

August 1992

_M6162/LM6262/LM6362 High Speed Operational Amplifie

LM6162/LM6262/LM6362 **High Speed Operational Amplifier**

General Description

The LM6362 family of high-speed amplifiers exhibits an excellent speed-power product, delivering 300 V/µs and 100 MHz gain-bandwidth product (stable for gains as low as +2 or -1) with only 5 mA of supply current. Further power savings and application convenience are possible by taking advantage of the wide dynamic range in operating supply voltage which extends all the way down to +5V.

These amplifiers are built with National's VIP™ (Vertically Integrated PNP) process which provides fast transistors that formance without the need for complex and expensive dielectric isolation.

are true complements to the already fast NPN devices. This advanced junction-isolated process delivers high speed per-

Features ■ High slew rate: 300 V/µs ■ High gain-bandwidth product: 100 MHz

■ Low supply current: 5 mA

■ Fast settling time: 120 ns to 0.1%

■ Low differential gain: <0.1%

■ Low differential phase: <0.1°

■ Wide supply range: 4.75V to 32V

■ Stable with unlimited capacitive load

■ Well behaved; easy to apply

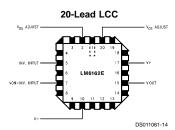
Applications

- Video amplifier
- Wide-bandwidth signal conditioning for image processing (FAX, scanners, laser printers)

DS011061-15

- Hard disk drive preamplifier
- Error amplifier for high-speed switching regulator

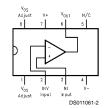
Connection Diagrams



Top View See NS Package Number E20A

10-Pin Ceramic Flatpak V_{OS} ADJUST □ INV INPUT NON-INV INPUT

Top View See NS Package Number W10A



See NS Package Number N08E, M08A or J08A

	Temperature Range	Package	NSC	
Military	Industrial	Commercial		Drawing
-55°C ≤ T _A ≤ +125°C	-25° C \leq T _A \leq +85 $^{\circ}$ C	0° C \leq T _A \leq +70 $^{\circ}$ C		
LM6162N	LM6262N	LM6362N	8-Pin Molded DIP	N08E
LM6162J/883			8-Pin Ceramic DIP	J08A
5962-9216501PA				
	LM6262M	LM6362M	8-Pin Molded Surface Mt.	M08A
LM6162E/883			20-Lead LCC	E20A
5962-92165012A				
LM6162W/883			10-Pin Ceramic Flatpak	W10A
5962-9216501HA				

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Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage (V+-V-) 36V Differential Input Voltage (Note 2) ±8V Common-Mode Input Voltage $(V^+-0.7V)$ to (Note 3) $(V^- + 0.7V)$

Output Short Circuit to GND

(Note 4)

Soldering Information

Dual-In-Line Package (N) Soldering (10 seconds)

Small Outline Package (M)

Vapor Phase (60 seconds)

Infrared (15 seconds)

220°C

See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering surface mount devices.

Storage Temperature Range $-65^{\circ}C \leq T_{J} \leq +150^{\circ}C$ 150°C Max Junction Temperature ±1100V ESD Tolerance (Note 5)

Operating Ratings

Temperature Range (Note 6)

LM6162 $-55^{\circ}C \leq T_{J} \leq +125^{\circ}C$ LM6262 $-25^{\circ}C \leq T_{J} \leq +85^{\circ}C$ LM6362 $0^{\circ}C \leq T_{J} \leq +70^{\circ}C$ Supply Voltage Range 4.75V to 32V

DC Electrical Characteristics

These limits apply for supply voltage = ± 15 V, $V_{CM} = 0$ V, and $R_L \ge 100~k\Omega$, unless otherwise specified. Limits in standard type-face are for $T_A = T_J = 25$ °C; limits in **boldface type** apply over the **Operating Temperature Range**.

Continuous

260°C

215°C

			Typical	LM6162	LM6262	LM6362	
Symbol	Parameter	Conditions	(Note 7)	Limit	Limit	Limit	Units
				(Note 8)	(Note 8)	(Note 8)	
V _{OS}	Input Offset Voltage		±3	±5	±5	±13	mV
				±8	±8	±15	max
ΔV _{OS}	Input Offset Voltage		7				μV/°C
ΔTemp	Average Drift						
I _{bias}	Input Bias Current		2.2	3	3	4	μA
				6	5	6	max
I _{os}	Input Offset Current		±150	±350	±350	±1500	nA
				±800	±600	±1900	max
Δlos	Input Offset Current		0.3				nA/°C
ΔTemp	Average Drift						
R _{IN}	Input Resistance	Differential	180				kΩ
C _{IN}	Input Capacitance		2.0				pF
A _{VOL}	Large Signal	$V_{OUT} = \pm 10V$, $R_L = 2 k\Omega$	1400	1000	1000	800	V/V
	Voltage Gain	(Note 9)		500	700	650	min
		$R_L = 10 \text{ k}\Omega$	6500				V/V
V _{CM}	Input Common-Mode	Supply = ±15V	+14.0	+13.9	+13.9	+13.8	V
	Voltage Range			+13.8	+13.8	+13.7	min
			-13.2	-12.9	-12.9	-12.9	V
				-12.7	-12.7	-12.8	max
		Supply = +5V	4.0	3.9	3.9	3.8	V
		(Note 9)		3.8	3.8	3.7	min
			1.6	1.8	1.8	1.9	V
				2.0	2.0	2.0	max
CMRR	Common-Mode	-10V ≤ V _{CM} ≤ +10V	100	83	83	76	dB
	Rejection Ratio			79	79	74	min
PSRR	Power Supply	±10V ≤ V _S ≤ ±16V	93	83	83	76	dB
	Rejection Ratio			79	79	74	min

DC Electrical Characteristics (Continued)

These limits apply for supply voltage = ± 15 V, V_{CM} = 0V, and $R_L \ge 100~k\Omega$, unless otherwise specified. Limits in standard type-face are for $T_A = T_J = 25$ °C; limits in **boldface type** apply over the **Operating Temperature Range**.

Symbol	Parameter	Conditions	Typical (Note 7)	LM6162 Limit	LM6262 Limit	LM6362 Limit	Units
-				(Note 8)	(Note 8)	(Note 8)	
Vo	Output Voltage	Supply = $\pm 15V$, R _L = 2 k Ω	+14.2	+13.5	+13.5	+13.4	V
	Swing			+13.3	+13.3	13.3	min
			-13.4	-13.0	-13.0	-12.9	V
				-12.7	-12.8	-12.8	max
Vo	Output Voltage Swing	Supply = +5V and	4.2	3.5	3.5	3.4	V
		$R_L = 2 k\Omega$ (Note 10)		3.3	3.3	3.3	min
			1.3	1.7	1.7	1.8	V
				2.0	1.9	1.9	max
I _{osc}	Output Short	Sourcing	65	30	30	30	mA
	Circuit Current			20	25	25	min
		Sinking	65	30	30	30	mA
				20	25	25	min
Is	Supply Current		5.0	6.5	6.5	6.8	mA
				6.8	6.7	6.9	max

AC Electrical Characteristics

These limits apply for supply voltage = ± 15 V, V_{CM} = 0V, R_L ≥ 100 k Ω , and C_L ≤ 5 pF, unless otherwise specified. Limits in standard typeface are for T_A = T_J = 25°C; limits in **boldface type** apply over the **Operating Temperature Range** .

			Typical	LM6162	LM6262	LM6362	
Symbol	Parameter	Conditions	(Note 7)	Limit	Limit	Limit	Units
				(Note 8)	(Note 8)	(Note 8)	
GBW	Gain-Bandwidth Product	f = 20 MHz	100	80	80	75	MHz
				55	65	65	min
		Supply = ±5V	70				MHz
SR	Slew Rate	A _V = +2 (Note 11)	300	200	200	200	V/µs
				180	180	180	min
		Supply = ±5V	200				V/µs
PBW	Power Bandwidth	V _{OUT} = 20 V _{PP}	4.5				MHz
t _s	Settling Time	10V step, to 0.1%	100				ns
		$A_V = -1$, $R_L = 2 k\Omega$					
φ _m	Phase Margin	A _V = +2	45				deg
	Differential Gain	NTSC, A _V = +2	<0.1				%
	Differential Phase	NTSC, A _V = +2	<0.1				deg
e _n	Input Noise Voltage	f = 10 kHz	10				nV/√Hz
i _n	Input Noise Current	f = 10 kHz	1.2				pA/√Hz

Note 1: Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device beyond its rated operating conditions.

Note 2: The ESD protection circuitry between the inputs will begin to conduct when the differential input voltage reaches 8V.

Note 3: a) In addition, the voltage between the V^+ pin and either input pin must not exceed 36V.

3V below the negative supply pin voltage, a substrate diode begins to conduct. Current through this pin must then be kept less than 20 mA to limit damage from self-heating.

Note 4: Although the output current is internally limited, continuous short-circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C.

Note 5: This value is the average voltage that the weakest pin combinations can withstand and still conform to the datasheet limits. The test circuit used consists of the human body model, 100 pF in series with 1500Ω .

Note 6: The typical thermal resistance, junction-to-ambient, of the molded plastic DIP (N package) is 105°C/W. For the molded plastic SO (M package), use 155°C/W. All numbers apply for packages soldered directly into a printed circuit board.

Note 7: Typical values are for $T_J = 25^{\circ}C$, and represent the most likely parametric norm.

Note 8: Limits are guaranteed, by testing or correlation.

AC Electrical Characteristics (Continued)

Note 9: Voltage Gain is the total output swing (20V) divided by the magnitude of the input signal required to produce that swing.

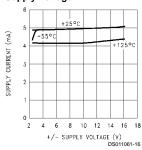
Note 10: For single-supply operation, the following conditions apply: V^+ = 5V, V^- = 0V, V_{CM} = 2.5V, V_{OUT} = 2.5V. Pin 1 and Pin 8 (V_{OS} Adjust pins) are each connected to pin 4 (V^-) to realize maximum output swing. This connection will increase the offset voltage.

Note 11: V_{IN} = 10V step. For ±5V supplies, V_{IN} = 1V step.

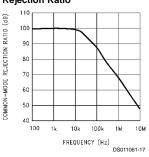
Note 12: A military RETS electrical test specification is available on request.

Typical Performance Characteristics R_L = 10 k Ω , T_A = 25°C unless otherwise noted

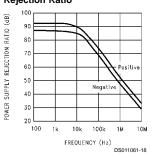
Supply Current vs Supply Voltage



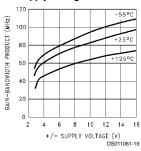
Common-Mode Rejection Ratio



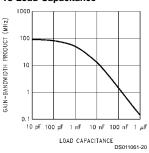
Power Supply Rejection Ratio



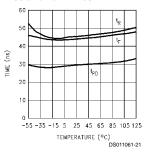
Gain-Bandwidth Product vs Supply Voltage



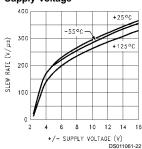
Gain-Bandwidth Product vs Load Capacitance



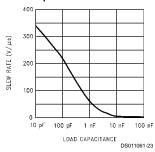
Propagation Delay, Rise and Fall Times



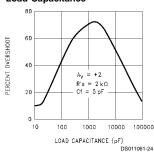
Slew Rate vs Supply Voltage



Slew Rate vs Load Capacitance

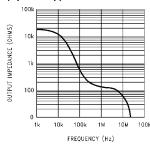


Overshoot vs Load Capacitance

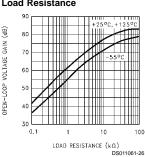


Typical Performance Characteristics R_L = 10 k Ω , T_A = 25°C unless otherwise noted (Continued)

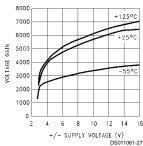
Output Impedance (Open-Loop)



Voltage Gain vs Load Resistance



Voltage Gain vs Supply Voltage

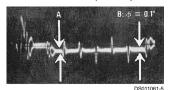


Differential Gain (Note)

Differential Gain (Note 13)

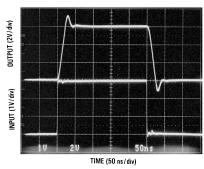


Differential Phase (Note 13)



Note 13: Differential gain and differential phase measured for four series LM6362 op amps configured with gain of +2 each, in series with a 1:16 attenuator and an LM6321 buffer. Error added by LM6321 is negligible. Test performed using Tektronix Type 520 NTSC test system.

Step Response; Av = +2

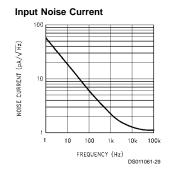


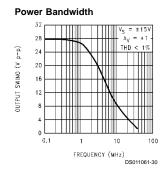
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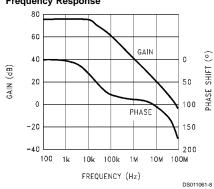


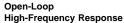
Input Noise Voltage NOISE VOLTAGE (nV/VHz) FREQUENCY (Hz)
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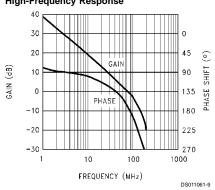




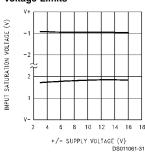




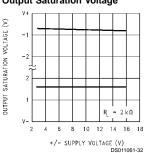




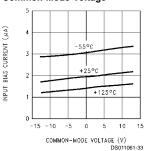
Common-Mode Input **Voltage Limits**



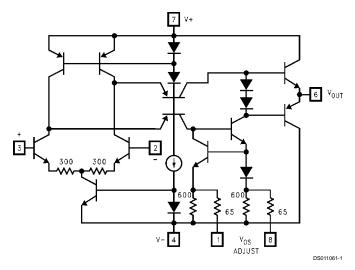
Output Saturation Voltage



Bias Current vs Common-Mode Voltage



Simplified Schematic



Application Tips

The LM6362 has been decompensated for a wider gain-bandwidth product than the LM6361. However, the LM6362 still offers stability at gains of 2 (and –1) or greater over the specified ranges of temperature, power supply voltage, and load. Since this decompensation involved reducing the emitter-degeneration resistors in the op amp's input stage, the DC precision has been increased in the form of lower offset voltage and higher open-loop gain.

Other op amps in this family include the LM6361, LM6364, and LM6365. If unity-gain stability is required, the LM6361 should be used. The LM6364 has been decompensated for operation at gains of 5 or more, with corresponding greater gain-bandwidth product (125 MHz, typical) and DC precision. The fully-uncompensated LM6365 offers gain-bandwidth product of 725 MHz, typical, and is stable for gains of 25 or more. All parts in this family, regardless of compensation, have the same high slew rate of 300 V/µs (typ).

The LM6362 is unusually tolerant of capacitive loads. Most op amps tend to oscillate when their load capacitance is greater than about 200 pF (in low-gain circuits). However, load capacitance on the LM6362 effectively increases its compensation capacitance, thus slowing the op amp's response and reducing its bandwidth. The compensation is not ideal, though, and ringing may occur in low-gain circuits with large capacitive loads.

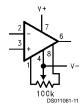
Power supply bypassing is not as critical for LM6362 as it is for other op amps in its speed class. However, bypassing will improve the stability and transient response of the LM6362, and is recommended for every design. 0.01 μ F to 0.1 μ F ceramic capacitors should be used (from each supply "rail" to ground); if the device is far away from its power supply source, an additional 2.2 μ F to 10 μ F of tantalum may be required for extra noise reduction.

Keep all leads short to reduce stray capacitance and lead inductance, and make sure ground paths are low-impedance, especially where heavier currents will be flowing. Stray capacitance in the circuit layout can cause signal coupling from one pin, input or lead to another, and can cause circuit gain to unintentionally vary with frequency.

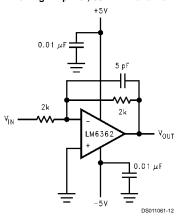
Breadboarded circuits will work best if they are built using generic PC boards with a good ground plane. If the op amps are used with sockets, as opposed to being soldered into the circuit, the additional input capacitance may degrade circuit frequency response. At low gains (+2 or -1), a feedback capacitor $C_{\rm f}$ from output to inverting input will compensate for the phase lag caused by capacitance at the inverting input. Typically, values from 2 pF to 5 pF work well; however, best results can be obtained by observing the amplifier pulse response and optimizing $C_{\rm f}$ for the particular layout.

Typical Applications

Offset Voltage Adjustment

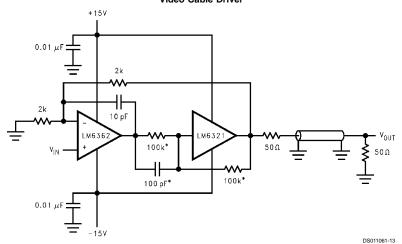


Inverting Amplifier, 30 MHz Bandwidth



Operation on ±15V supplies results in wider bandwidth, 50 MHz (typ).

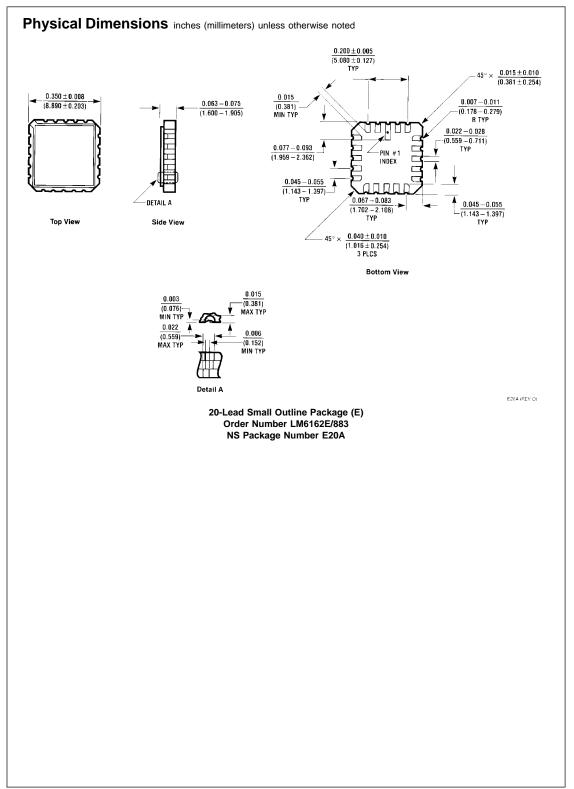
Video Cable Driver

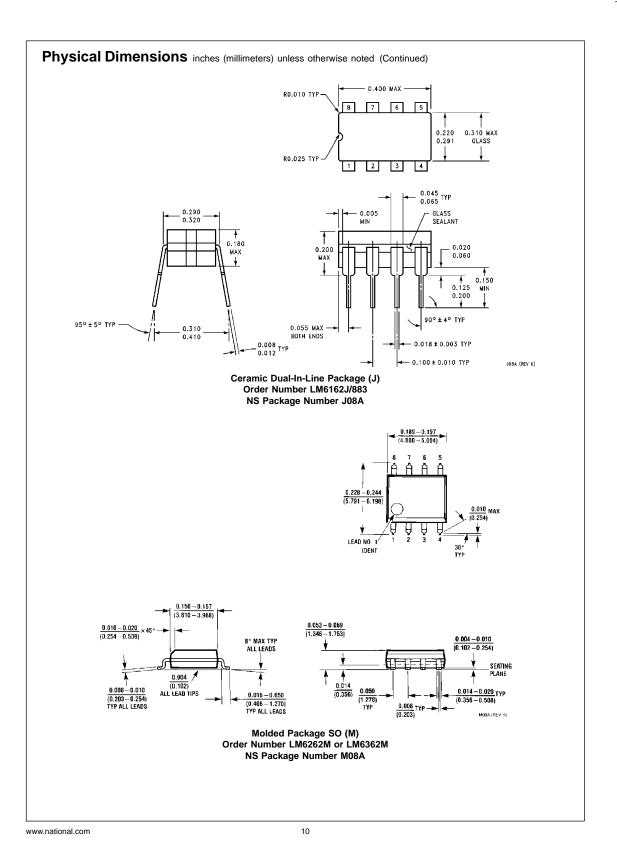


* Network required when operating on supply voltage over ±5V, for overvoltage protection of LM6321. If ±5V supplies are used, omit network and connect output of LM6362 directly to input of LM6321.

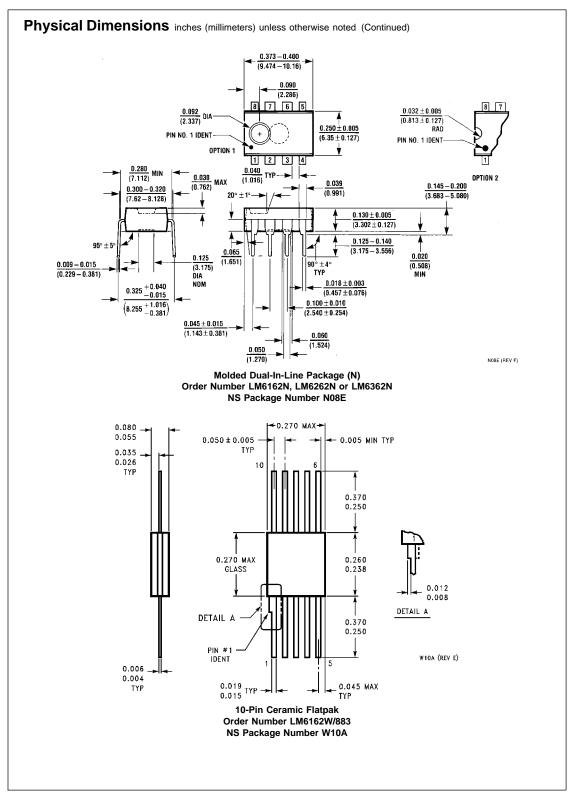
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