# 220 mA dual output LCD bias power supply

Rev. 2 — 20 December 2023

**Product data sheet** 

# 1. General description

The NEX10001 is a power supply IC providing programmable positive and negative output voltages for TFT-LCD panels. The device features an integrated boost converter, LDO regulator and a negative charge pump. The device requires only a single inductor allowing for a small Bill-Of-Material and the smallest PCB area.

The NEX10001 has an input voltage range of 2.7 V to 5 V and can operate from single-cell Li-ion, Ni-Li and Li-Polymer batteries It can supply up to 220 mA output current. The NEX10001 has excellent line transient and load transient responses.

## 2. Features and benefits

- Input voltage range: 2.7 V to 5 V
- Programmable output voltages:
  - OUTP: +4 V to +6.5 V (0.1 V step)
  - OUTN: -4 V to -6.5 V (0.1 V step)
- Maximum output current: 220 mA
- · 85% high efficiency at 80 mA combined output
- ±1% output voltage accuracy
- · Excellent transient response
- I<sup>2</sup>C Interface:
  - Flexible output voltage programming
- · Integrated active output discharge function
- UVLO and OTSD protection
- Wafer level chip-scale package; 15 bumps;
  1.16 × 1.96 × 0.62 mm body (0.4 mm pitch)
- ESD protection:
  - HBM: ANSI/ESDA/JEDEC JS-001 class 2 exceeds 2000 V
  - CDM: ANSI/ESDA/JEDEC JS-002 class C2a exceeds 500 V
- Specified from -40 °C to +85 °C

# 3. Applications

- Smart phone TFT-LCD
- Tablet TFT-LCD

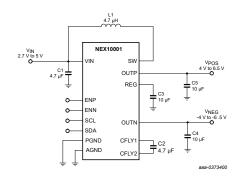
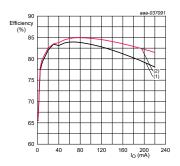


Fig. 1. Typical application



$$V_{POS} = 5.4 \text{ V}; V_{NEG} = -5.4 \text{ V}$$

$$(1) V_{IN} = 3.7 V$$

(2) 
$$V_{IN} = 4.5 V$$

Fig. 2. Efficiency versus output current (I<sub>O</sub>)



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# 4. Ordering information

**Table 1. Ordering information** 

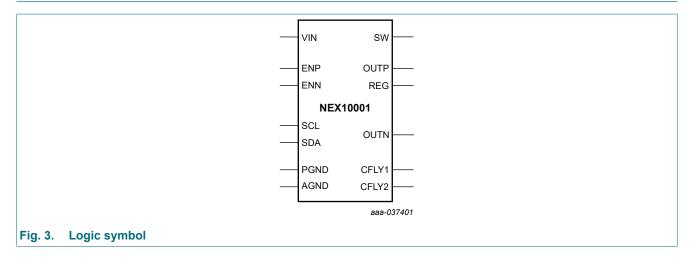
Type number	Package						
	Temperature range	Name	Description	Version			
NEX10001UB	-40 °C to +85 °C	WLCSP15	wafer level chip-scale package; 15 bumps; 1.16 × 1.96 × 0.62 mm body	SOT8054-1			

# 5. Marking

Table 2. Marking codes

Type number	Marking code
NEX10001UB	NX01A

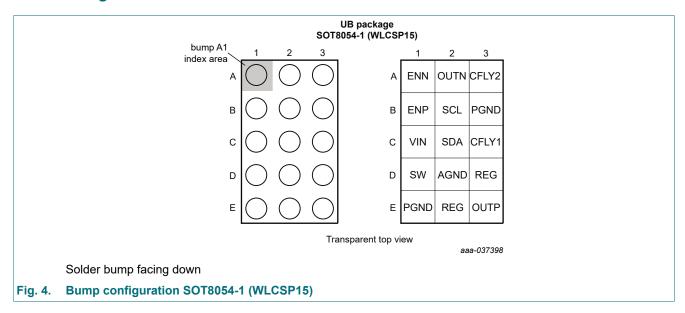
# 6. Functional diagram



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# 7. Pinning information

## 7.1. Pinning



## 7.2. Pin description

Table 3. Pin description

Symbol Bump		I/O	Description
	WLCSP15		
ENN	A1	I	Enable pin for V <sub>NEG</sub> rail
OUTN	A2	0	Output pin of the negative charge pump (V <sub>NEG</sub> )
CFLY2	A3	I/O	Negative charge pump flying capacitor pin
ENP	B1	I	Enable pin for V <sub>POS</sub> rail
SCL	B2	I/O	I <sup>2</sup> C interface clock signal pin
PGND	B3	-	Power ground
VIN	C1	I	Input voltage supply pin
SDA	C2	I/O	I <sup>2</sup> C interface data signal pin
CFLY1	C3	I/O	Negative charge pump flying capacitor pin
SW	D1	I/O	Switch pin of the boost converter
AGND	D2	-	Analog ground
REG	D3	I/O	Boost converter output pin
PGND	E1	-	Power ground
REG	E2	I/O	Boost converter output pin
OUTP	E3	0	Output pin of the LDO (V <sub>POS</sub> )

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# 8. Device comparison

#### **Table 4. Device Comparison**

Type number	Output current (I <sub>OUT</sub> )
NEX10001UB	220 mA
NEX10000UB[1]	80 mA

<sup>[1]</sup> Please see detail parameters from NEX10000UB data sheet.

# 9. Limiting values

## **Table 5. Limiting values**

In accordance with the Absolute Maximum Rating System (IEC 60134). Voltages are referenced to GND (ground = 0 V).

Symbol	Parameter	Conditions	Min	Max	Unit
VI	input voltage	pins ENN, ENP, VIN	-0.3	+5.5	V
Vo	output voltage	pin OUTN	-7	+0.3	V
		pin OUTP	-0.3	+7.0	V
V <sub>IO</sub>	input/output voltage	pins SCL, SDA	-0.3	+5.5	V
		pins CFLY1, REG	-0.3	+7.1	V
		pin SW	-0.3	+7.5	V
		pin CFLY2	-7	+0.3	V
Tj	junction temperature		-40	+150	°C
T <sub>amb</sub>	ambient temperature		-40	+85	°C
T <sub>stg</sub>	storage temperature		-65	+150	°C
ESD rati	ngs				
$V_{ESD}$	electrostatic discharge	HBM: ANSI/ESDA/JEDEC JS-001 class 2	-	±2000	V
		CDM: ANSI/ESDA/JEDEC JS-002 class C2a	-	±500	V

# 10. Recommended operating conditions

Table 6. Recommended operating conditions

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V <sub>IN</sub>	input voltage	supply voltage pin VIN	2.7	-	5.0	V
Io	output current	[1]	-	-	220	mA
L	inductor		2.2	-	4.7	μΗ
Co	output capacitance	pins OUTN, OUTP, REG	4.7	10	-	μF
C <sub>IO</sub>	input/output capacitance	pins CFLY1, CFLY2	2.2	-	4.7	μF
C <sub>I</sub>	input capacitance	pin VIN	-	4.7	-	μF
T <sub>amb</sub>	ambient temperature		-40	-	+85	°C
Tj	junction temperature		-40	-	+125	°C

<sup>[1]</sup> to support I<sub>POS</sub> + I<sub>NEG</sub> > 360 mA,  $V_{IN} \ge 3 \text{ V}$  is required.

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# 11. Thermal Information

#### **Table 7. Thermal information**

Symbol Parameter		Package	Unit
		SOT8054-1 (WLCSP15)	
$R_{\theta JA}$	Junction to ambient thermal resistance	76.4[1]	°C/W

<sup>[1]</sup> Measured in still air-free convection condition (conforms to EIA/JESD51-2) on high effective thermal conductivity JESD51-9 with a test board PCB.

## 12. Electrical characteristics

#### **Table 8. Electrical characteristics**

At recommended operating conditions;  $V_{IN} = 3.7 \text{ V}$ ;  $V_{ENN} = V_{ENP} = V_{IN}$ ;  $V_{POS} = 5.4 \text{ V}$ ;  $V_{NEG} = -5.4 \text{ V}$ ; (unless otherwise noted) voltages are referenced to GND (ground = 0 V).

Symbol	Parameter	Conditions	T <sub>amb</sub> =	Unit		
			Min	Typ[1]	Max	
Supply						
V <sub>IN</sub>	input voltage range		2.7	-	5.0	V
$V_{UVLO}$	undervoltage lockout	V <sub>IN</sub> rising	2.35	2.4	2.5	V
	threshold	V <sub>IN</sub> falling	2.1	2.2	2.25	V
Iq	quiescent current	V <sub>ENP</sub> = V <sub>ENN</sub> = 3.7 V, switching	-	0.79	1.04	mA
I <sub>sd</sub>	shutdown current	V <sub>ENP</sub> = V <sub>ENN</sub> = 0 V, no switching	-	-	1	μA
T <sub>sd</sub>	shutdown temperature	[2]	-	140	-	°C
T <sub>sd(hys)</sub>	shutdown temperature hysteresis	[2]	-	30	-	°C
Logic EN	N, ENP					
V <sub>IH</sub>	HIGH-level input voltage	V <sub>IN</sub> = 2.7 V to 5 V	1.27	-	-	V
V <sub>IL</sub>	LOW-level input voltage	V <sub>IN</sub> = 2.7 V to 5 V	-	-	0.53	V
R <sub>EN</sub>	pull-down resistor		-	220	-	ΚΩ
Logic SC	L, SDA					
V <sub>IH</sub>	HIGH-level input voltage	V <sub>IN</sub> = 2.7 V to 5 V	1.27	-	-	V
V <sub>IL</sub>	LOW-level input voltage	V <sub>IN</sub> = 2.7 V to 5 V	-	-	0.53	V
<b>Boost Co</b>	nverter					_
I <sub>LIM</sub>	boost converter current limit		1.3	1.6	1.9	Α
f <sub>SW</sub>	switching frequency	Boost converter switching frequency	1.15	1.4	1.65	MHz

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Symbol	Parameter	Conditions		T <sub>amb</sub> =	Unit		
					Typ[1]	Max	
LDO outp	ut V <sub>POS</sub>				<u>'</u>		
V <sub>POS</sub>	positive output voltage range			4.0	-	6.5	V
V <sub>POS_acc</sub>	positive output voltage accuracy			-1%	-	1%	
I <sub>POS</sub>	positive output current capability			220	-	-	mA
I <sub>POS_lim</sub>	positive output current limit	Register 0x02 LDO_CL = 0		255	275	310	mA
		Register 0x02 LDO_CL = 1		-	370	-	mA
V <sub>DROP</sub>	dropout voltage	$V_{REG} = V_{POS(nom)} = 5.4 \text{ V}, I_{OUT} = 80 \text{ mA}$		-	66	-	mV
		$V_{REG} = V_{POS(nom)} = 5.4 \text{ V}, I_{OUT} = 220 \text{ mA}$		-	192	-	mV
$\Delta V_{LR}$	line regulation	V <sub>IN</sub> = 2.7 V to 5 V, I <sub>OUT</sub> = 220 mA	[2]	-	5	-	mV
$\Delta V_{LDR}$	load regulation	V <sub>IN</sub> = 3.7 V, ΔI <sub>OUT</sub> = 220 mA	[2]	-	15	-	mV
R <sub>D</sub>	discharge resistor			-	23	-	Ω
Negative of	charge pump output V <sub>NEG</sub>						
$V_{NEG}$	negative output voltage range			-6.5	-	-4.0	V
V <sub>NEG_acc</sub>	negative output voltage accuracy			-1%	-	1%	V
I <sub>NEG</sub>	negative output current capability			220	-	-	mA
I <sub>NEG_lim</sub>	negative output current limit			245	275	310	mA
f <sub>OSC</sub>	negative charge pump switching frequency			0.85	1	1.15	MHz
$\Delta V_{LR}$	line regulation	V <sub>IN</sub> = 2.7 V to 5 V, I <sub>OUT</sub> = 220 mA	[2]	-	5	-	mV
$\Delta V_{LDR}$	load regulation	V <sub>IN</sub> = 3.7 V, ΔI <sub>OUT</sub> = 220 mA	[2]	-	10	-	mV
R <sub>D</sub>	discharge resistor			-	17.5	-	Ω
I <sup>2</sup> C Interfa	ce						
V <sub>IH</sub>	HIGH-level input voltage	SDA, SCL, 2.7 V ≤ V <sub>IN</sub> ≤ 5.0 V		1.27	-	-	V
V <sub>IL</sub>	LOW-level input voltage	SDA, SCL, 2.7 V ≤ V <sub>IN</sub> ≤ 5.0 V		-	-	0.53	V
f <sub>SCL</sub>	SCL clock frequency	pin SCL		-	-	400	KHz

 <sup>[1]</sup> All typical values are measured at T<sub>amb</sub> = 25 °C.
 [2] Guaranteed by bench test, not fully tested in production.

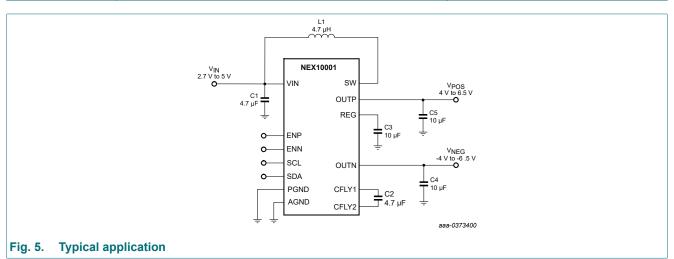
#### 220 mA dual output LCD bias power supply

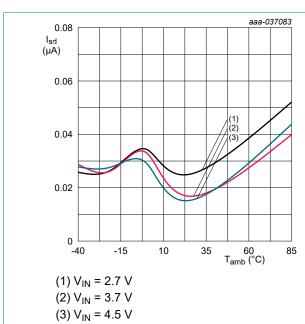
# 13. Typical characteristics

#### Table 9. Component list for typical characteristics circuit

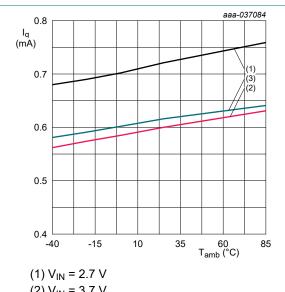
 $V_{IN}$  =  $V_{ENN}$  =  $V_{ENP}$  = 3.7 V;  $V_{POS}$  = 5.4 V;  $V_{NEG}$  = -5.4 V;  $T_{amb}$  = 25 °C; unless otherwise noted; voltages are referenced to GND (ground = 0 V). See Fig. 5

Components	Description	Manufacture and part number
C1	4.7 μF, 10 V, 0603, X5R, ceramic	TDK-C1608X5R1E475K080AC
C2	4.7 μF, 16 V, 0603, X5R, ceramic	TDK-C1608X5R1C225KTK00E
C3, C4, C5	10 μF, 10 V, 0603, X5R, ceramic	TDK-C1608X5R1E106K080AC
L1	4.7 μF, 2.2 A, 180 mΩ, 2.5 mm × 2.0 mm × 1.2 mm	TDK-TMS252012ALM-4R7MTAA
U1	NEX10001UB	NEXPERIA





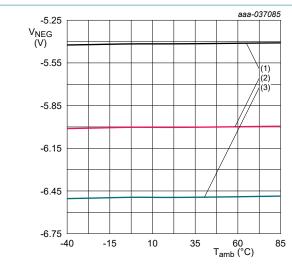
Shutdown current (I<sub>sd</sub>) versus ambient Fig. 6. temperature (T<sub>amb</sub>)



- $(2) V_{IN} = 3.7 V$
- $(3) V_{IN} = 4.5 V$

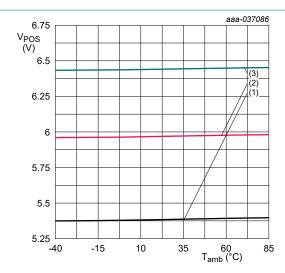
Fig. 7. Quiescent current (I<sub>a</sub>) versus ambient temperature (T<sub>amb</sub>)

### 220 mA dual output LCD bias power supply



- (1)  $V_{NEG} = -5.4 \text{ V}$
- (2)  $V_{NEG} = -6 V$
- (3)  $V_{NEG} = -6.5 V$

Fig. 8. Negative output voltage (V<sub>NEG</sub>) versus ambient temperature (T<sub>amb</sub>)



- (1)  $V_{POS} = 5.4 \text{ V}$
- (2)  $V_{POS} = 6 V$
- $(3) V_{POS} = 6.5 V$

Fig. 9. Positive output voltage (V<sub>POS</sub>) versus ambient temperature (T<sub>amb</sub>)

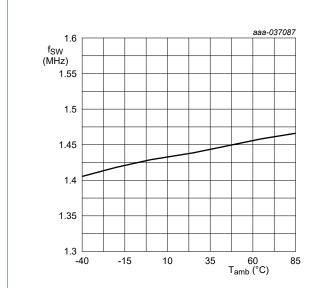
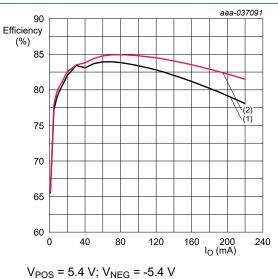


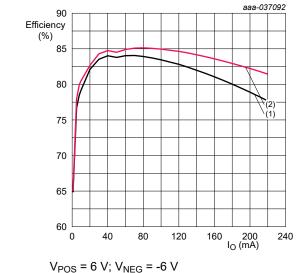
Fig. 10. Boost converter switching frequency  $(f_{SW})$  versus ambient temperature  $(T_{amb})$ 



- (1)  $V_{IN} = 3.7 V$
- $(2) V_{IN} = 4.5 V$

Fig. 11. Efficiency versus output current (I<sub>O</sub>)

#### 220 mA dual output LCD bias power supply

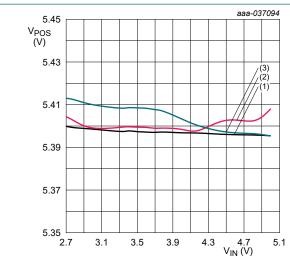


$$V_{POS} = 6 V$$
;  $V_{NEG} = -6 V$ 

(1) 
$$V_{IN} = 3.7 \text{ V}$$

$$(2) V_{IN} = 4.5 V$$

Fig. 12. Efficiency versus output current (I<sub>O</sub>)

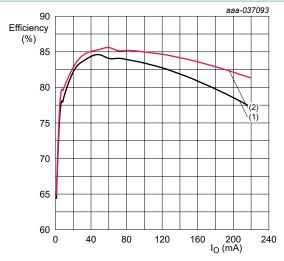


 $V_{POS} = 5.4 V$ 

(1)  $I_{POS} = 0 \text{ mA}$ 

(2)  $I_{POS} = 80 \text{ mA}$ (3)  $I_{POS} = 200 \text{ mA}$ 

Fig. 14.  $V_{POS}$  versus  $V_{IN}$  line regulation

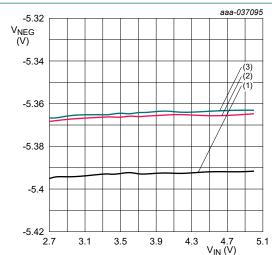


 $V_{POS} = 6.5 \text{ V}; V_{NEG} = -6.5 \text{ V}$ 

(1)  $V_{IN} = 3.7 \text{ V}$ 

(2)  $V_{IN} = 4.5 V$ 

Fig. 13. Efficiency versus output current (I<sub>O</sub>)



 $V_{NEG} = -5.4 V$ 

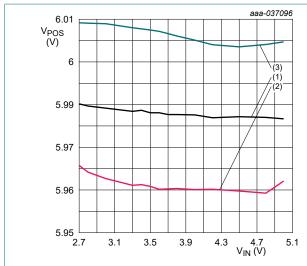
(1)  $I_{NEG} = 0 \text{ mA}$ 

(2)  $I_{NEG} = 80 \text{ mA}$ 

(3)  $I_{NEG} = 200 \text{ mA}$ 

Fig. 15. V<sub>NEG</sub> versus V<sub>IN</sub> line regulation

#### 220 mA dual output LCD bias power supply



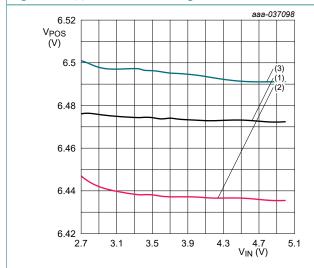
 $V_{POS} = 6 V$ 

(1)  $I_{POS} = 0 \text{ mA}$ 

(2)  $I_{POS} = 80 \text{ mA}$ 

(3)  $I_{POS} = 200 \text{ mA}$ 

Fig. 16. V<sub>POS</sub> versus V<sub>IN</sub> line regulation



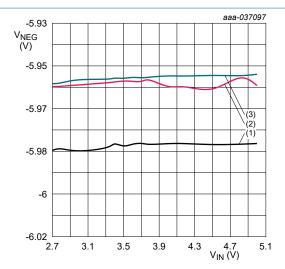
 $V_{POS} = 6.5 V$ 

(1)  $I_{POS} = 0 \text{ mA}$ 

(2)  $I_{POS} = 80 \text{ mA}$ 

(3)  $I_{POS} = 200 \text{ mA}$ 

Fig. 18.  $V_{POS}$  versus  $V_{IN}$  line regulation



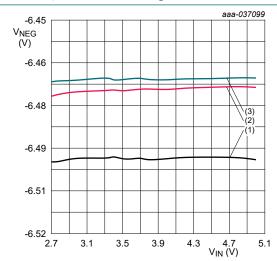
 $V_{NEG} = -6 V$ 

(1)  $I_{NEG} = 0 \text{ mA}$ 

(2)  $I_{NEG} = 80 \text{ mA}$ 

(3)  $I_{NEG} = 200 \text{ mA}$ 

Fig. 17. V<sub>NEG</sub> versus V<sub>IN</sub> line regulation



 $V_{NEG}$  = -6.5 V

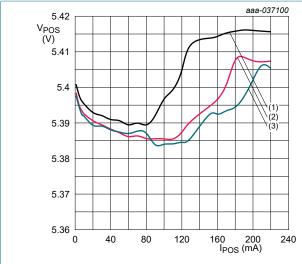
(1)  $I_{NEG} = 0 \text{ mA}$ 

(2)  $I_{NEG} = 80 \text{ mA}$ 

(3)  $I_{NEG} = 200 \text{ mA}$ 

Fig. 19. V<sub>NEG</sub> versus V<sub>IN</sub> line regulation

#### 220 mA dual output LCD bias power supply



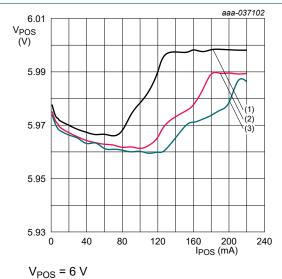
$$V_{POS} = 5.4 V$$

(1) 
$$V_{IN} = 2.7 \text{ V}$$

(2) 
$$V_{IN} = 3.7 \text{ V}$$

$$(3) V_{IN} = 4.5 V$$

Fig. 20. V<sub>POS</sub> versus positive output current (load regulation)



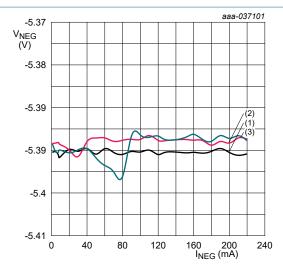
(1)  $V_{IN} = 2.7 V$ 

$$(1) V_{IN} = 2.7 V$$

(2) 
$$V_{IN} = 3.7 V$$

$$(3) V_{IN} = 4.5 V$$

Fig. 22. V<sub>POS</sub> versus positive output current (load regulation)



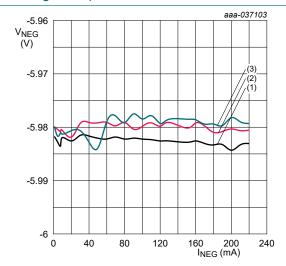
$$V_{NEG}$$
 = -5.4  $V$ 

(1) 
$$V_{IN} = 2.7 \text{ V}$$

(2) 
$$V_{IN} = 3.7 \text{ V}$$

(3) 
$$V_{IN} = 4.5 V$$

Fig. 21. V<sub>NEG</sub> versus negative output current (load regulation)



$$V_{NEG} = -6 V$$

(1) 
$$V_{IN} = 2.7 \text{ V}$$

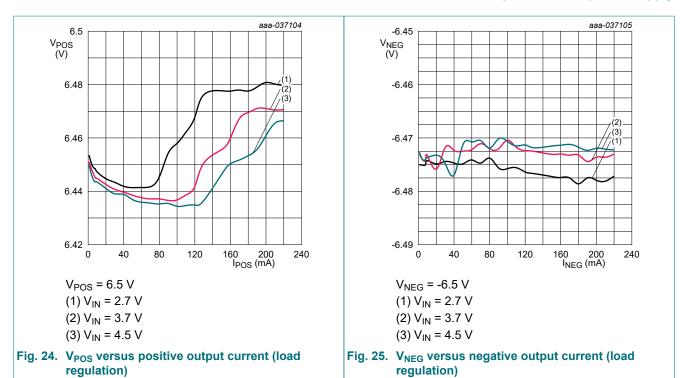
(2) 
$$V_{IN} = 3.7 V$$

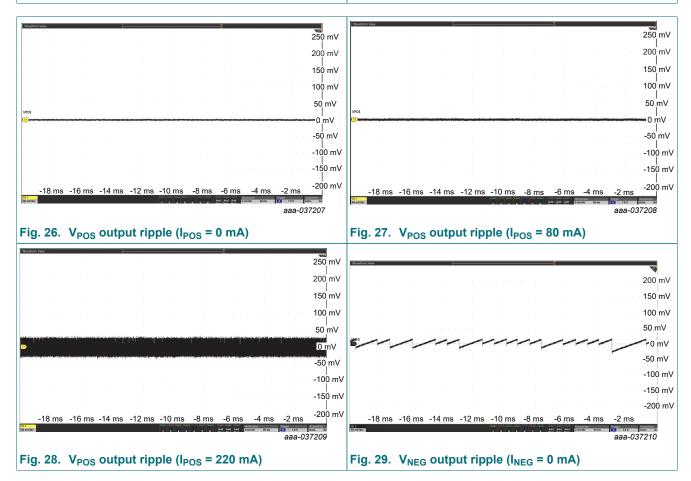
$$(3) V_{IN} = 4.5 V$$

Fig. 23. V<sub>NEG</sub> versus negative output current (load regulation)

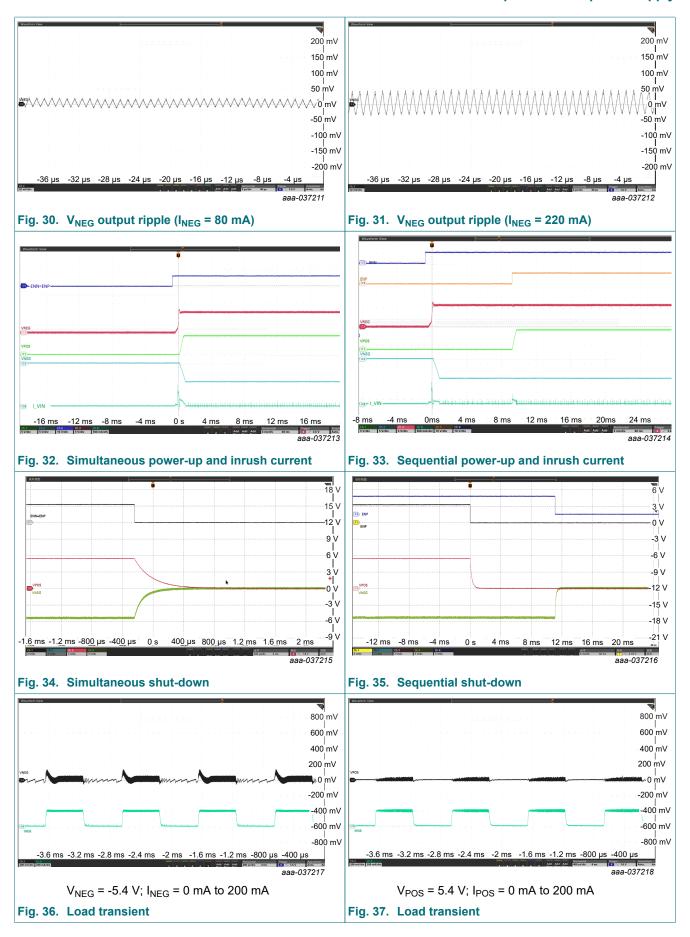
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#### 220 mA dual output LCD bias power supply

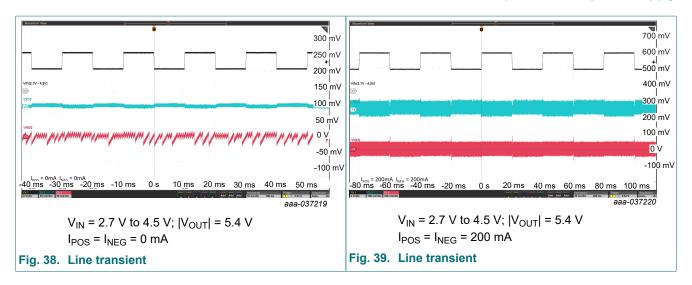




#### 220 mA dual output LCD bias power supply



#### 220 mA dual output LCD bias power supply



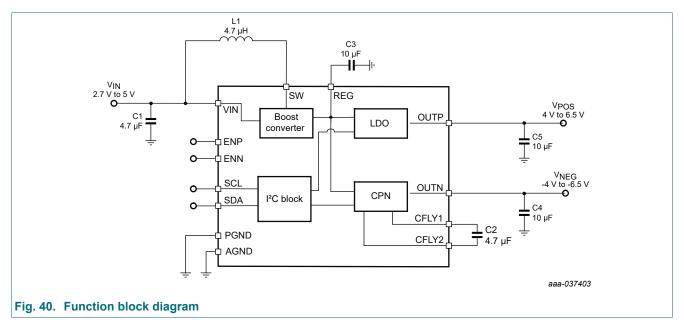
## 14. Detailed description

#### 14.1. Overview

The NEX10001, supporting input voltage range from 2.7 V to 5 V, operates with a single inductor scheme to provide high efficiency with small solution size. The synchronous boost converter generates a positive voltage that is regulated down by an integrated Low Dropout (LDO) regulator, outputting the positive supply rail ( $V_{POS}$ ). The negative supply rail ( $V_{NEG}$ ) is generated by an integrated negative charge pump (CPN) driven from the boost converter output pin, REG.The device is programmable over an I²C compatible interface for both  $V_{POS}$  and  $V_{NEG}$ . The device topology allows a 100% asymmetry of the output voltages and currents.

## 14.2. Function block diagram

The NEX10001 function block diagram is shown in Fig. 40.



#### 220 mA dual output LCD bias power supply

### 14.3. Feature description

#### 14.3.1. Undervoltage lockout (UVLO)

The NEX10001 integrates an undervoltage lockout block (UVLO) that enables the device once the voltage on the VIN pin exceeds the UVLO threshold (2.5 V maximum). No output voltage will be generated if the enable signals are not pulled HIGH. The device, as well as all converters (boost converter, LDO, CPN), will be disabled as soon as the VIN voltage falls below the UVLO threshold. The UVLO threshold is designed with a hysteresis, the NEX10001 will continue operating as long as VIN stays above 2.25 V. This guarantees a proper operation even in the event of extensive line transients when the battery gets suddenly heavily loaded.

#### 14.3.2. Thermal shutdown

The device has a build-in temperature sensor which monitors the internal junction temperature. When the junction temperature exceeds 140 °C, NEX10001 will shut down. When the junction temperature falls below the thermal recovery temperature, approximately 110 °C, the device restarts by using soft-start sequence.

#### 14.3.3. Boost converter

#### 14.3.3.1. Boost converter operation

The synchronous boost converter uses current mode topology with a typical switching frequency of 1.4 MHz, we recommend using 4.7  $\mu$ H inductor. The converter is internally compensated and provides a regulated output voltage automatically adjusted depending on the programmed  $V_{POS}$  and  $V_{NEG}$  voltages. The boost converter operates either in continuous conduction mode (CCM) or Pulse Frequency Modulation mode (PFM), depending on the load current to provide the highest efficiency.

#### 14.3.3.2. Power-up and soft-start (Boost Converter)

The Boost converter starts switching as soon as ENP or ENN is pulled high and the voltage on VIN pin is higher than UVLO threshold.

The boost converter starts up with internal soft start to avoid drawing excessive inrush current from the supply. The output voltage  $V_{REG}$  is slowly ramped up to its target value. Typical start-up waveforms for low-current applications are show in Fig. 32

#### 14.3.3.3. Power-down (Boost Converter)

The boost converter stops switching when VIN is below the UVLO threshold or when both ENP and ENN are pulled low.

### 14.3.3.4. Output voltage (Boost Converter)

 $V_{REG}$ , the output voltage of boost converter is automatically adjusted according to the programmed  $V_{POS}$  and  $V_{NEG}$  voltages.  $V_{REG}$  will maintain a fixed voltage difference with the large values in  $V_{POS}$  and  $V_{NEG}$ .

#### 14.3.3.5. Isolation (Boost Converter)

The boost converter output (REG) is isolated from the input supply VIN, providing a true shutdown.

#### 14.3.4. LDO regulator

#### 14.3.4.1. LDO operation

The Low Dropout regulator (LDO) generates the positive voltage rail,  $V_{POS}$ , by regulating down the output voltage of the boost converter ( $V_{REG}$ ). The high PSRR LDO helps filtering the output ripple of the boost converter to provide low ripple  $V_{POS}$  voltage.

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#### 14.3.4.2. Power-up and soft-start (LDO)

The LDO starts operating as soon as the ENP signal is pulled HIGH and  $V_{IN}$  voltage is above the UVLO threshold and the boost converter has reached its Power Good threshold. In the case where the enable signal is already HIGH when  $V_{IN}$  exceeds the UVLO threshold, the boost converter will start first and the LDO will only start after the boost converter has reached its target voltage.

The LDO integrates a soft start that slowly ramps up its output voltage  $V_{POS}$  regardless of the output capacitor and the target voltage, as long as the LDO current limit is not reached, the typical start-up time is 540  $\mu$ s. Typical start-up waveform for LDO are shown in Fig. 32.

#### 14.3.4.3. Power-down and discharge (LDO)

The LDO stops operating when  $V_{IN}$  is below the UVLO threshold or when ENP is pulled LOW. The NEX10001 integrates actively discharge function for  $V_{POS}$ . Typical power down waveform for LDO is shown in Fig. 34.

#### 14.3.4.4. Setting output voltage (LDO)

The output voltage of the LDO is programmable via  $I^2C$  interface, from 4.0 V to 6.5 V in 100 mV steps. For more details, please refer to the  $V_{POS}$  register (address: 0x00)

#### 14.3.4.5. Isolation (LDO)

The LDO isolates the  $V_{POS}$  rail from  $V_{REG}$  (boost converter output) as long as the rail is not enabled in order to ensure flexible start-up like  $V_{NEG}$  before  $V_{POS}$ .

#### 14.3.5. Negative charge pump

#### 14.3.5.1. **CPN** operation

The negative charge pump (CPN) generates the negative voltage rail,  $V_{NEG}$ , by inverting and regulating the output voltage of the boost converter ( $V_{REG}$ ). The charge pump uses 4 switches and an external flying capacitor to generate the negative rail. Two of the switches are turned on in the first phase to charge the flying capacitor up to  $V_{REG}$ , and in the second phase they are turned-off and the two others turn on to pump the energy negatively out of the OUTN capacitor.

#### 14.3.5.2. Power-up and soft-start (CPN)

The CPN starts operating as soon as the ENN signal is pulled HIGH and VIN voltage is above the UVLO threshold and the boost converter has reached its power good threshold. In the case where the enable signal is already HIGH when VIN reaches the UVLO threshold, the boost converter will start first and the CPN will only start after the boost converter has reached its target voltage. Typical startup waveform for CPN is shown in Fig. 32.

The CPN integrates a soft start that slowly ramps up its output voltage  $V_{NEG}$  within a time. The time is determined by the output voltage and the output capacitor value.

The estimated start-up time can be calculated using the following formula:  $t_{\text{STARTUP}} = \frac{c_{OUT} \times v_{NEG}}{I_{STARTUP}}$ 

Where:

 $t_{STARTUP}$  = start-up time of the  $V_{NFG}$  rail

 $C_{OUT}$  = output capacitance of the  $V_{NEG}$  rail

V<sub>NEG</sub> = target output voltage

I<sub>STARTUP</sub> = the output current of V<sub>NEG</sub> to charge up the output capacitor at start-up

#### 14.3.5.3. Power-down and discharge (CPN)

The CPN will stop operating when  $V_{IN}$  is below the UVLO threshold or when ENN is pulled LOW. NEX10001 integrates actively discharge function for  $V_{NEG}$ . Typical start-up waveform for  $V_{NEG}$  is shown in Fig. 32.

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### 14.3.5.4. Setting output voltage (CPN)

The output voltage of the CPN is programmable via  $I^2C$  interface, from -6.5 V to -4.0 V in 100 mV steps. For more details, please refer to the  $V_{NEG}$  register (address: 0x01).

## 14.3.5.5. Isolation (CPN)

The CPN isolates the  $V_{NEG}$  rail from  $V_{REG}$  (boost converter output) as long as the rail is not enabled in order to ensure flexible start-up like  $V_{POS}$  before  $V_{NEG}$ .

#### 14.3.6. I<sup>2</sup>C serial interface

NEX10001 integrates an industry standard I<sup>2</sup>C compatible interface as a slave. The slave address is 0x3E.

### 14.3.7. Register configuration

Table 10. Register map

Address	Register name	R/W	Bit7	Bit6	Bit5	Bit4	Bit3	Bit2	Bit1	Bit0	Default
0x00[1]	V <sub>POS</sub> register	R/W	reserved		V <sub>POS</sub> [4:0]					0x0E	
0x01[1]	V <sub>NEG</sub> register	R/W	reserved			V <sub>NEG</sub> [4:0]					0x0E
0x2	CL and Vendor ID	R/W		reserved		LDO_CL	rese	rved	1	1	0x03

<sup>[1] 0</sup>x00 and 0x01: the relationship between  $V_{POS}$ ,  $V_{NEG}$  registers and  $V_{POS}$ ,  $V_{NEG}$  voltages are shown in Table 11.

Table 11. V<sub>POS</sub>and V<sub>NEG</sub> voltages

V <sub>POS</sub> [4:0]			<b>V<sub>NEG</sub> voltage</b>
0x00	4.0	0x00	-4.0
0x01	4.1	0x01	-4.1
0x02	4.2	0x02	-4.2
0x03	4.3	0x03	-4.3
0x04	4.4	0x04	-4.4
0x05	4.5	0x05	-4.5
0x06	4.6	0x06	-4.6
0x07	4.7	0x07	-4.7
0x08	4.8	0x08	-4.8
0x09	4.9	0x09	-4.9
0x0A	5.0	0x0A	-5.0
0x0B	5.1	0x0B	-5.1
0x0C	5.2	0x0C	-5.2
0x0D	5.3	0x0D	-5.3
0x0E	5.4 (Default value)	0x0E	-5.4 (Default value)
0x0F	5.5	0x0F	-5.5
0x10	5.6	0x10	-5.6
0x11	5.7	0x11	-5.7
0x12	5.8	0x12	-5.8
0x13	5.9	0x13	-5.9
0x14	6.0	0x14	-6.0
0x15	6.1	0x15	-6.1
0x16	6.2	0x16	-6.2

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V <sub>POS</sub> [4:0]	V <sub>POS</sub> voltage	V <sub>NEG</sub> [4:0]	V <sub>NEG</sub> voltage
0x17	6.3	0x17	-6.3
0x18	6.4	0x18	-6.4
0x19	6.5	0x19	-6.5

## Table 12. Definition of 0x02 register

Table 121 Definition of the Togleton				
Bit	Field	Description	Value	Action
7:5	Reserved	Reserved	Reserved	Reserved
4	LDO_CL	V <sub>POS</sub> current limit	0	current limit 280 mA
			1	current limit 370 mA
3:2	Reserved	Reserved	Reserved	Reserved
1:0	Vender ID	Vendor ID	11	Read only Vender ID

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# 15. Package outline

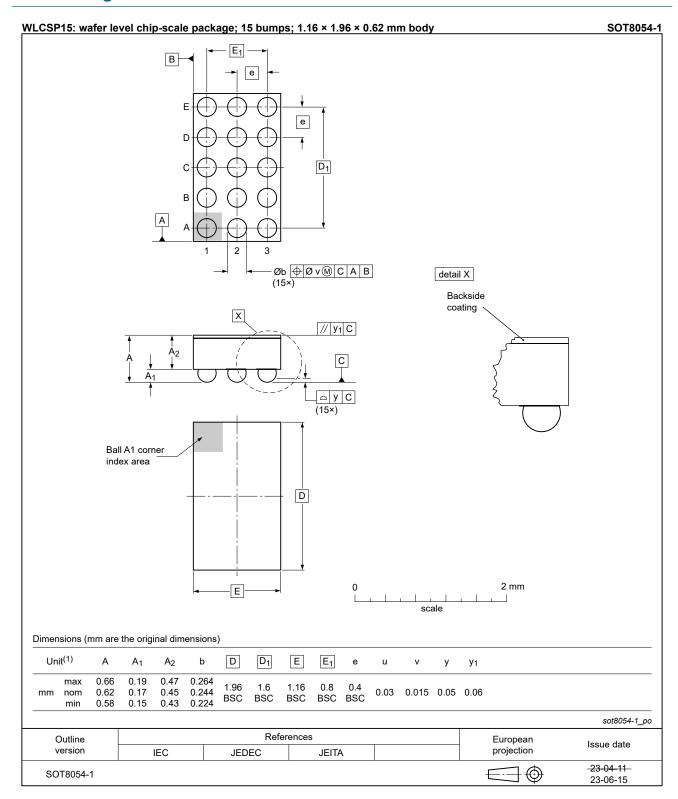


Fig. 41. Package outline SOT8054-1 (WLCSP15)

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# 16. Abbreviations

#### **Table 13. Abbreviations**

Acronym	Description
ССМ	Continuous Conduction Mode
CDM	Charged Device Model
CPN	Negative Charge Pump
ESD	ElectroStatic Discharge
НВМ	Human Body Model
LDO	Low Dropout
OTSD	Over Temperature Shut Down
PCB	Printed Circuit Board
PFM	Pulse Frequency Modulation
PSRR	Power Supply Rejection Ratio
TFT-LCD	Thin Film Transistor-Liquid Crystal Display
UVLO	Undervoltage Lockout

# 17. Revision history

#### **Table 14. Revision history**

Document ID	Release date	Data sheet status	Change notice	Supersedes
NEX10001UB v. 2	20231220	Product data sheet	-	NEX10001UB v. 1
Modifications:	<u>Section 5</u> : marking code has changed			
NEX10001UB v. 1	20231106	Product data sheet	-	-

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# 18. Legal information

#### **Data sheet status**

Document status [1][2]	Product status [3]	Definition
Objective [short] data sheet	Development	This document contains data from the objective specification for product development.
Preliminary [short] data sheet	Qualification	This document contains data from the preliminary specification.
Product [short] data sheet	Production	This document contains the product specification.

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- [2] The term 'short data sheet' is explained in section "Definitions".
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