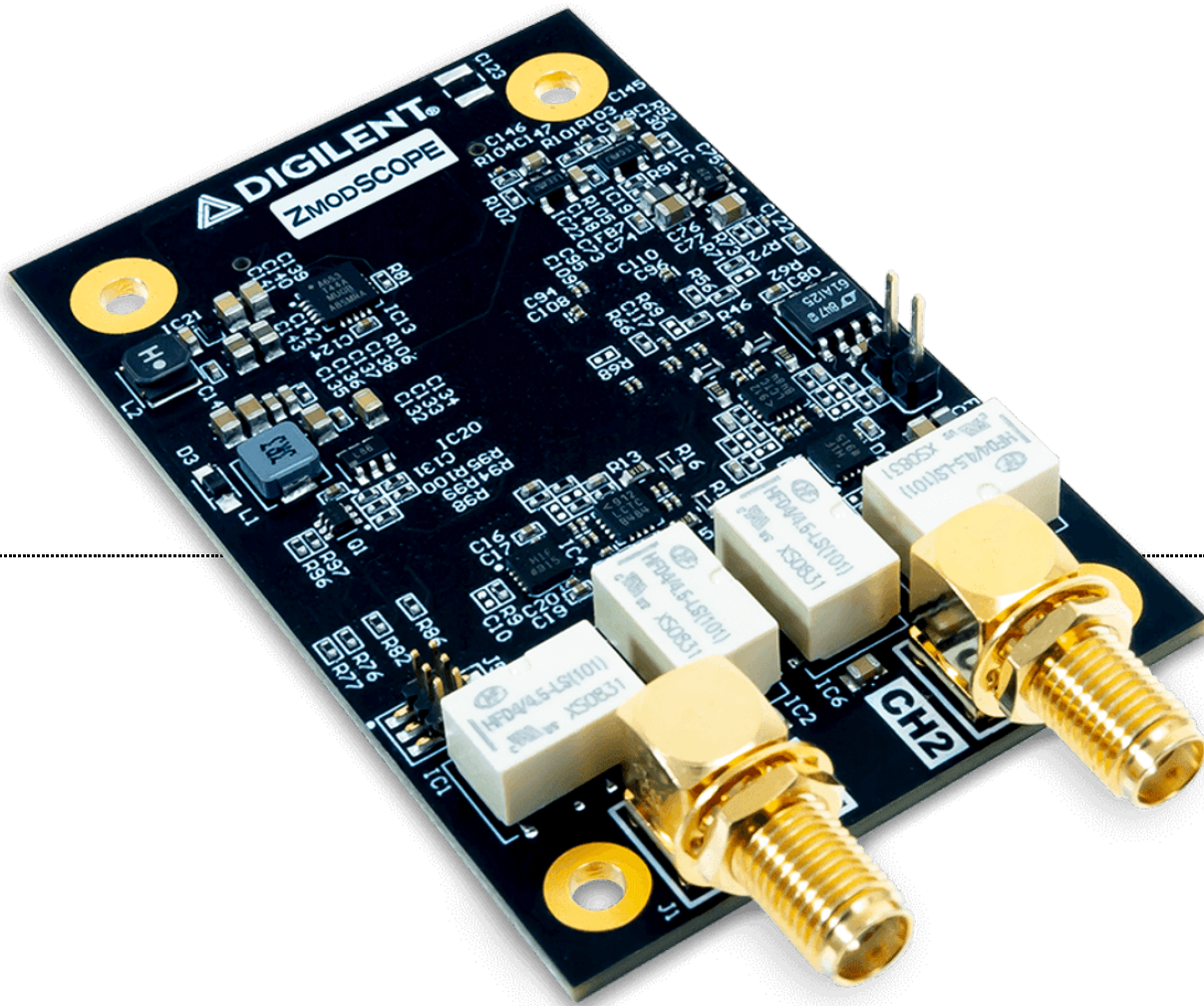


# Zmod Scope Reference Manual

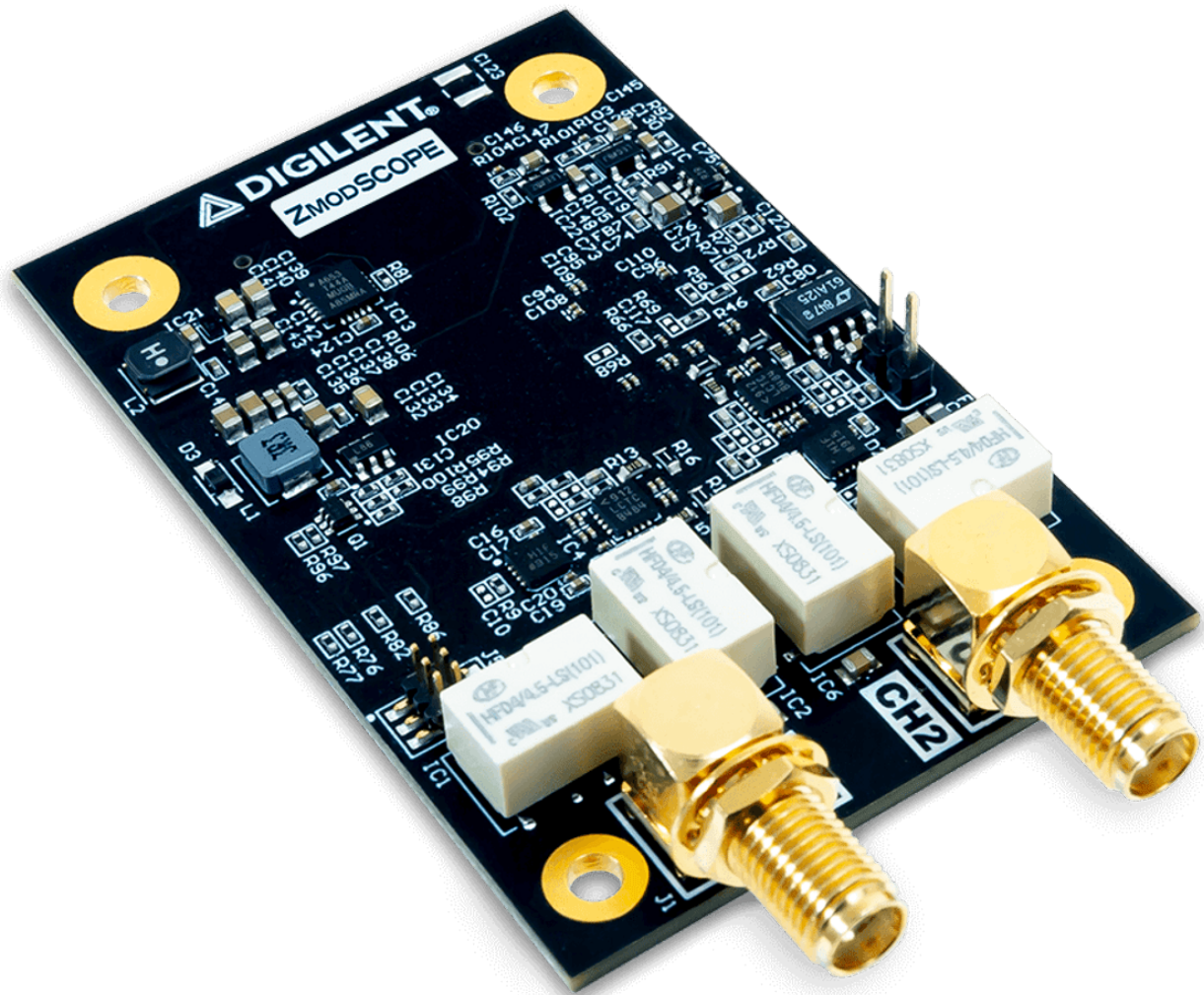
Formerly known as the Zmod ADC().

The Digilent Zmod Scope is an open-source hardware SYZYGY™ <sup>1)</sup>compatible pod containing a dual-channel ADC() and the associated front end. The Zmod Scope is intended to be used with any SYZYGY™ compatible carrier board having the required capabilities.

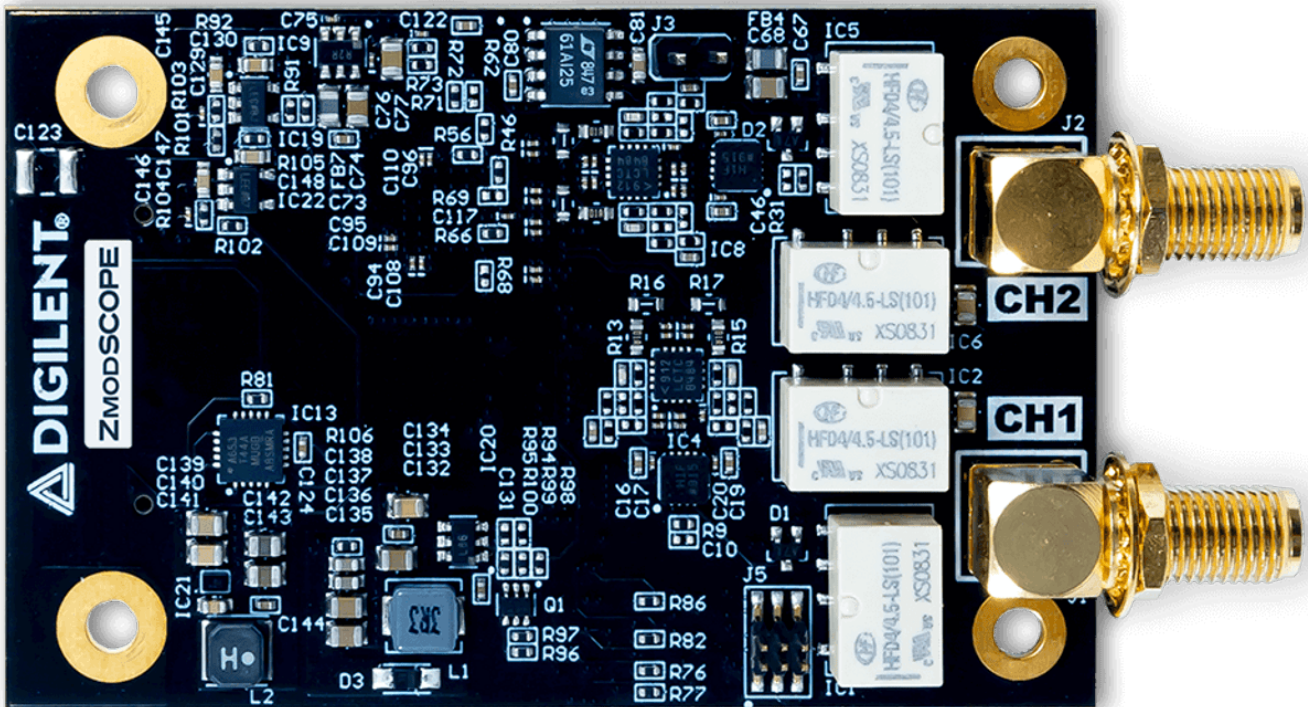
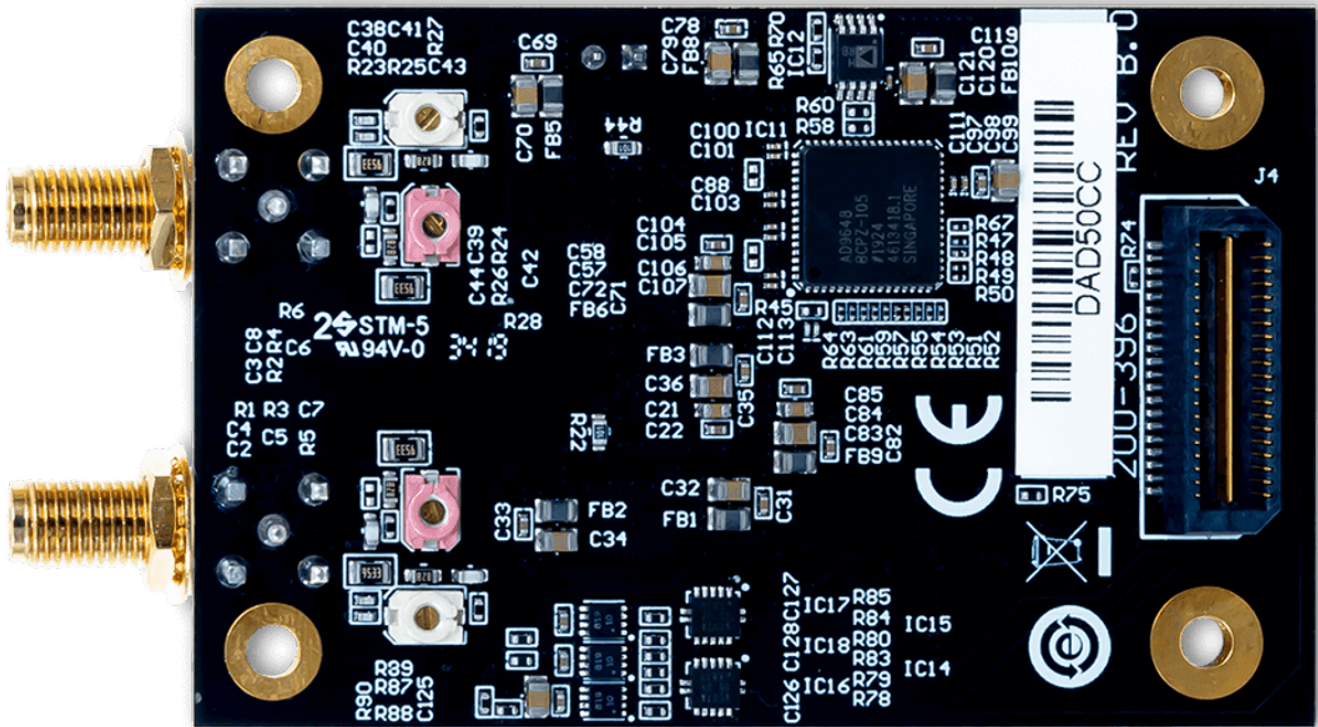


Feedback

([https://digilent.com/reference/\\_media/zmod/scope/zmodscope-generic-obl-1000.png](https://digilent.com/reference/_media/zmod/scope/zmodscope-generic-obl-1000.png))







The Zmod Scope was designed to be a piece in a modular, HW and SW open-source ecosystem. The Zmod Scope can acquire two signals, with simultaneous sampling. Combined with a SYZYGY™ carrier, other SYZYGY™ compatible pods, Zmod Scope can be used for a variety of applications: data acquisition systems, closed loop controllers, scopes, etc.

- There are multiple members in the Zmod Scope family:
- Zmod Scope 1410-40: 14-bit, 40MSPS

- Zmod Scope 1410-105: 14-bit, 105MSPS
- Zmod Scope 1410-125: 14-bit, 125MSPS
- Zmod Scope 1210-40: 12-bit, 40MSPS
- Zmod Scope 1210-125: 12-bit, 125MSPS
- Zmod Scope 1010-40: 10-bit, 40MSPS
- Zmod Scope 1010-125: 10-bit, 125MSPS

The Zmods in the family are similar, with small loading and performance differences, as explained in the current user manual. In this document, all the members of the family are generically referred as Zmod Scope and only aspects which differ between members are explicitly shown.

## Features

Table 1. Zmod Scope family features  $\square$

Features/Version	1410-40	1410-105	1410-125	1210-40	1210-125
ADC(s)	AD9251BCPZ-40	AD9648BCPZ-105	AD9648BCPZ-125	AD9231BCPZ-40	AD9648BCPZ-125
Input Channels	2				
Input range - Low Range	$\pm 1V$				
Input range - High Range	$\pm 25V$				
Resolution [bits]	14	14	14	12	12
Absolute resolution - Low Range	0.13mV	0.13mV	0.13mV	0.51mV	0.51mV
Absolute resolution - High Range	3.2mV	3.2mV	3.2mV	12.8mV	12.8mV
Accuracy [% of Input range]	$\pm 0.2\%$				
Sample rate - max [MSPS]	40 MSPS	105 MSPS	125 MSPS	40 MSPS	125 MSPS

Analog Bandwidth @ 3dB	20 MHz	70 MHz	70 MHz	20 MHz	70 MHz
Analog Bandwidth @ 0.5dB	8 MHz	30 MHz	30 MHz	8 MHz	30 MHz
Analog Bandwidth @ 0.1dB	4 MHz	20 MHz	20 MHz	4 MHz	20 MHz
Input resistance [MΩ]	1 MΩ/18 pF				

## 1. Architectural Overview and Block Diagram

This document describes the Zmod Scope's circuits, with the intent of providing a better understanding of its electrical functions, operations, and a more detailed description of the hardware's features and limitations. It is not intended to provide enough information to enable complete duplication of the Zmod Scope, but can help users to design custom configurations for programmable parts in the design.

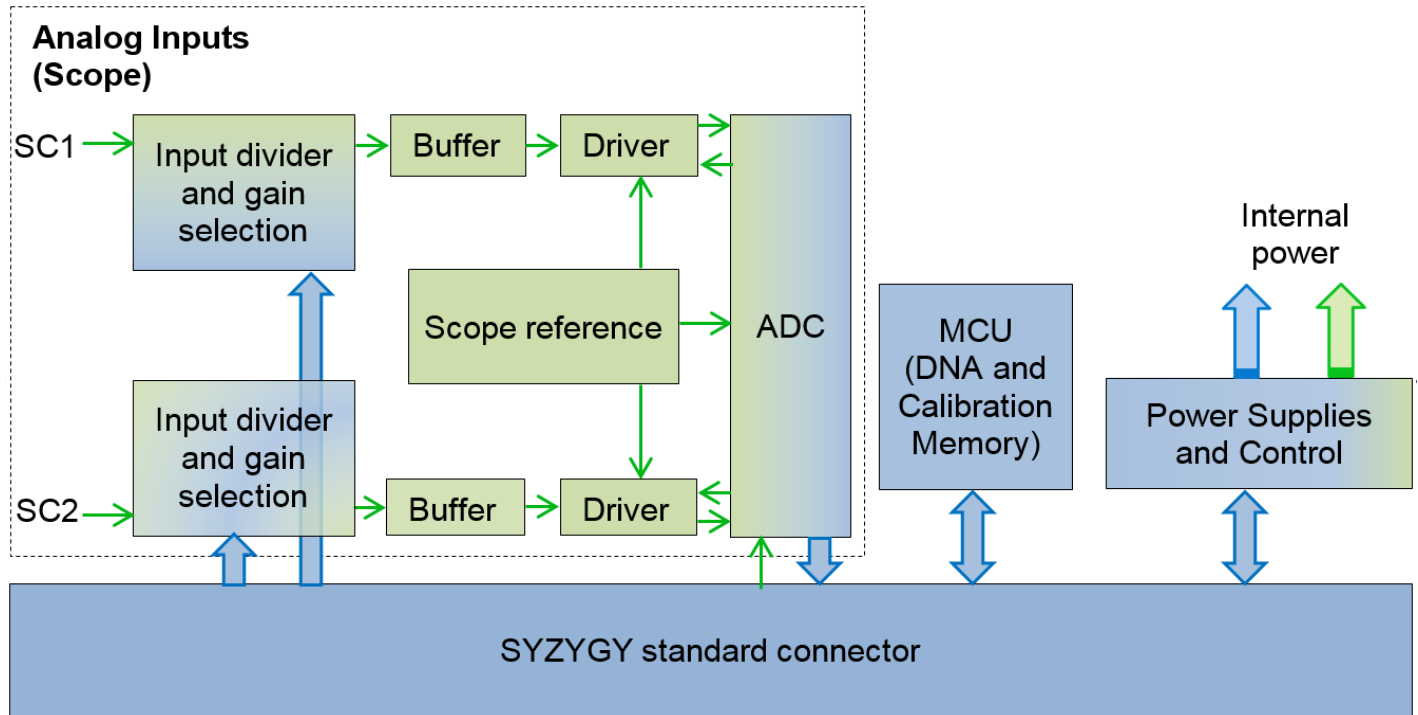
Zmod Scope's block diagram is presented in Fig. 2 below. The core of the Analog Zmod Scope is a dual channel, high speed, low power ADC, as shown in Table 1. The carrier board is responsible to configure the internal registers of the ADC circuit, provide the acquisition clock and receive the data.

The **Analog Input** block is also called the **Scope**, because of similar structure and behavior to such a front end. The signals in this circuitry use a "SC" indexes to indicate they are related to the scope block. Signals and equations also use certain naming conventions. Analog voltages are prefixed with a "V" (for voltage), and suffixes and indexes are used in various ways: to specify the location in the signal path (IN, MUX, BUF, ADC, etc.); to indicate the related instrument (SC, etc.); to indicate the channel (1 or 2); and to indicate the type of signal (P, N, or diff). Referring to the block diagram in Fig. 2 below:

- The **Analog Inputs/Scope** instrument block includes:
  - **Input Divider and Gain Selection:** high bandwidth input adapter/divider. High or low-gain can be selected by the FPGA
  - **Buffer:** high impedance buffer
  - **Driver:** provides appropriate signal levels and protection to the ADC.
  - **Scope Reference:** generates and buffers reference voltages for the scope stages
  - **ADC:** the analog-to-digital converter for both scope channels.
- The **Power Supplies and Control** block generates all internal supply voltages.
- The **MCU** works as a I2C memory for two different purposes:
  - The **DNA** includes the standard SYZYGY™ (<https://syzygyfpga.io>) pod identification information.
  - The **Calibration Memory** stores all calibration parameters. Except for the "Probe Calibration" trimmers in the scope Input divider, the Zmod Scope includes no analog calibration circuitry. Instead, a calibration operation is performed at manufacturing (or by

the user), and parameters are stored in memory. The application software uses these parameters to correct the acquired data and the generated signals

In the sections that follow, schematics are not shown separately for identical blocks. For example, the Scope Input Divider and Gain Selection schematic is only shown for channel 1 since the schematic for channel 2 is identical. Indexes are omitted where not relevant. As examples, in equation 1 below,  $V_{SCOPE-SMA}$  does not contain the channel index (because the equation applies to both channels 1 and 2).



([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/zmodadcblockdiagram.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/zmodadcblockdiagram.png?id=zmod%3Ascope%3Areference-manual))

Figure 2. Zmod Scope Block diagram. □

## 2. Scope

### 2.1. Scope Input Divider and Gain Selection

Fig. 3 shows the scope input divider and gain selection stage.

$C_5$  and  $C_6$  are capacitive trimmers, 3...10pF, -0/+50% tolerance. The worse case range is 4.5...10pF. All other capacitors are 1% tolerance, all the resistors are 0.1%.

The IC1 relay switches between two symmetrical R-C dividers. Each of them provide:

- Scope input impedance = 1MΩ || 18pF
- Two different attenuations for high-gain/low-gain (25:1)
- Controlled capacitance, much higher than the parasitical capacitance of subsequent stages
- Constant attenuation over a large frequency range (trimmer adjusted)

The maximum voltage rating for scope inputs is limited to:

$$-50V < V_{SCOPE-SMA} < 50V \quad (1)$$

The DC low gain is:

$$\frac{V_{SC-LG}}{V_{SCOPE-SMA}} = \frac{R_5}{R_1 + R_3 + R_5} = 0.04 \quad (2)$$

The High Range (at low gain):

$$-25V \leq V_{SCOPE-SMA} \leq 25V \quad (3)$$

The high gain is:

$$\frac{V_{SC-HG}}{V_{SCOPE-SMA}} = \frac{R_4 + R_6}{R_2 + R_4 + R_6} = 0.96 \quad (4)$$

The Low Range (at high gain):

$$-1V \leq V_{SCOPE-SMA} \leq 1V \quad (5)$$

The two dividers are designed to have the same equivalent impedance (both active and reactive):

$$R_{ech} = R_1 + R_3 + R_5 = R_2 + R_4 + R_6 = 1Mohm \quad (6)$$

Experiments shown that there are significant parasitic capacities of the layout and buffer input stage:  $C_{PH}$  (high gain divider), parallel to  $C_6$ , and  $C_{PL}$  (low gain divider), parallel to  $C_7$ . The trimmers should compensate for these parasitic capacities and adjust for perfect matching:

$$C_3 * R_2 = (C_{PH} + C_6) * (R_4 + R_6) \quad (7)$$

$$(C_{PH} + C_6) = \frac{C_3 * R_2}{R_4 + R_6} = 18pF \quad (8)$$

$$(C_4 + C_5) * (R_1 + R_3) = (C_{PL} + C_7) * R_5 \quad (9)$$

$$(C_{PL} + C_7) = (C_4 + C_5) * \frac{(R_1 + R_3)}{R_5} \quad (10)$$

With the chosen values, the correct adjustment results in about mid-position of trimmers  $C_5$  and  $C_6$ :

$$C_5 = C_6 = 7pF \quad (11)$$

which solves the parasitic capacities as:

$$C_{PH} = 11pF \quad (12)$$

$$C_{PL} = 8.8pF \quad (13)$$



The Low Gain and High Gain dividers have very close equivalent capacitance, within the tolerances and model approximations:

$$C_{HGech} = \frac{C_3 * R_2}{R_2 + R_4 + R_6} = C_{ech} = 17.28pF \quad (14)$$

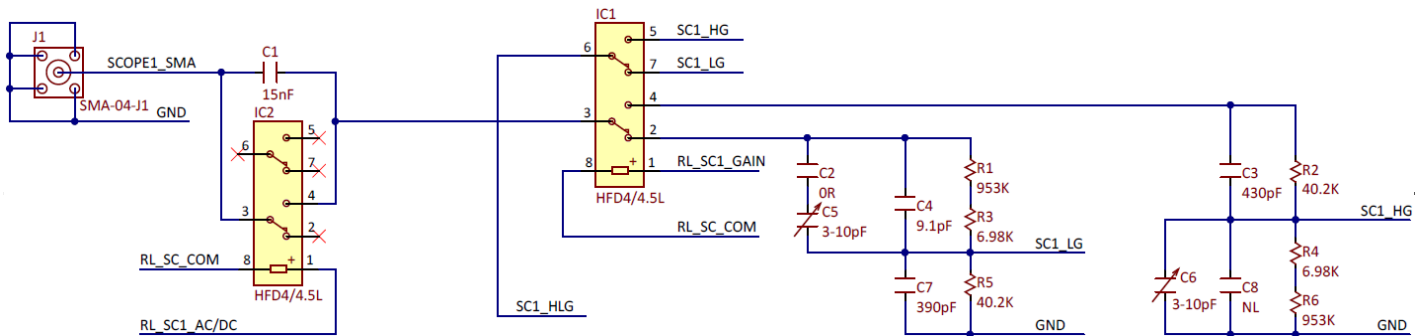
$$C_{LGech} = \frac{(C_7 + C_{PL}) * R_5}{R_1 + R_3 + R_5} = 16.03p \quad (15)$$

Experiments show that the equivalent capacitances are even closer than the values above, about 18pF. The computing error mainly derives from trimmer position approximation.

$$C_{ech} = 18p \quad (16)$$

The IC2 relay shorts the C1 capacitor when DC coupling is desired. Otherwise, C1 forms a High Pass filter with the selected divider, for AC coupling, with the corner frequency:

$$f_c = \frac{1}{2 * \pi * R_{ech} * (C_{ech} + C_1)} \approx \frac{1}{2 * \pi * R_{ech} * C_1} = 10.6Hz \quad (17)$$



([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/inputdivider.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/inputdivider.png?id=zmod%3Ascope%3Areference-manual))

Figure 3. Input divider and gain selection. □

IC1 and IC2 in Fig. 3, are HFD4/4.5L (<https://hongfa.com/product/detail/70689a00-9d73-4802-b0ac-e36561332f57>) latching relays (possible replacement: IM42GR

(<https://www.te.com/commerce/DocumentDelivery/DDEController?Action=srchtrrv&DocNm=108-98001&DocType=SS&DocLang=EN>). The schematic shows the “reset” position. A relay is “set” when a positive voltage is applied at the coil terminals and it is “reset” when a negative voltage is applied (see the polarity on the schematic symbol). The relay keeps state when no current flows through the coil (the terminals are driven at the same voltage or at least one of the terminals is “open”). The nominal coil voltage is 4.5V.

The IC16, IC17 and IC18 in Fig. 4 are A3910EEEETR-T

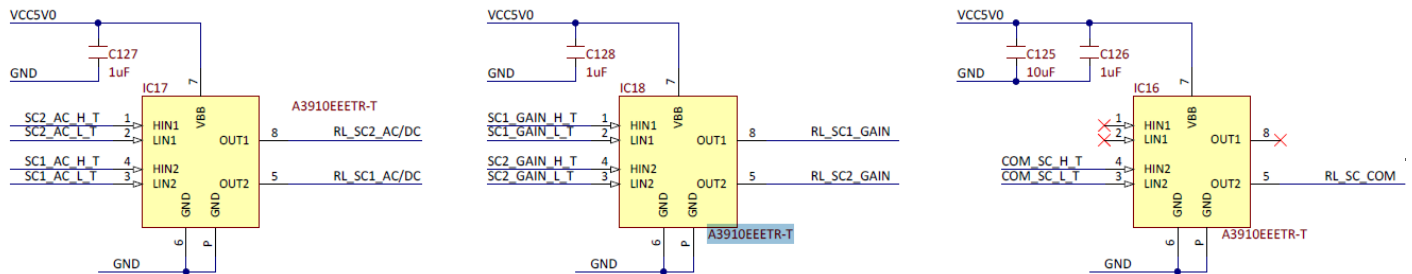
(<https://www.digikey.com/en/datasheets/allegromicrosystemslc/allegro-microsystems-lca3910datasheetshx>) drivers for the relays. They feature:

- Low RDS(on) outputs
- Standby mode with zero current drain




- Small  $2 \times 2$  DFN package
- Crossover Current protection
- Thermal Shutdown protection

Normally, all of them have both HIN and LIN inputs “low” or both “High” driving the “OUT” pins “High Z”. To “set” a relay, OUT2 of IC16 is set “Low” (HIN=“Low”, LIN=“High”) and the corresponding OUT pin of IC17 or IC18 is set “High” (HIN=“High”, LIN=“Low”). All other OUT pins are set “High Z”. To “reset” a relay, OUT2 of IC16 is set “High” (HIN=“High”, LIN=“Low”) and the corresponding OUT pin of IC17 or IC18 is set “Low” (HIN=“Low”, LIN=“High”). All other OUT pins are set “High Z”.

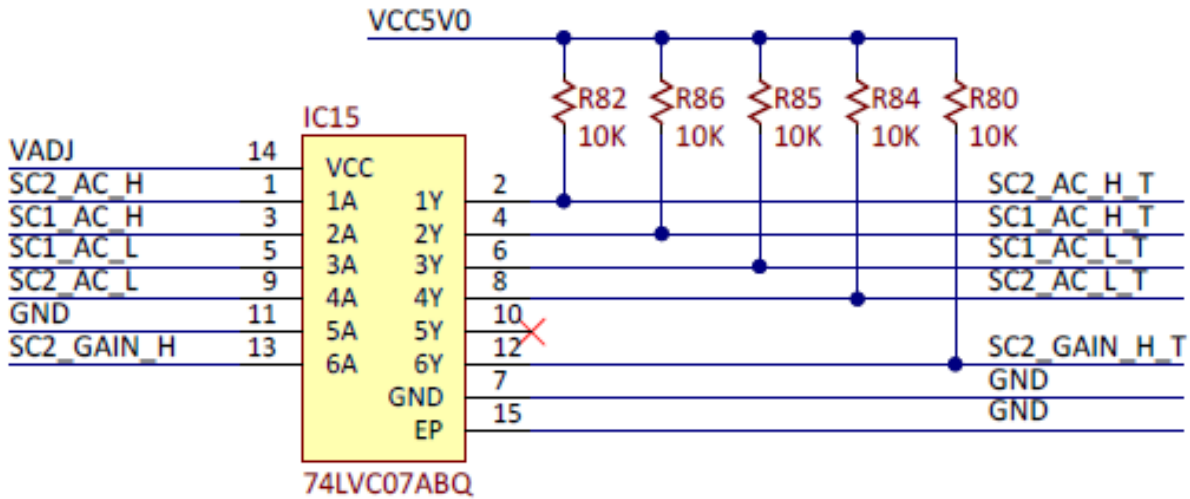
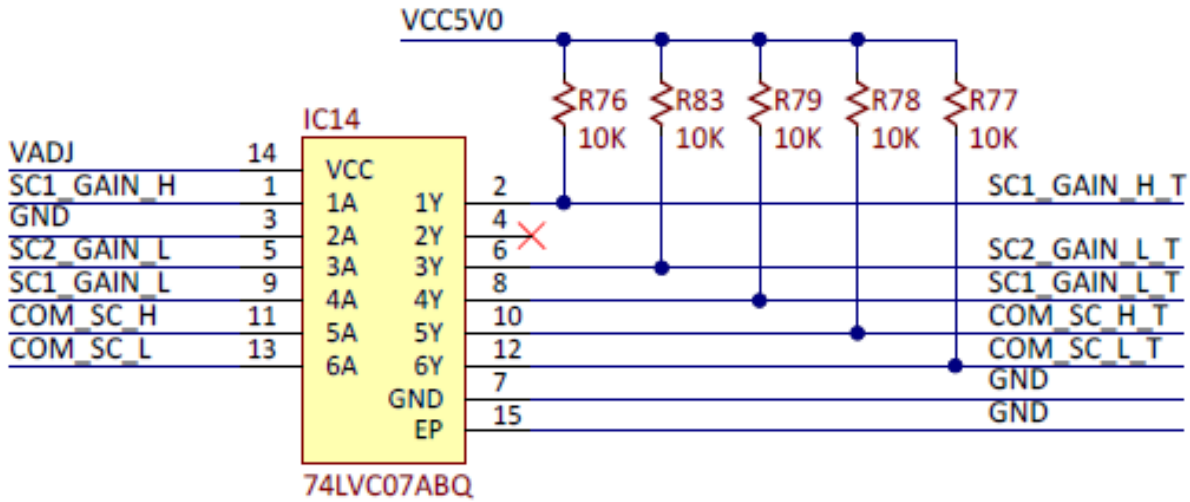


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Figure 4. Relay Drivers. □

IC14 and IC15 in Fig. 5 are  74LVC07ABQ (<https://www.nxp.com/products/analog-logic-ics/asynchronous-interface-logic/buffers-inverters-drivers/74LVC07ABQ.html>) open-drain gates used as level translators from  $V_{ADJ} = 1.8V$  to  $V_{CC5V0} = 5V$ .

- 5 V tolerant inputs and outputs (open-drain) for interfacing with 5 V logic
- Wide supply voltage range from 1.2 V to 5.5 V
- CMOS low power consumption
- Direct interface with TTL levels
- Inputs accept voltages up to 5 V
- Complies with JEDEC standard:
- JESD8-7A (1.65 V to 1.95 V)
- JESD8-5A (2.3 V to 2.7 V)
- JESD8-C/JESD36 (2.7 V to 3.6 V)
- ESD protection:
- HBM JESD22-A114F exceeds 2000 V
- MM JESD22-A115-B exceeds 200 V
- CDM JESD22-C101E exceeds 1000 V
- Specified from  $-40\text{ }^{\circ}\text{C}$  to  $+85\text{ }^{\circ}\text{C}$  and  $-40\text{ }^{\circ}\text{C}$  to  $+125\text{ }^{\circ}\text{C}$



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Figure 5. Level Translators. □

## 2.2. Scope Buffer

The scope buffer stage provides very high impedance as load for the input divider.

The gain is:

$$\frac{V_{BUFF}}{V_{SC-HLG}} = 1 \quad (20)$$

The actual input and output range (for nominal usage) is:

$$-1V < V_{SC-HLG} = V_{BUFF} < 1V \quad (21)$$

## 2.3. Scope Reference

The scope reference stage generates the 1V reference voltage for the  $\text{ADC}_0$ , as well as other internal reference voltages.

$$V_{REFADC} = 1V \quad (23)$$

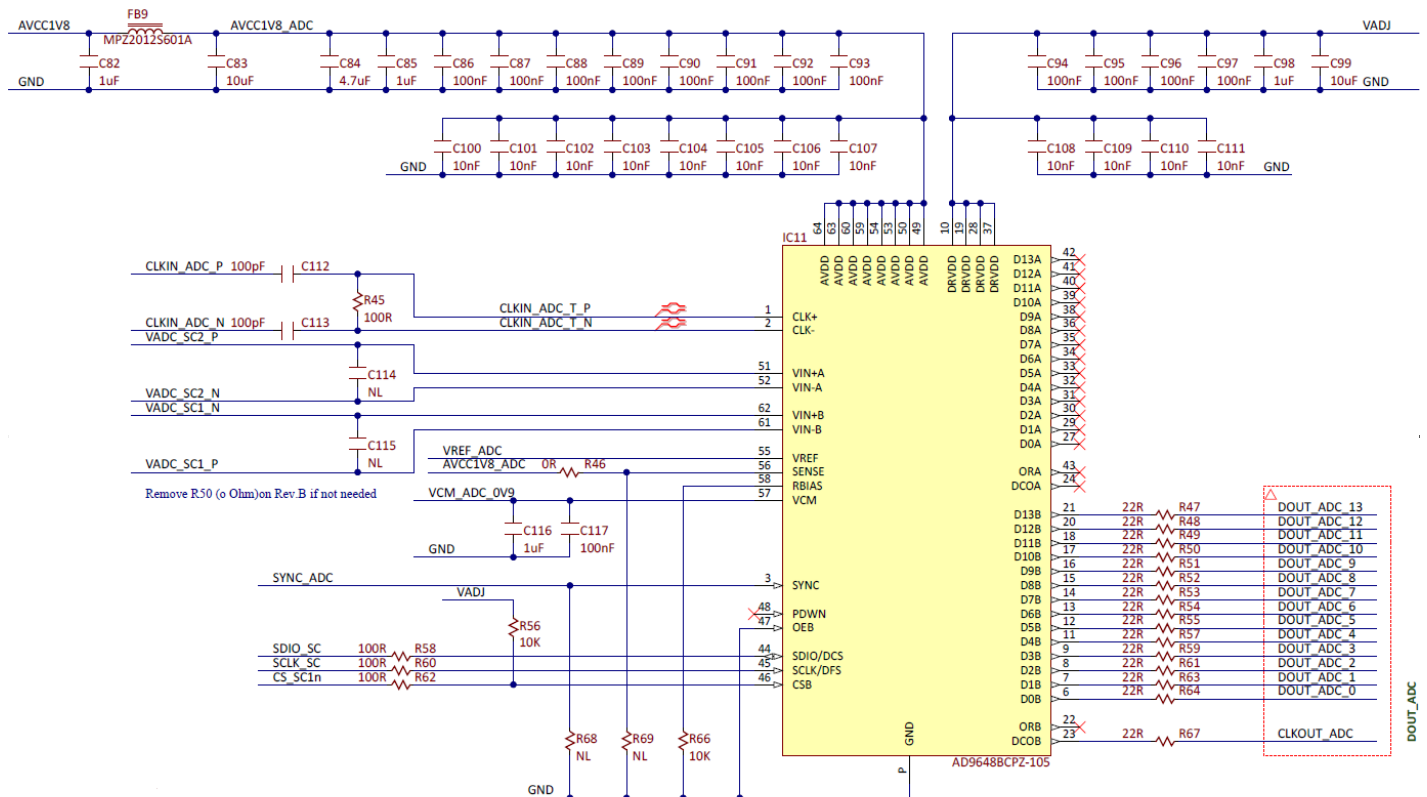
## 2.4. Scope Driver

The  $\text{ADC}_0$  driver is used for:

- Driving the differential inputs of the  $\text{ADC}_0$  (with low impedance outputs)
- Providing the common mode voltage for the  $\text{ADC}_0$
- $\text{ADC}_0$  protection.

## 2.5. Scope ADC

The Zmod Scope uses a dual channel, high speed, low power, 14-bit, 105MS/s  $\text{ADC}_0$ , as shown in Fig. 11.



[https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/adc.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/adc.png?id=zmod%3Ascope%3Areference-manual)

Figure 11. ADC (C114 and C115 are NoLoad for Zmod Scope XXXX-105 and Scope XXXX-125, respectively 39pF for Zmod Scope XXXX-40) []

The differential inputs are driven via a low-pass filter comprised of C114 together with R13, R15, R16, R17 in the buffer stage. The differential clock is AC-coupled and the line is impedance matched. The clock is internally divided by 4 to operate the `ADC_0`. The `ADC_0` generates the common mode reference voltage (VCM\_SC) to be used in the buffer stage.

The digital stage of the `ADC_0` and the corresponding FPGA bank are supplied at 1.8V by the SYZYGY™ voltage  $V_{adj}$ .

The multiplexed mode is used, to combine the two channels on a single data bus and minimize the number of used FPGA pins. CLKOUT\_SC is provided to the FPGA for synchronizing data.

## 2.6. Scope Signal Scaling

The nominal differential `ADC_0` input voltage range is:

$$-1V < V_{ADC\ diff} < 1V \quad (33)$$

The total scope gains (from the SMA connectors to the `ADC_0` inputs) are:

$$Low\ gain = \frac{V_{ADC\ diff}}{V_{SCOPE-SMA}} = 0.038 \quad (38)$$

$$High\ gain = \frac{V_{ADC\ diff}}{V_{SCOPE-SMA}} = 0.91 \quad (39)$$

Considering the `ADC_0` input voltage range shown in [33](#):

$$at\ low\ gain : -26.3V < V_{SCOPE-SMA} < 26.3V$$

$$at\ high\ gain : -1.1V < V_{SCOPE-SMA} < 1.1V \quad (40)$$

To cover component value tolerances and to allow software calibration, only the ranges below are specified.

$$at\ low\ gain : -25V < V_{SCOPE-SMA} < 25V \quad (41)$$

$$at\ high\ gain : -1V < V_{SCOPE-SMA} < 1V \quad (42)$$

With the 14-bit `ADC_0`, the absolute resolution of the scope is (see [Table 1](#)):

$$at\ low\ gain : \frac{52.6V}{2^n} \quad (43)$$

$$at\ high\ gain : \frac{2.12V}{2^n} n \quad (44)$$

For  $V_{in}$  voltage value at the input of the Scope channel, the Zmod Scope sends a signed n bit integer, N. This value is used to compute  $V_{in}$ :



$$V_{in} = \frac{N \cdot Range \cdot (1 + CG)}{2^{n-1}} + CA \quad (45)$$

where:

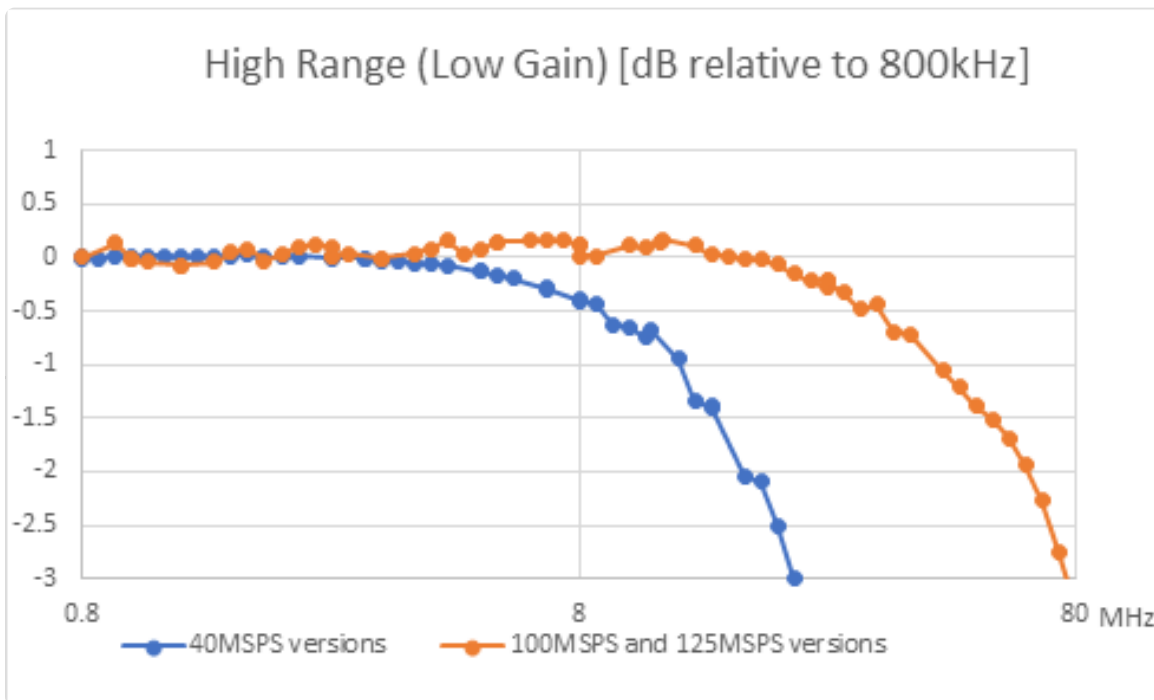
- n = the number of bits (14 for Zmod Scope 14XX, 12 for Zmod Scope 12XX, 10 for Zmod Scope 10XX)
- N = the n bit, 2's complement integer number returned by the `ADC()`
- $V_{in}$  = the corrected value of the input voltage
- CA = calibration Additive constant (for the appropriate channel and gain; see Table 4)
- CG = calibration Gain constant (for the appropriate channel and gain; see Table 4)
- Range = the ideal Range of the Scope input stage (approximation of the values in equation 40):
  - 1.086 (for low range:  $\pm 1V$ ) or
  - 26.25 (for high range:  $\pm 25V$ )

## 2.7 Scope Spectral Characteristics

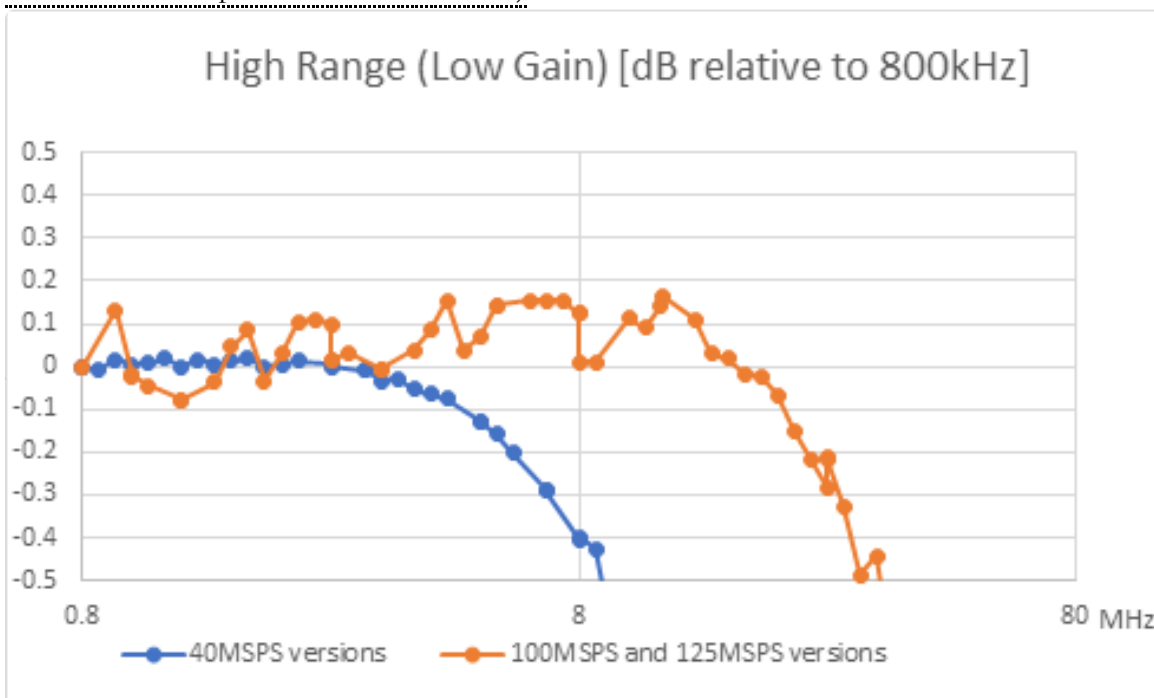
Fig. 12 shows a typical spectral characteristic of the scope input stage. A PXIe-5433 80 MHz Function/Arbitrary Waveform Generator was used to generate the input signal of 0.9V, for High Gain Scale, respectively 10V for the Low Gain scale. A Tektronix DPO5204B scope was used for measuring the reference signal (at the scope SMA connector) and the output signal (at the input of the `ADC()`). A differential probe was used to read the output signal on the pads of the unloaded C115. The signal swept from 800kHz to 80MHz. The effective values of the input and output signals were recorded for each frequency. The measurements were further processed to display the input stage frequency characteristics, as shown in Fig. 10.

For Zmod Scope XXXX-100MHz and Zmod Scope XXXX-125MHz, for both scales, the 3dB bandwidth is 70MHz+. The 0.5dB bandwidth is 30MHz and the 0.1dB bandwidth is 20MHz. For Zmod Scope XXXX-40MHz, for both scales, the 3dB bandwidth is 20MHz+. The 0.5dB bandwidth is 8MHz and the 0.1dB bandwidth is 4MHz

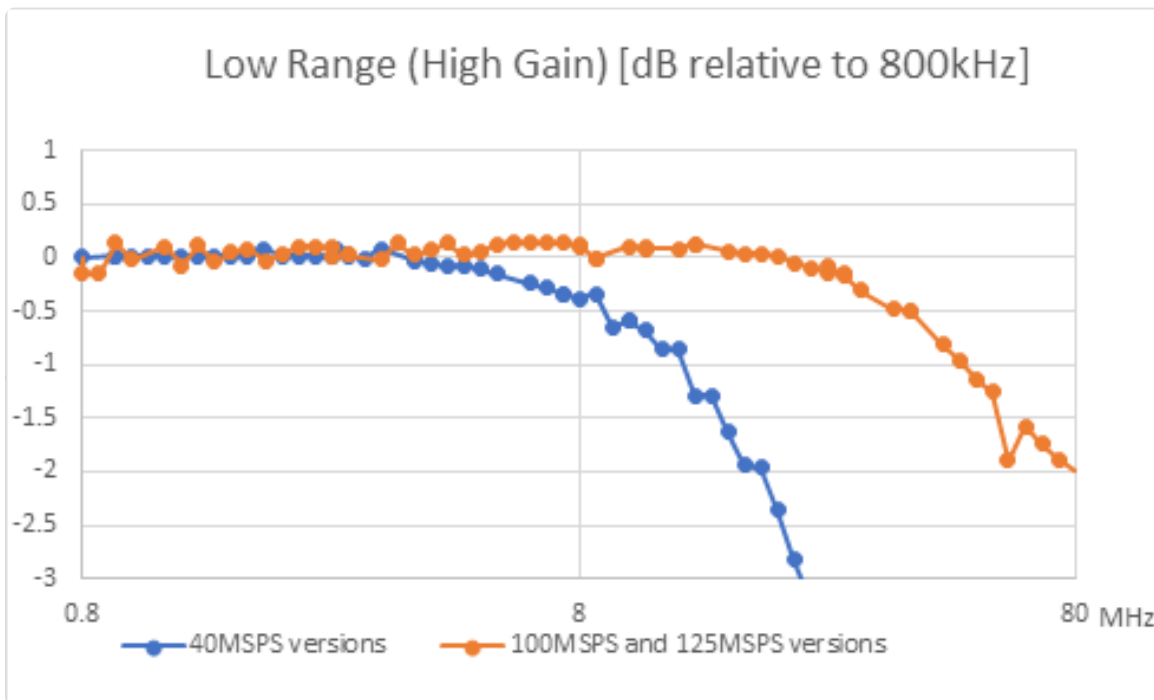
The standard -3dB bandwidth definition is derived from filter theory. At cutout frequency, the scope attenuates the spectral components by 0.707, assuming an error of  $\sim 30\%$ , way too high for a measuring instrument. The bandwidth with a specified flatness is useful to better define the scope spectral performances. The bandwidth @ 0.5dB, means a flatness error of a max 5.6%, while bandwidth @ 0.1dB means flatness error of a max 1.1%.



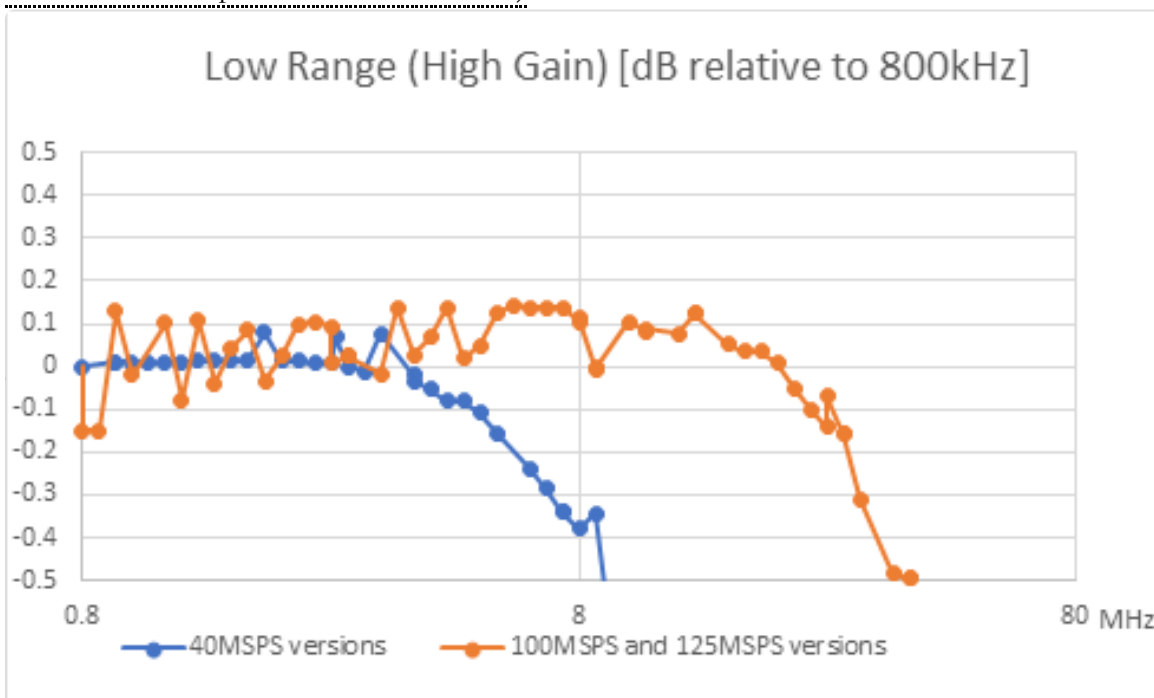
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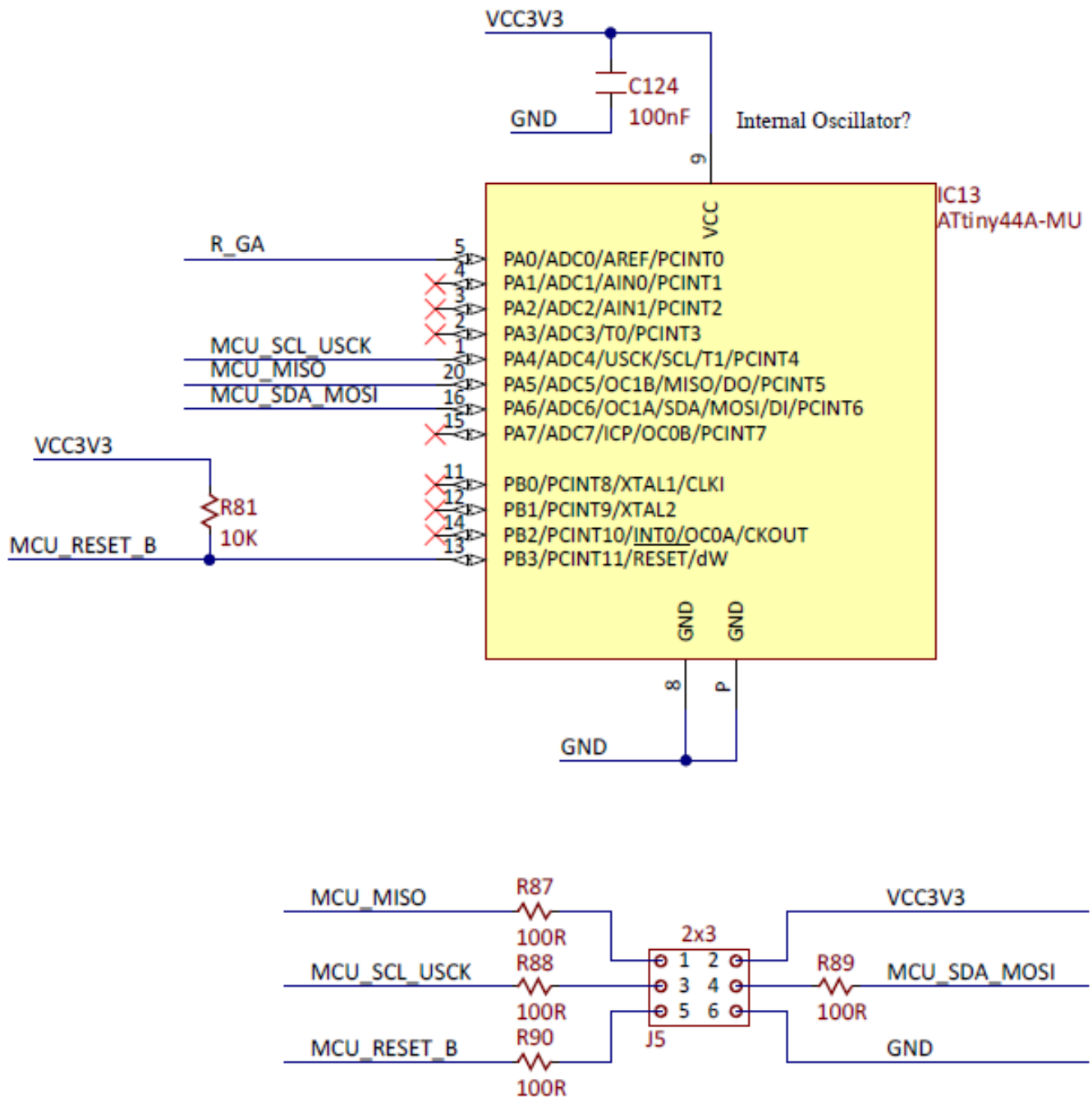
([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/bwdclrhg05dballversions.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/bwdclrhg05dballversions.png?id=zmod%3Ascope%3Areference-manual))

Figure 12. Scope input stage Bandwidth - Low Gain scale (up) and High Gain scale (down). 0.5dB detail (right) □

### 3. MCU

The  ATtiny44 (<https://www.microchip.com/wwwproducts/en/ATTINY44A>) MCU in Fig. 13 works as a I2C memory, storing the SYZYGY™ DNA information and the Calibration Coefficients. The J5 connector is used for programming the MCU and the SYZYGY™ DNA at manufacturing.

The DNA and the Factory Calibration Coefficients are stored in the Flash memory of the MCU, which appears to the I2C interface as “read-only”. The User Calibration Coefficients are stored in the `EEPROM()` memory of the MCU, which is write-protected via a magic number at a magic address. The memory structure can be consulted below.



([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/mcu.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/mcu.png?id=zmod%3Ascope%3Areference-manual))

Figure 13. The MCU

- Program Memory Type: Flash
- Program Memory Size (KB): 4
- CPU Speed (MIPS/DMIPS): 20
- SRAM Bytes: 256
- Data `EEPROM()`/HEF (bytes): 256



- Digital Communication Peripherals: 1-SPI, 1-I2C
- Capture/Compare/PWM Peripherals: 1 Input Capture, 1 CCP, 4PWM
- Timers: 1 x 8-bit, 1 x 16-bit
- Number of Comparators: 1
- Temperature Range (C): -40 to 85
- Operating Voltage Range (V): 1.8 to 5.5
- Pin Count: 14
- Low Power: Yes

Table 2. The Flash memory structure []

Address	Function	Size (Bytes)
0x8000 - 0x80FF	DNA	256
0x8100 - 0x817F	Factory Calibration	128
0x8180 - 0x83FF	Future use	896

### 3.1. SYZYGY™ DNA


The Zmod Scope is compliant with  SYZYGY™ Specification (<https://syzygyfpga.io/specification/>). It contains an MCU able to calculate the Geographical Address and provide the DNA information via I2C. The DNA is stored in the MCU FLASH at the address range: 0x8000 - 0x80FF with the following structure:

Table 3. The Zmod Scope DNA structure []

Contents	Type	Size(Bytes)	Value	Address
DNA full data length	uint16	2	91	0x8000
DNA header length	uint16	2	40	0x8002
SYZYGY DNA major version	uint8	1	1	0x8004
SYZYGY DNA minor version	uint8	1	0	0x8005
Required SYZYGY DNA major version	uint8	1	1	0x8006
Required SYZYGY DNA minor version	uint8	1	0	0x8007
Maximum operating 5V load (mA)	uint16	2	400	0x8008

<b>Contents</b>	<b>Type</b>	<b>Size(Bytes)</b>	<b>Value</b>	<b>Address</b>
Maximum operating 3.3V load (mA)	uint16	2	100	0x800A
Maximum VIO load (mA)	uint16	2	270	0x800C
Attribute flags	uint16	2	0	0x800E
Minimum operating VIO (10 mV steps)	uint16	2	180	0x8010
Maximum operating VIO (10 mV steps)	uint16	2	180	0x8012
Minimum operating VIO (10 mV steps)	uint16	2	170	0x8014
Maximum operating VIO (10 mV steps)	uint16	2	190	0x8016
Minimum operating VIO (10 mV steps)	uint16	2	0	0x8018
Maximum operating VIO (10 mV steps)	uint16	2	0	0x801A
Minimum operating VIO (10 mV steps)	uint16	2	0	0x801C
Maximum operating VIO (10 mV steps)	uint16	2	0	0x801E
Manufacturer name length	uint8	1	12	0x8020
Product name length	uint8	1	a)	0x8021
Product model / Part number length	uint8	1	a)	0x8022
Product version / revision length	uint8	1	1	0x8023
Serial number length	uint8	1	12	0x8024
RESERVED	uint8	1	0	0x8025
CRC-16 (most significant byte)	uint8	1	b)	0x8026
CRC-16 (least significant byte)	uint8	1	c)	0x8027
END DATA HEADER				

Contents	Type	Size(Bytes)	Value	Address
Manufacturer name	string	12	Digilent Inc	0x8028
Product name	string	a)	e)	d)
Product model / Part number	string	a)	e)	d)
Product version / revision	string	1	f)	d)
Serial number	string	12	210396000000	d)
Product ID	uint32	4	g)	0x80FC

a)	13	for Zmod Scope 1410
	17	for Zmod Scope 1410-105
	17	for Zmod Scope XXXX-125
	16	for Zmod Scope XXXX-40
b)	CRC computed over the addresses 0x8000-0x8025: most significant byte	
c)	CRC computed over the addresses 0x8000-0x8025: least significant byte	
d)	the subsequent address (hexadecimal)	
e)	Zmod Scope 1410, Zmod Scope 1410-105, Zmod Scope 1410-125, Zmod Scope 1210-125, Zmod Scope 1010-125, Zmod Scope 1410-40, Zmod Scope 1210-40, Zmod Scope 1010-40 - upon case	
f)	A, B, C, etc - upon case	
g)	0x80100200 Zmod Scope 1410-105	
	0x80101200 Zmod Scope 1010-40	
	0x80102200 Zmod Scope 1210-40	
	0x80103200 Zmod Scope 1410-40	

	0x80104200 Zmod Scope 1010-125	
	0x80105200 Zmod Scope 1210-125	
	0x80106200 Zmod Scope 1410-125	

## 3.2. Calibration Memory

The analog circuitry described in previous chapters includes passive and active electronic components. The datasheet specs show parameters (resistance, capacitance, offsets, bias currents, etc.) as typical values and tolerances. The equations in previous chapters consider typical values. Component tolerances affect DC and AC performances of the Zmod Scope. To minimize these effects, the design uses:

- 0.1% resistors and 1% capacitors in all the critical analog signal paths
- Capacitive trimmers for balancing the Scope Input Divider and Gain Selection
- No other mechanical trimmers (as these are big, expensive, unreliable and affected by vibrations, aging, and temperature drifts)
- Software calibration, at manufacturing
- User software calibration, as an option

A software calibration is performed on each device as a part of the manufacturing test. Reference signals are connected to the Scope inputs. A set of measurements is used to identify all the DC errors (Gain, Offset) of each analog stage. Correction (Calibration) parameters are computed and stored in the Calibration Memory, on the Zmod Scope device, both as Factory Calibration Data and User Calibration Data. The WaveForms software allows the user performing an in-house calibration and overwrite the User Calibration Data. Returning to Factory Calibration is always possible.

The Software reads the calibration parameters from the Zmod Scope MCU via the I2C bus and uses them to correct the acquired signals. The structure of the calibration data is shown below:

Table 4. The Calibration Data Structure □

Heading 1	Name	Size (Bytes)	Type	Flash Address (Factory Calibration)	<u>EEPROM()</u> Address (User Calibration)
Magic ID		1	uchar 0xAD	0x8100	0x7000
Calibration Time		4	unix timestamp	0x8101	0x7001
Channel 1 LG Gain	CG	4	float32	0x8105	0x7005



<b>Heading 1</b>	<b>Name</b>	<b>Size (Bytes)</b>	<b>Type</b>	<b>Flash Address (Factory Calibration)</b>	<b><u>EEPROM()</u> Address (User Calibration)</b>
Channel 1 LG Offset	CA	4	float32	0x8109	0x7009
Channel 1 HG Gain	CG	4	float32	0x810D	0x700D
Channel 1 HG Offset	CA	4	float32	0x8111	0x7011
Channel 2 LG Gain	CG	4	float32	0x8115	0x7015
Channel 2 LG Offset	CA	4	float32	0x8119	0x7019
Channel 2 HG Gain	CG	4	float32	0x811D	0x701D
Channel 2 HG Offset	CA	4	float32	0x8121	0x7021
Reserved Area		68	-	0x8125	0x7025
Log		22	string	0x8169	0x7069
CRC		1	uchar	0x817F	0x707F

Table 5. The EEPROM Memory Map []

<b>Address</b>	<b>Function</b>	<b>Size (Bytes)</b>
0x7000 - 0x707F	User Calibration	128
0x7080 - 0x70FF	Future Use	128

At the power up the EEPROM() memory is protected against write operations. To disable the write protection one has to write a magic number to a magic address over I2C. To re-enable the write protection one has to write any other number to the magic address.

Table 6. The Write Protection Disable magic number and address []

Magic Number	Magic Address
0xD2	0x6FFF

## 4. Power Supplies and Control

This block includes the internal power supplies.


The Zmod Scope gets the digital rails from the carrier board, via the SYZYGY connector:

- VCC5V0 - used for relays and analog supplies
- VCC3V3 - used for the MCU and analog supplies
- Vadj = 1.8V - used for the ADC() digital rail

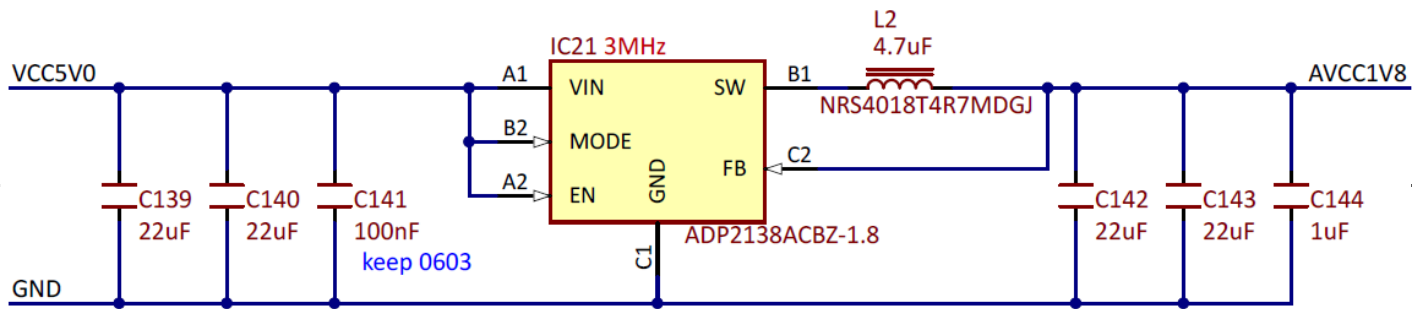
The internal analog rails sequence is:

- AVCC1V8 - ADC() analog rail
- AVCC3V0 - ADC() driver
- AVCC-2V5, AVCC4V5 - Scope buffer, reference voltage

### 4.1. AVCC1V8

The analog supply AVCC1V8 is built from VCC5V0 using IC21, an  ADP2138 (<http://www.analog.com/en/power-management/switching-regulators-integrated-fet-switches/adp2138/products/product.html>) Fixed Output Voltage, 800mA, 3MHz, Step-Down DC-to-DC converter. To insure low output voltage ripple a second LC filter (FB9 in Fig. 9) is added and forced PWM mode is selected.


- Input voltage: 2.3 V to 5.5 V
- Peak efficiency: 95%
- 3 MHz() fixed frequency operation
- Typical quiescent current: 24  $\mu$ A
- Very small solution size
- 6-lead, 1 mm  $\times$  1.5 mm WLCSP package
- Fast load and line transient response
- 100% duty cycle low dropout mode
- Internal synchronous rectifier, compensation, and soft start
- Current overload and thermal shutdown protections
- Ultra-low shutdown current: 0.2  $\mu$ A (typical)
- Forced PWM and automatic PWM/PSM modes



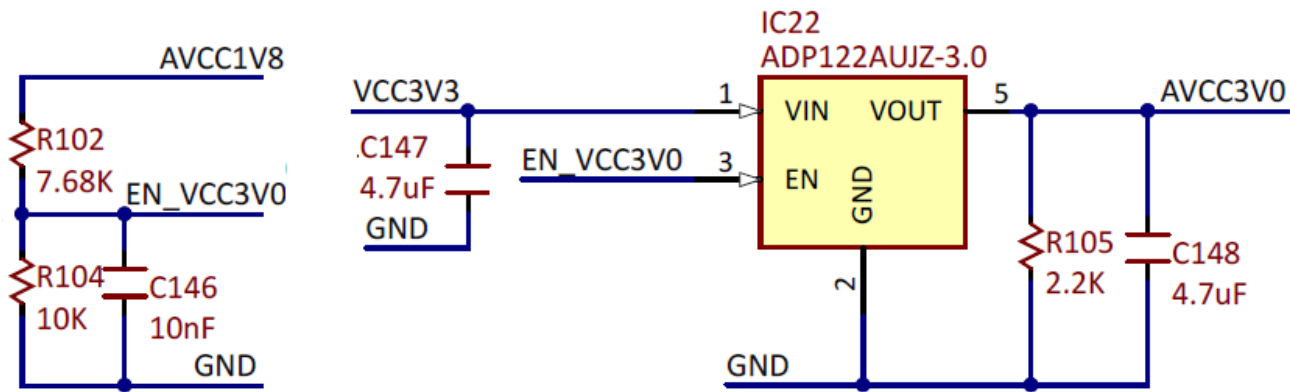
([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/avcc1v8.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/avcc1v8.png?id=zmod%3Ascope%3Areference-manual))

Figure 14. AVCC1V8 □

## 4.2. AVCC3V0

The analog supply AVCC3V0 is built from VCC3V3 using IC22, an  ADP122 (<https://www.analog.com/en/products/adp122.html>) 5.5 V Input, 300 mA, Low Quiescent Current, CMOS Linear Regulator, Fixed Output Voltage. To reduce noise and reduce the crosstalk between supplied circuits, the rail uses individual LC filters: FB3 in Fig. 10, FB6 (Channel 2 ADC<sub>0</sub>) Driver - not shown), FB7 in Fig. 7.

- Input voltage supply range: 2.3 V to 5.5 V
- 300 mA maximum output current
- Fixed and adjustable output voltage versions
- Very low dropout voltage: 85 mV at 300 mA load
- Low quiescent current: 45  $\mu$ A at no load
- Low shutdown current: <1  $\mu$ A
- Initial accuracy:  $\pm$ 1% accuracy
- Up to 31 fixed-output voltage options available from
- 1.75 V to 3.3 V
- Adjustable-output voltage range
- 0.8 V to 5.0 V (ADP123)
- Excellent PSRR performance: 60 dB at 100 kHz ( )
- Excellent load/line transient response
- Optimized for small 1.0  $\mu$ F ceramic capacitors
- Current limit and thermal overload protection
- Logic controlled enable
- Compact packages: 5-lead TSOT and 6-lead 2 mm  $\times$  2 mm LFCSP



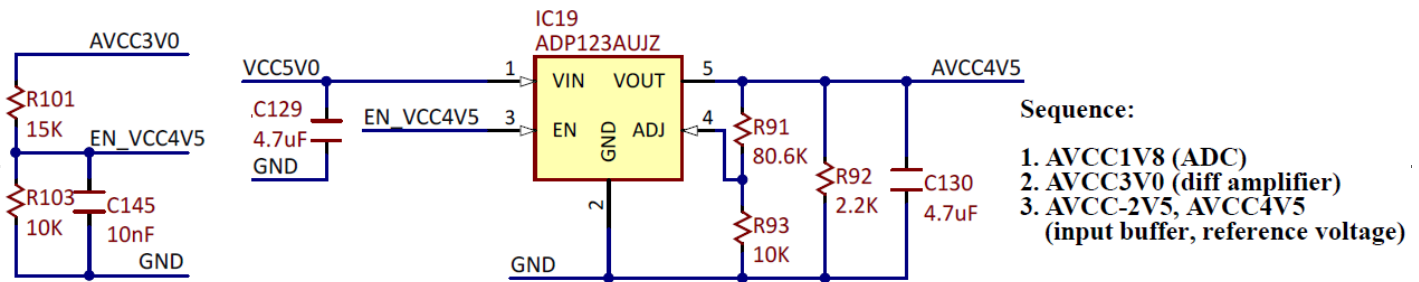
([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/avcc3v0.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/avcc3v0.png?id=zmod%3Ascope%3Areference-manual))

Figure 15. AVCC3V0 □

### 4.3. AVCC4V5

The analog supply AVCC4V5 is built from VCC5V0 using IC19, an [ADP123](https://www.analog.com/en/products/adp123.html) 5.5 V Input, 300 mA, Low Quiescent Current, CMOS Linear Regulator, Adjustable Output Voltage. To reduce noise and reduce the crosstalk between supplied circuits, the rail uses individual LC filters: FB2 in Fig. 6, FB5 (Channel 2 ADC<sub>0</sub> Buffer - not shown), FB8 in Fig. 9.

- Input voltage supply range: 2.3 V to 5.5 V
- 300 mA maximum output current
- Fixed and adjustable output voltage versions
- Very low dropout voltage: 85 mV at 300 mA load
- Low quiescent current: 45  $\mu$ A at no load
- Low shutdown current: <1  $\mu$ A
- Initial accuracy:  $\pm$ 1% accuracy
- Up to 31 fixed-output voltage options available from
- 1.75 V to 3.3 V
- Adjustable-output voltage range
- 0.8 V to 5.0 V (ADP123)
- Excellent PSRR performance: 60 dB at 100 kHz ()
- Excellent load/line transient response
- Optimized for small 1.0  $\mu$ F ceramic capacitors
- Current limit and thermal overload protection
- Logic controlled enable
- Compact packages: 5-lead TSOT and 6-lead 2 mm  $\times$  2 mm LFCSP



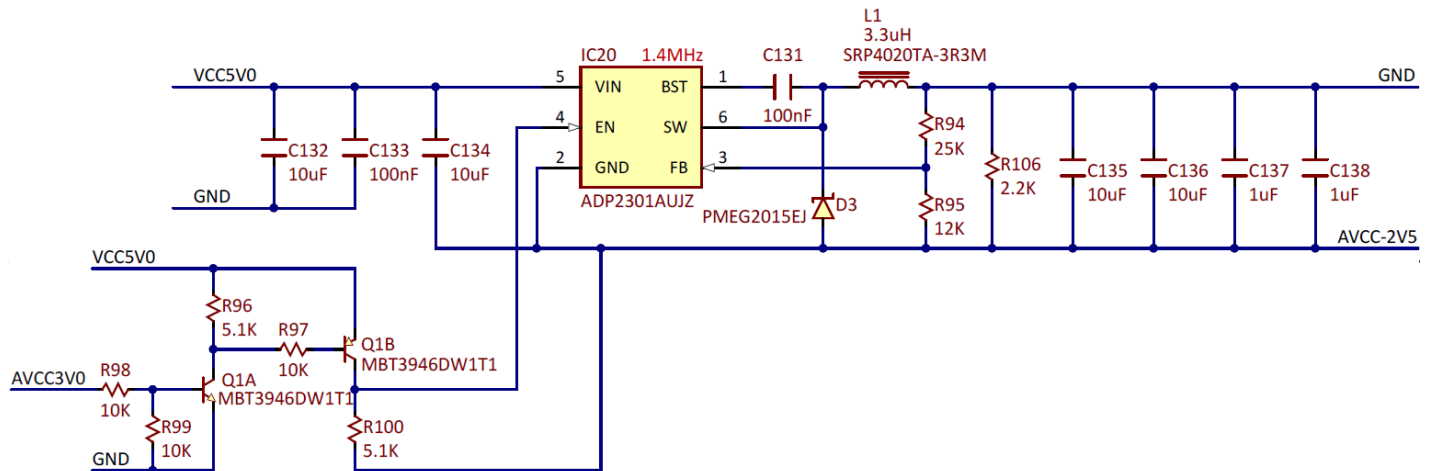
([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/avcc4v5.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/avcc4v5.png?id=zmod%3Ascope%3Areference-manual))

Figure 16. AVCC4V5

## 4.4. AVCC-2V5

The AVCC-2V5 analog power supply is implemented with the [ADP2301](http://www.analog.com/en/power-management/switching-regulators-integrated-fet-switches/adp2301/products/product.html) (http://www.analog.com/en/power-management/switching-regulators-integrated-fet-switches/adp2301/products/product.html) Step-Down regulator in an inverting Buck-Boost configuration. See application Note [AN-1083: Designing an Inverting Buck Boost Using the ADP2300 and ADP2301](http://www.analog.com/static/imported-files/application_notes/AN-1083.pdf) (http://www.analog.com/static/imported-files/application\_notes/AN-1083.pdf). To reduce noise and reduce the crosstalk between supplied circuits, the rail uses individual LC filters: FB1 in Fig. 6, FB4 (Channel 2 [ADC](#)) Buffer - not shown). The ADP2301 features:

- 1.2 A maximum load current
- $\pm 2\%$  output accuracy over temperature range
- 1.4 MHz switching frequency
- High efficiency up to 91%
- Current-mode control architecture
- Output voltage from 0.8 V to  $0.85 \times V_{IN}$
- Automatic PFM/PWM mode switching
- Integrated high-side MOSFET and bootstrap diode,
- Internal compensation and soft start
- Undervoltage lockout (UVLO), Overcurrent protection (OCP) and thermal shutdown (TSD)
- Available in ultrasmall, 6-lead TSOT package

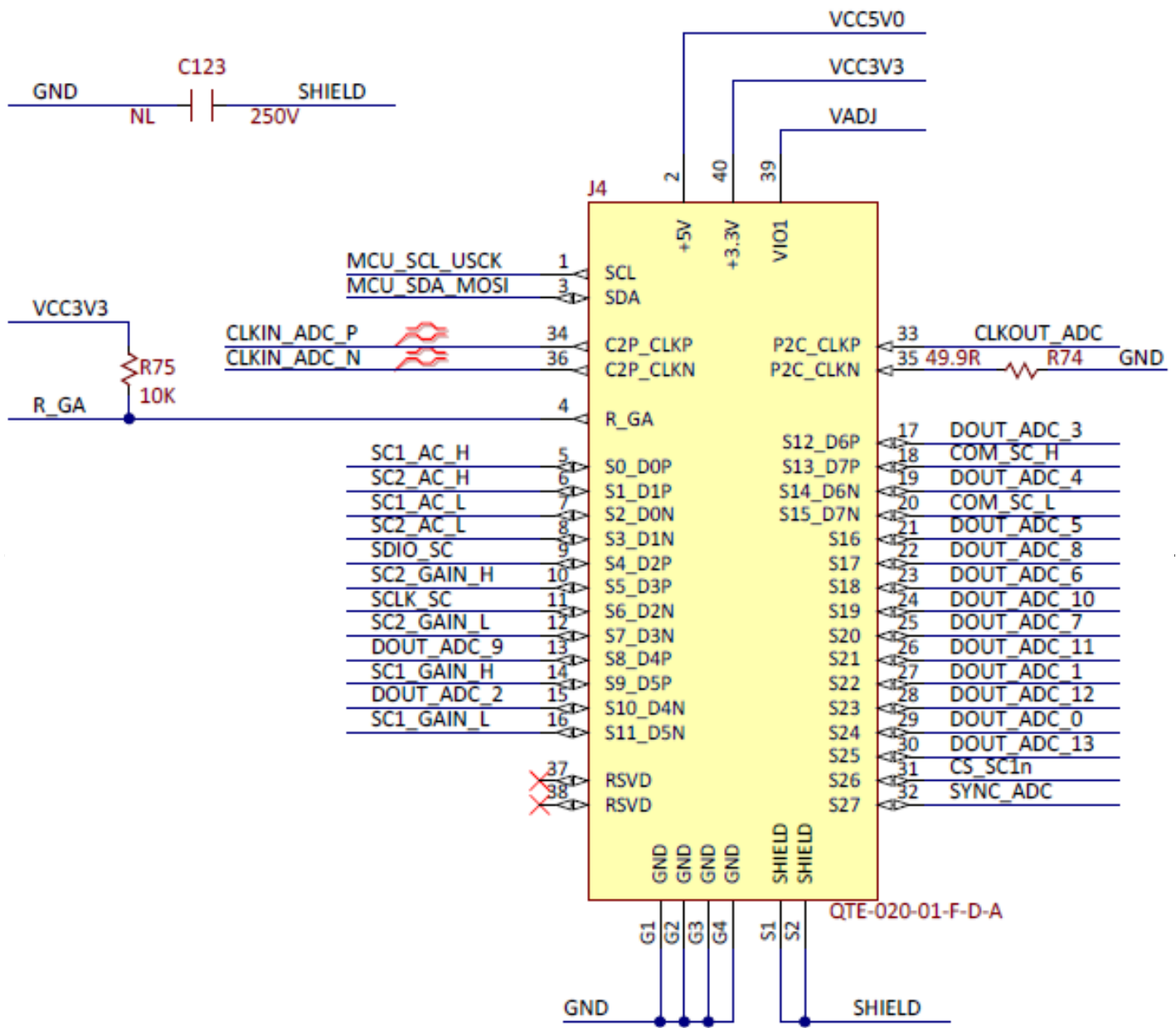


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## 5. The SYZYGY™ Connector

The SYZYGY™ connector provides the interface with the carrier board. The used signals are:

- Power rails
  - VCC5V0
  - VCC3V3
  - VADJ - needs to be set by the carrier board to 1.8V
  - GND\_()
  - Shield
- SYZYGY™ I2C bus:
  - MCU\_SCLUSCK
  - MCU\_SDA\_MOSI\_()
- ADC\_() differential input clock
  - CLKIN\_ADC\_()\_P
  - CLKIN\_ADC\_()\_N
- ADC\_() single ended output clock:
  - CLKOUT\_ADC\_() (coupled with GND\_() in the differential P2C pair)
- R\_GA for geographical address identification
- SYNC\_ADC\_() for ADC\_() internal clock divider synchronization
- ADC\_() data bus: DOUT\_ADC\_()\_0...13
- ADC\_() SPI bus:
  - CS\_()\_SC1n
  - SCLK\_()\_SC
  - SDIO\_SC
- relay control
  - SCx\_yy\_z



([https://digilent.com/reference/\\_detail/reference/test-and-measurement/zmodadc/syzygyconn.png?id=zmod%3Ascope%3Areference-manual](https://digilent.com/reference/_detail/reference/test-and-measurement/zmodadc/syzygyconn.png?id=zmod%3Ascope%3Areference-manual))

Figure 18. SYZYGY™ connector []

## 6. The SYZYGY™ compatibility table

Table 7. The SYZYGY™ compatibility table []

Parameter	Value
Maximum 5V supply current	400mA
Maximum 3.3V supply current	100mA



Parameter	Value
VIO supply voltage	1.8V
Maximum VIO supply current	270mA
Total number of I/O	28
Number of differential I/O pairs	0
Width	Single

Written by Mircea Dabacan, PhD, Technical University of Cluj-Napoca Romania

1

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