provides a nFAULT pin to report to the system if a fault condition such as OCP, OVP, or over-temperature protection (OTP) occurs. nFAULT is an open-drain output and is pulled low during fault conditions. If used, nFAULT should be pulled high via an external pull-up resistor.

#### **Over-Current Protection (OCP)**

OCP circuitry limits the current through the HS-FET and LS-FET by disabling the gate driver. If the over-current (OC) limit threshold is reached and lasts for longer than the OC deglitch time (t<sub>OCP</sub>), then all the MOSFETs in the H-bridge are disabled and nFAULT is pulled low. The driver remains disabled until the device is reset by pulling nRESET low, or cycling the power on the MP6545.

OC conditions on both the HS and LS devices (i.e. an over-current condition to ground, supply, or across a motor winding) all result in an OC shutdown.



#### **Over-Voltage Protection (OVP)**

If  $V_{IN}$  exceeds the OVP threshold  $(V_{OVP})$ , then the outputs are disabled and nFAULT is pulled low. The driver remains disabled until the device is reset by pulling nRESET low, or cycling the power on the MP6545.

#### Input **Under-Voltage** Lockout (UVLO) **Protection**

If V<sub>IN</sub> falls below the under-voltage lockout (UVLO) threshold (V<sub>IN UVLO</sub>) at any time, then all circuitry in the device is disabled and the internal logic resets. Once V<sub>IN</sub> exceeds V<sub>UVLO</sub>, the device resumes normal operation.

#### Thermal Shutdown

If the die temperature exceeds safe limits, then all the MOSFETs in the H-bridge are disabled and nFAULT is pulled low. Once the die temperature falls to a safe level, the MP6545 automatically resumes normal operation.



#### APPLICATION INFORMATION

#### Selecting the External Components

Bypass the two VIN pins to GND using a minimum 100nF ceramic capacitor with X7R dielectrics, placed as close to the IC as possible. Place an additional 1µF to 10µF ceramic capacitor close to the 100nF capacitor. Depending on the supply impedance and distance to other large capacitors, electrolytic bulk capacitor may also be required to stabilize VIN.

Connect a 100nF ceramic capacitor rated for V<sub>IN</sub> between the CPA and CPB pins. Connect a 1µF, 16V ceramic capacitor between the VIN and VCP pins.

Connect a 1µF, 16V ceramic capacitor with X7R dielectrics from the VREG pin to GND.

#### **PCB Layout Guidelines**

Efficient PCB layout is critical to achieve stable operation. For the best results, refer to Figure 3 and Figure 4, and follow the guidelines below:

- 1. Place the supply bypass and charge pump capacitors as close to the IC as possible, ideally adjacent to the pins on the same PCB laver.
- 2. Each VIN pin requires a bypass capacitor.
- 3. Place as much copper on the long pads as possible.
- 4. Place large copper areas on the pads and the device's outer copper layer.
- 5. The thermal pad should be soldered directly to the copper on the PCB.
- 6. Add thermal vias to transfer heat to the other PCB layers.

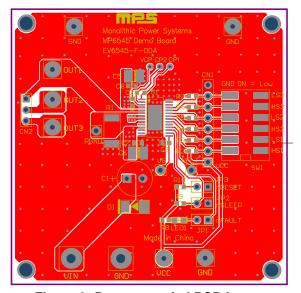


Figure 3: Recommended PCB Layout (MP6545GF)

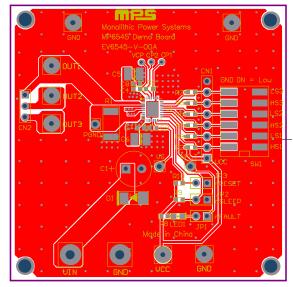
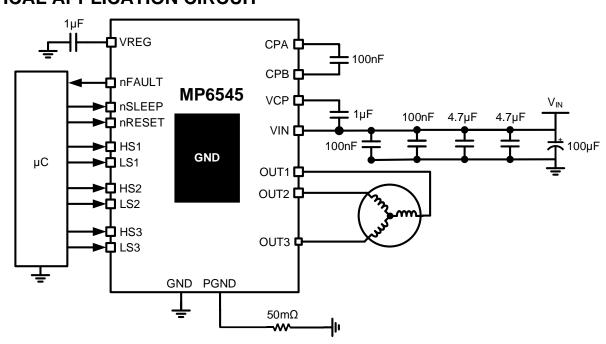


Figure 4: Recommended PCB Layout (MP6545GV)



# **TYPICAL APPLICATION CIRCUIT**

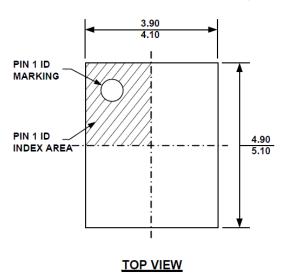


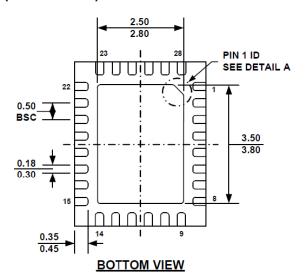
**Figure 5: Typical Application Circuit** 

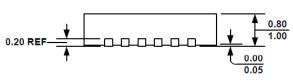


## **PACKAGE INFORMATION**

## QFN-28 (4mm×5mm)





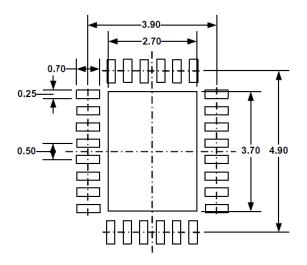


PIN 1 ID OPTION A
0.30x45° TYP.

PIN 1 ID OPTION B
R0.25 TYP.

SIDE VIEW

<u>DETAIL A</u>



RECOMMENDED LAND PATTERN

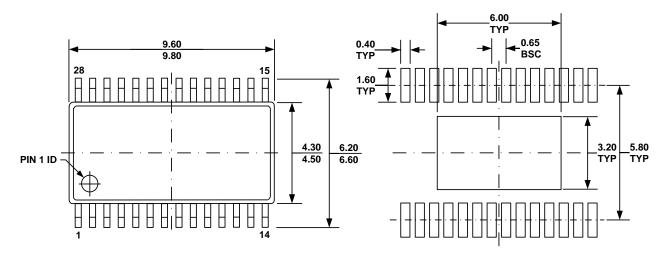
#### NOTE:

- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) EXPOSED PADDLE SIZE DOES NOT INCLUDE MOLD FLASH.
- 3) LEAD COPLANARITY SHALL BE 0.10 MILLIMETER MAX.
- 4) DRAWING CONFORMS TO JEDEC MO-220, VARIATION VHGD-3.
- 5) DRAWING IS NOT TO SCALE.



## PACKAGE INFORMATION (continued)

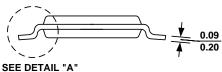
#### TSSOP-28EP



#### **TOP VIEW**

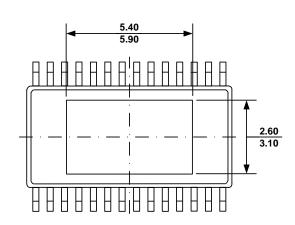
#### **RECOMMENDED LAND PATTERN**



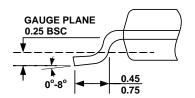


#### **FRONT VIEW**

**SIDE VIEW** 



**BOTTOM VIEW** 



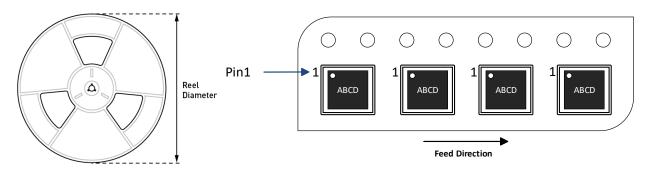
**DETAIL "A"** 

#### NOTE:

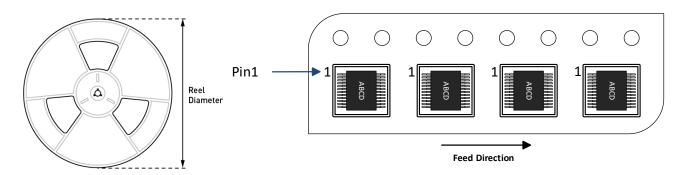
- 1) ALL DIMENSIONS ARE IN MILLIMETERS.
- 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSION OR GATE BURR.
- 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION.
- 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.10 MILLIMETERS MAX.
- 5) DRAWING CONFORMS TO JEDEC MO-153, VARIATION AET.
- 6) DRAWING IS NOT TO SCALE.



## **CARRIER INFORMATION**



Part Number	Package	Quantity/	Quantity/	Reel	Carrier	Carrier
	Description	Reel	Tube	Diameter	Tape Width	Tape Pitch
MP6545GV-Z	QFN-28 (4mmx5mm)	5000	N/A	13in	12mm	8mm



Part Number	Package	Quantity/	Quantity/	Reel	Carrier	Carrier
	Description	Reel	Tube	Diameter	Tape Width	Tape Pitch
MP6545GF-Z	TSSOP-28EP	2500	50	13in	16mm	8mm



## **REVISION HISTORY**

	Revision #	Revision Date	Description	Pages Updated
Ī	1.0	10/13/2022	Initial Release	-
	1.1	5/12/2023	Updated the MSL rating for the MP6545GV to "2" in the Ordering Information section	2

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