



# VFC32

# Voltage-to-Frequency and Frequency-to-Voltage CONVERTER

### **FEATURES**

- OPERATION UP TO 500kHz
- EXCELLENT LINEARITY ±0.01% max at 10kHz FS ±0.05% max at 100kHz FS
- V/F OR F/V CONVERSION
- MONOTONIC
- VOLTAGE OR CURRENT INPUT

# **APPLICATIONS**

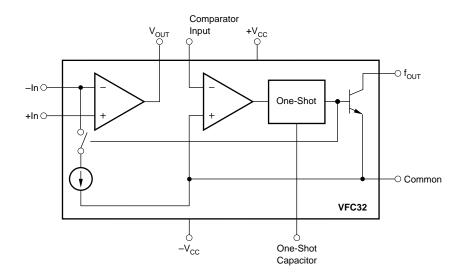
- INTEGRATING A/D CONVERTER
- SERIAL FREQUENCY OUTPUT
- ISOLATED DATA TRANSMISSION
- FM ANALOG SIGNAL MOD/DEMOD
- MOTOR SPEED CONTROL
- TACHOMETER

# DESCRIPTION

The VFC32 voltage-to-frequency converter provides an output frequency accurately proportional to its input voltage. The digital open-collector frequency output is compatible with all common logic families. Its integrating input characteristics give the VFC32 excellent noise immunity and low nonlinearity.

Full-scale output frequency is determined by an external capacitor and resistor and can be scaled over a wide range. The VFC32 can also be configured as a frequency-to-voltage converter.

The VFC32 is available in 14-pin plastic DIP, SO-14 surface-mount, and metal TO-100 packages. Commercial, industrial, and military temperature range models are available.



International Airport Industrial Park • Mailing Address: PO Box 11400 • Tucson, AZ 85734 • Street Address: 6730 S. Tucson Blvd. • Tucson, AZ 85706

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# **SPECIFICATIONS**

At  $T_A = +25$ °C and  $V_{CC} = \pm 15$ V, unless otherwise noted.

		VFC32KP, KU			VFC32BM			VFC32SM				
PARAMETER	CONDITIONS	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS	
INPUT (V/F CONVERTER)	$F_{OUT} = V_{IN}/7.5 R_1 C_1$											
Voltage Range <sup>(1)</sup>												
Positive Input		>0		+0.25mA	*		*	*		*	V	
				x R <sub>1</sub>								
Negative Input		>0		-10	*		*	*		*	V	
Current Range <sup>(1)</sup>		>0		+0.25	*		*	*		*	mA	
Bias Current			00	400							A	
Inverting Input			20	100		*	*		*	*	nA	
Noninverting Input Offset Voltage <sup>(2)</sup>			100	250 4		* *	*		*	*	nA mV	
Differential Impedance		300    10	650    10	4	*	*	*	*	*	*	kΩ    pF	
Common-mode		300    10	030    10		~	~		~	~		K22    þi	
Impedance		300    3	500    3		*	*		*	*		MΩ    pF	
•	<u> </u>	000    0	000    0		•			-			10122    P1	
INPUT (F/V CONVERTER)	$V_{OUT} = 7.5 R_1 C_1 F_{IN}$	50 II 40	450    40								1:0 II = E	
Impedance		50    10	150    10		*	*		*	*		kΩ    pF	
Logic "1" Logic "0"			+1.0 -0.05		* *		*	*		*	V V	
Pulse-width Range		0.1	-0.05	150k/F <sub>MAX</sub>	*		*	*		*		
		0.1		TOOKT MAX	~		~	~		~	μs	
ACCURACY	0.0411 0											
Linearity Error <sup>(3)</sup>	0.01Hz ≤ Oper		10.005	10.040(4)							0/ -4 505/5	
	Freq ≤ 10kHz		±0.005	±0.010 <sup>(4)</sup>		*	*		*	*	% of FSR <sup>(5)</sup>	
	0.1Hz ≤ Oper Freq ≤ 100kHz		±0.025	±0.05		*	*		*	*	% of FSR	
	l '		±0.025	±0.05		*	*		*	*	% 01 F5R	
	0.5Hz ≤ Oper Freq ≤ 500kHz		±0.05			*			*		% of FSR	
Offset Error Input	1 1eq 5 300KHZ		±0.03			~			~		/6 UT 3K	
Offset Votlage <sup>(2)</sup>			1	4		*	*		*	*	mV	
Offset Drift <sup>(6)</sup>			±3			*			*	"	ppm of FSR/°C	
Gain Error <sup>(2)</sup>			5			*			*		% of FSR	
Gain Drift <sup>(6)</sup>	f = 10kHz		±75			±50	±100		±70	±150	ppm/°C	
Full Scale Drift	f = 10kHz		±75			±50	±100		±70	±150	ppm of FSR/°C	
(offset drift and												
gain drift)(6, 7)												
Power Supply	$f = DC$ , $\pm V_{CC} = 12VDC$											
Sensitivity	to 18VDC			±0.015			*			*	% of FSR/%	
OUTPUT (V/F CONVERTE	R) (open collector output)											
Voltage, Logic "0"	I <sub>SINK</sub> = 8mA	0	0.2	0.4	*	*	*	*	*	*	V	
Leakage Current,	I <sub>SINK</sub> = OTTA	"	0.2	0.4	•	^	•	^	^	^	V V	
Logic "1"	V <sub>o</sub> = 15V		0.01	1.0		*	*		*	*	μА	
Voltage, Logic "1"	External Pull-up Resistor		0.01	1.0		~	~		~	~	μΑ	
Voltago, Logio	Required (see Figure 4)			V <sub>PU</sub>			*			*	V	
Pulse Width	For Best Linearity		0.25/F <sub>MAX</sub>	* PU		*			*	"	s	
Fall Time	$I_{OUT} = 5\text{mA}, C_{LOAD} = 500\text{pF}$		O.ZO/I MAX	400			*			*	ns	
OUTPUT (F/V CONVERTE			-				· ·			<u> </u>		
					_							
Voltage	I <sub>o</sub> ≤ 7mA	0 to +10			*			*			V	
Current	V <sub>o</sub> ≤ 7VDC	+10		,	*			*			mA	
Impedance	Closed Loop	1		1			*			*	Ω	
Capacitive Load	Without Oscillation	ļ		100			*			*	pF	
DYNAMIC RESPONSE												
Full Scale Frequency				500(8)	*			*			kHz	
Dynamic Range		6			*			*			decades	
Settling Time	(V/F) to Specified Linearity	1	100			l .						
0 1 10	for a Full Scale Input Step		(9)			*			*			
Overload Recovery	< 50% Overload		(9)			*			*			
POWER SUPPLY												
Rated Voltage			±15					1			V	
Voltage Range		±11		±20					*		V	
Quiescent Current			±5.5	±6.0		*	*		*		mA	
TEMPERATURE RANGE												
Specification		0		+70	-25		+85	-55		+125	°C	
	i .	I 05	1			I	+125	-55	1	+125	°C	
Operating		-25		+85	-55		+123	-33		+125		

<sup>\*</sup> Specification the same as VFC32KP.

NOTES: (1) A 25% duty cycle (0.25mA input current) is recommended for best linearity. (2) Adjustable to zero. See Offset and Gain Adjustment section. (3) Linearity error is specified at any operating frequency from the straight line intersecting 90% of full scale frequency and 0.1% of full scale frequency. See Discussion of Specifications section. Above 200kHz, it is recommended all grades be operated below +85°C. (4) ±0.015% of FSR for negative inputs shown in Figure 5. Positive inputs are shown in Figure 1. (5) FSR = Full Scale Range (corresponds to full scale frequency and full scale input voltage). (6) Exclusive of external components' drift. (7) Positive drift is defined to be increasing frequency with increasing temperature. (8) For operations above 200kHz up to 500kHz, see Discussion of Specifications and Installation and Operation sections. (9) One pulse of new frequency plus 1µs.

#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage	±22V
Output Sink Current (FOUT)	50mA
Output Current (Vour)	+20mA
Input Voltage, -Input	
Input Voltage, +Input	±Supply
Comparator Input	±Supply
Storage Temperature Range:	
VFC32BM, SM	
VFC32KP, KU	25°C to +85°C

#### PACKAGE/ORDERING INFORMATION

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER <sup>(1)</sup>	TEMPERATURE RANGE
VFC32KP	14-Pin Plastic DIP	010	0°C to 70°C
VFC32BM	TO-100 Metal	007	–25°C to +85°C
VFC32SM	TO-100 Metal	007	-55°C to +125°C
VFC32KU	SO-14 SOIC	235	0°C to +70°C

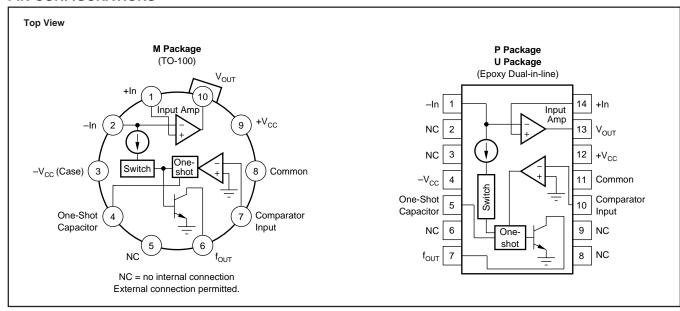
NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.

# ELECTROSTATIC DISCHARGE SENSITIVITY

This integrated circuit can be damaged by ESD. Burr-Brown recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### **PIN CONFIGURATIONS**



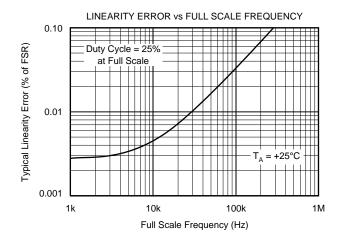
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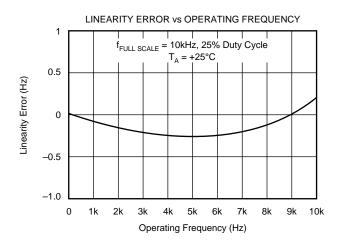
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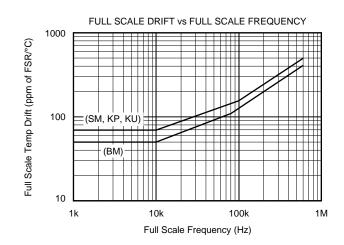
VFC32

# **TYPICAL PERFORMANCE CURVES**

At  $T_{\text{A}}$  = +25°C and  $V_{\text{CC}}$  =  $\pm 15 \text{V},$  unless otherwise noted.







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# APPLICATION INFORMATION

Figure 1 shows the basic connection diagram for frequency-to-voltage conversion.  $R_1$  sets the input voltage range. For a 10V full-scale input, a  $40k\Omega$  input resistor is recommended. Other input voltage ranges can be achieved by changing the value of  $R_1$ .

$$R_1 = \frac{V_{FS}}{0.25 \text{mA}} \tag{1}$$

 $R_1$  should be a metal film type for good stability. Manufacturing tolerances can produce approximately  $\pm 10\%$  variation in output frequency. Full-scale output frequency can be trimmed by adjusting the value of  $R_1$ —see Figure 3.

The full-scale output frequency is determined by C<sub>1</sub>. Values shown in Figure 1 are for a full-scale output frequency of 10kHz. Values for other full-scale frequencies can be read from Figure 2. Any variation in C<sub>1</sub>—tolerance, temperature drift, aging—directly affect the output frequency. Ceramic NPO or silver-mica types are a good choice.

For full-scale frequencies above 200kHz, use larger capacitor values as indicated in Figure 2, with  $R_1 = 20k\Omega$ .

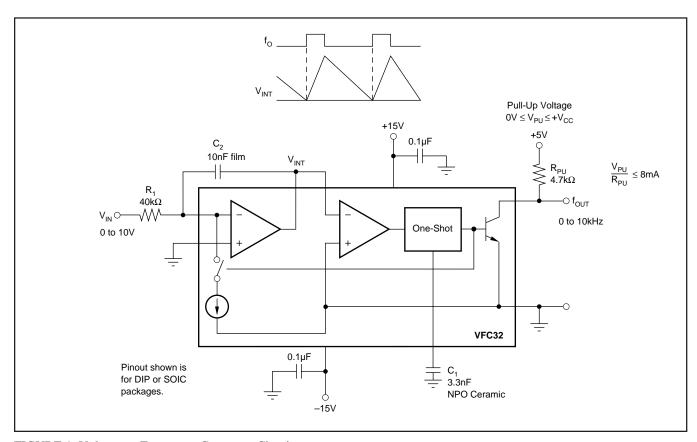
The value of the integrating capacitor, C<sub>2</sub>, does not directly influence the output frequency, but its value must be chosen within certain bounds. Values chosen from Figure 2 produce

approximately 2.5Vp-p integrator voltage waveform. If  $C_2$ 's value is made too low, the integrator output voltage can exceed its linear output swing, resulting in a nonlinear response. Using  $C_2$  values larger than shown in Figure 2 is acceptable.

Accuracy or temperature stability of C<sub>2</sub> is not critical because its value does not directly affect the output frequency. For best linearity, however, C<sub>2</sub> should have low leakage and low dielectric absorption. Polycarbonate and other film capacitors are generally excellent. Many ceramic types are adequate, but some low-voltage ceramic capacitor types may degrade nonlinearity. Electrolytic types are not recommended.

#### FREQUENCY OUTPUT PIN

The frequency output terminal is an open-collector logic output. A pull-up resistor is usually connected to a 5V logic supply to create standard logic-level pulses. It can, however, be connected to any power supply up to  $+V_{\rm CC}$ . Output pulses have a constant duration and positive-going during the one-shot period. Current flowing in the open-collector output transistor returns through the Common terminal. This terminal should be connected to logic ground.



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FIGURE 1. Voltage-to-Frequency Converter Circuit.

VFC32

#### FREQUENCY-TO-VOLTAGE CONVERSION

Figure 4 shows the VFC32 connected as a frequency-to-voltage converter. The capacitive-coupled input network  $C_3$ ,  $R_6$  and  $R_7$  allow standard 5V logic levels to trigger the comparator input. The comparator triggers the one-shot on the falling edge of the frequency input pulses. Threshold voltage of the comparator is approximately -0.7V. For frequency input waveforms less than 5V logic levels, the  $R_6/R_7$  voltage divider can be adjusted to a lower voltage to assure that the comparator is triggered.

The value of  $C_1$  is chosen from Figure 2 according to the full-scale input frequency.  $C_2$  smooths the output voltage waveform. Larger values of  $C_2$  reduce the ripple in the output voltage. Smaller values of  $C_2$  allow the output voltage to settle faster in response to a change in input frequency. Resistor  $R_1$  can be trimmed to achieve the desired output voltage at the full-scale input frequency.

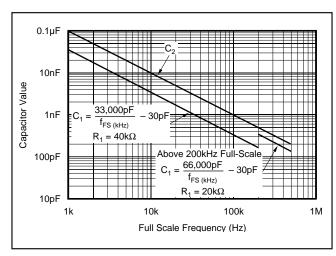


FIGURE 2. Capacitor Value Selection.

#### PRINCIPLES OF OPERATION

The VFC32 operates on a principle of charge balance. The signal input current is equal to  $V_{\rm IN}/R_1$ . This current is integrated by input op amp and  $C_2$ , producing a downward ramping integrator output voltage. When the integrator output ramps to the threshold of the comparator, the one-shot is triggered. The 1mA reference current is switched to the integrator input during the one-shot period, causing the integrator output ramp upward. After the one-shot period, the integrator again ramps downward.

The oscillation process forces a long-term balance of charge (or average current) between the input signal current and the reference current. The equation for charge balance is:

$$I_{IN} = I_{R(AVERAGE)}$$
 (2)

$$\frac{V_{IN}}{R_1} = f_O t_{OS} (1mA) \tag{3}$$

Where:

 $f_{o}$  is the output frequency  $t_{os}$  is the one-shot period, equal to  $t_{os} = 7500 \text{ C}_{1} \text{ (Farads)}$  (4)

The values suggested for  $R_1$  and  $C_1$  are chosen to produce a 25% duty cycle at full-scale frequency output. For full-scale frequencies above 200kHz, the recommended values produce a 50% duty cycle.

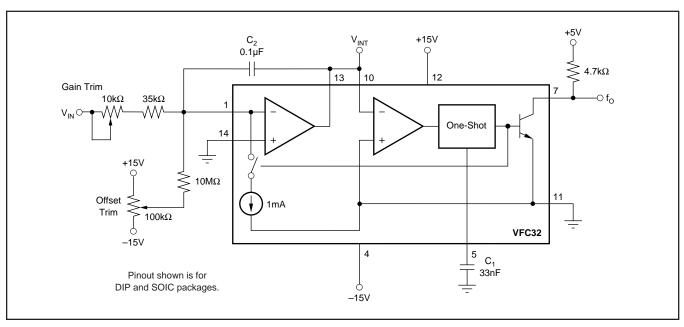


FIGURE 3. Gain and Offset Voltage Trim Circuit.

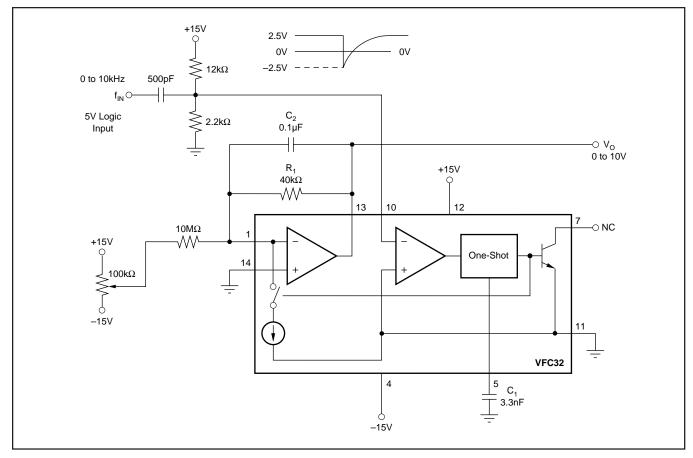
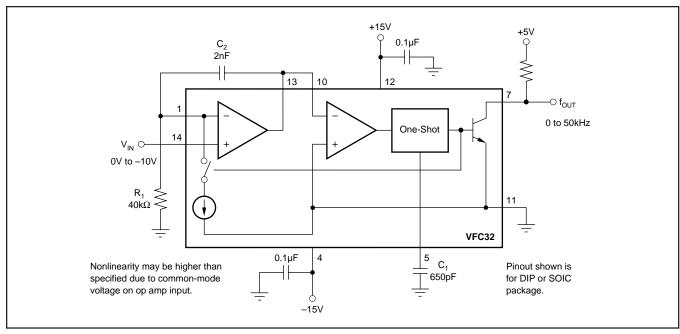


FIGURE 4. Frequency-to-Voltage Converter Circuit.



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FIGURE 5. V/F Converter—Negative Input Voltage.





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#### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
VFC32BM	OBSOLETE	TO-100	LME	10		TBD	Call TI	Call TI
VFC32KP	ACTIVE	PDIP	N	14	25	TBD	CU NIPDAU	Level-NA-NA-NA
VFC32KU	ACTIVE	SOIC	D	14	58	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
VFC32KU/2K5	ACTIVE	SOIC	D	14	2500	Pb-Free (RoHS)	CU NIPDAU	Level-3-260C-168 HR
VFC32KU/2K5E4	ACTIVE	SOIC	D	14	2500	TBD	Call TI	Call TI
VFC32KUE4	ACTIVE	SOIC	D	14	58	TBD	Call TI	Call TI

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

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(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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