

## Galvanically isolated 4 A single gate driver for SiC MOSFETs



### Features

- High voltage rail up to 1700 V
- Driver current capability: 4 A sink/source @25 °C
- dV/dt transient immunity  $\pm 100$  V/ns in full temperature range
- Overall input-output propagation delay: 75 ns
- Separate sink and source option for easy gate driving configuration
- 4 A Miller CLAMP dedicated pin option
- UVLO function
- Gate driving voltage up to 26 V
- 3.3 V, 5 V TTL/CMOS inputs with hysteresis
- Temperature shutdown protection
- Standby function
- 4.8 kV<sub>PK</sub> isolation
- Narrow body SO8

### Application

- Motor driver for home appliances, factory automation, industrial drives and fans.
- 600/1200 V inverters
- Battery chargers
- Induction heating
- Welding
- UPS
- Power supply units
- DC-DC converters
- Power Factor Correction

### Description

The **STGAP2SICSN** is a single gate driver which provides galvanic isolation between the gate driving channel and the low voltage control and interface circuitry.

The gate driver is characterized by 4 A capability and rail-to-rail outputs, making the device also suitable for mid and high power applications such as power conversion and motor driver inverters in industrial applications. The device is available in two different configurations. The configuration with separated output pins allows to independently optimize turn-on and turn-off by using dedicated gate resistors. The configuration featuring single output pin and Miller CLAMP function prevents gate spikes during fast commutations in half-bridge topologies. Both configurations provide high flexibility and bill of material reduction for external components.

The device integrates protection functions: UVLO with optimized value for SiC MOSFETs and thermal shutdown are included to easily design high reliability systems. Dual input pins allow choosing the control signal polarity and also implementing HW interlocking protection in order to avoid cross-conduction in case of controller malfunction. The input to output propagation delay results are contained within 75 ns, providing high PWM control accuracy. A standby mode is available in order to reduce idle power consumption.

#### Product status link

[STGAP2SICSN](#)

#### Product label



# 1 Block diagram

Figure 1. Block diagram - Single output and Miller Clamp configuration

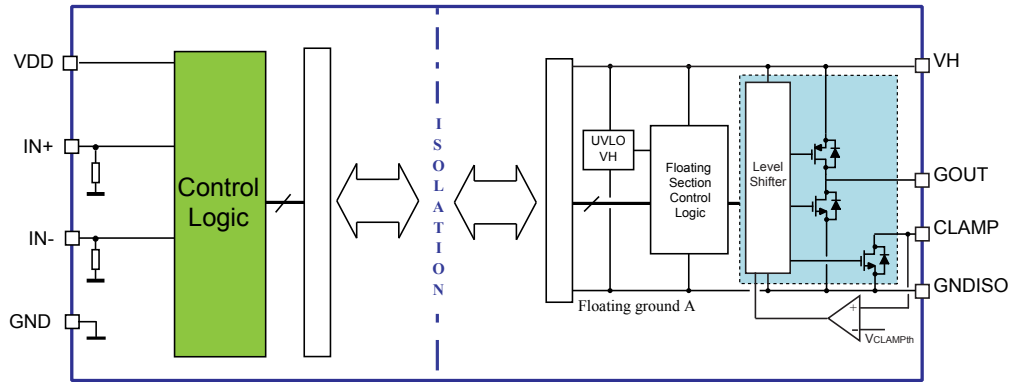
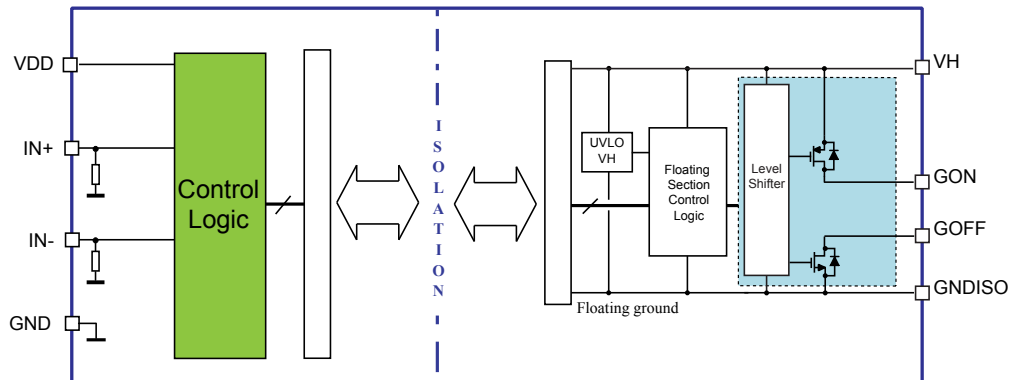


Figure 2. Block diagram - Separate outputs configuration



## 2 Pin description and connection diagram

Figure 3. Pin connection (top view), Separated outputs option

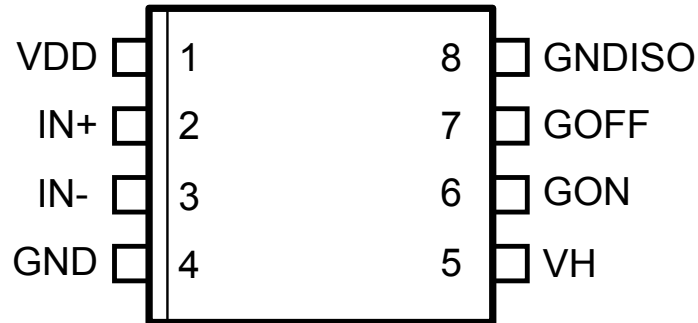


Figure 4. Pin connection (top view), Single output and Miller CLAMP option

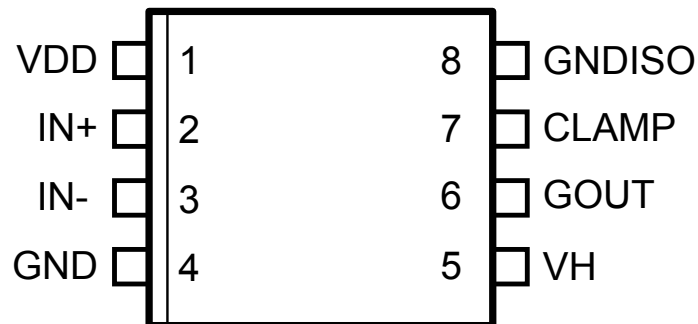


Table 1. Pin Description

Pin #		Pin name	Type	Function
Figure 3	Figure 4			
1	1	VDD	Power supply	Driver logic supply voltage.
2	2	IN+	Logic input	Driver logic input, active high.
3	3	$\overline{\text{IN}}^-$	Logic input	Driver logic input, active low.
4	4	GND	Power supply	Driver logic ground.
5	5	VH	Power supply	Gate driving positive voltage supply.
	6	GOUT	Analog output	Sink/Source output.
	7	CLAMP	Analog output	Active Miller Clamp.
6		GON	Analog output	Source output.
7		GOFF	Analog output	Sink output.
8	8	GNDISO	Power supply	Gate driving Isolated ground.

### 3 Electrical data

#### 3.1 Absolute maximum ratings

**Table 2. Absolute maximum ratings**

Symbol	Parameter	Test condition	Min.	Max.	Unit
VDD	Logic supply voltage vs. GND		-0.3	6.5	V
V <sub>LOGIC</sub>	Logic pins voltage vs. GND		-0.3	6.5	V
VH	Positive supply voltage (VH vs. GNDISO)		-0.3	28	V
V <sub>OUT</sub>	Voltage on gate driver outputs (GON, GOFF, CLAMP vs. GNDISO)		-0.3	VH + 0.3	V
T <sub>J</sub>	Junction temperature		-40	150	°C
T <sub>S</sub>	Storage temperature		-50	150	°C
ESD	HBM (human body model)			2	kV

#### 3.2 Thermal data

**Table 3. Thermal data**

Symbol	Parameter	Package	Value	Unit
R <sub>th(JA)</sub>	Thermal resistance junction to ambient	SO-8	123	°C/W

#### 3.3 Recommended operating conditions

**Table 4. Recommended operating conditions**

Symbol	Parameter	Test conditions	Min.	Max.	Unit
VDD	Logic supply voltage vs. GND		3.1	5.5	V
V <sub>LOGIC</sub>	Logic pins voltage vs. GND		0	5.5	V
VH	Positive supply voltage (VH vs. GNDISO)		Max(V <sub>Hon</sub> )	26	V
V <sub>ISO-OP</sub>	Input to output operative voltage (GND to GNDISO)	DC or peak	-1700	+1700	V
F <sub>SW</sub>	Maximum switching frequency. <sup>(1)</sup>			1	MHz
t <sub>OUT</sub>	Output Pulse width		100		ns
T <sub>J</sub>	Operating Junction Temperature		-40	125	°C

1. Actual limit depends on power dissipation and T<sub>J</sub>

## 4 Electrical characteristics

**Table 5. Electrical characteristics**

Symbol	Pin	Parameter	Test conditions	Min.	Typ.	Max.	Unit
<b>Dynamic characteristics</b>							
$t_{Don}$	IN+, IN-	Input to output propagation delay ON		50	75	90	ns
$t_{Doff}$	IN+, IN-	Input to output propagation delay OFF		50	75	90	ns
$t_r$		Rise time	$C_L = 4.7$ nF see Figure 12		30		ns
$t_f$		Fall time			30		ns
PWD		Pulse Width Distortion   $t_{Don} - t_{Doff}$				20	ns
$t_{deglitch}$	IN+, IN-	Inputs deglitch filter			20	40	ns
CMTI <sup>(1)</sup>		Common-mode transient immunity, $ dV_{ISO}/dt $	$V_{CM} = 1500$ V, see Figure 13	100			V/ns
<b>Supply voltage</b>							
$V_{H_{on}}$	VH	VH UVLO turn-on threshold		14.6	15.5	16.4	V
$V_{H_{off}}$	VH	VH UVLO turn-off threshold		13.9	14.8	15.7	V
$V_{H_{hyst}}$	VH	VH UVLO hysteresis		600	750	950	mV
$I_{QH_{U}}$	VH	VH undervoltage quiescent supply current	$V_H = 7.0$ V		1.3	1.8	mA
$I_{QH}$	VH	VH quiescent supply current			1.3	1.8	mA
$I_{QH_{SBY}}$	VH	Standby VH quiescent supply current	Standby mode		400	550	$\mu$ A
SafeClp	GOUT / GOFF	GOFF active clamp	$I_{GOFF} = 0.2$ A; VH floating		2	2.3	V
$I_{QDD}$	VDD	VDD quiescent supply current			1.0	1.3	mA
$I_{QDD_{SBY}}$	VDD	Standby VDD quiescent supply current	Standby mode		40	65	$\mu$ A
<b>Logic inputs</b>							
$V_{il}$	IN+, IN-	Low-level logic threshold voltage		$0.29 \cdot V_{DD}$	$0.33 \cdot V_{DD}$	$0.37 \cdot V_{DD}$	V
$V_{ih}$	IN+, IN-	High-level logic threshold voltage		$0.62 \cdot V_{DD}$	$0.66 \cdot V_{DD}$	$0.70 \cdot V_{DD}$	V
$I_{INh}$	IN+, IN-	INx logic "1" input bias current	$INx = 5$ V	33	50	70	$\mu$ A
$I_{INl}$	IN+, IN-	INx logic "0" input bias current	$INx = GND$			1	$\mu$ A
$R_{pd}$	IN+, IN-	Inputs pull-down resistors	$INx = 5$ V	70	100	150	k $\Omega$
<b>Driver buffer section</b>							
$I_{GON}$	GOUT / GON	Source short-circuit current	$T_J = 25$ °C		4		A

Symbol	Pin	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{GON}$	GOUT / GON	Source short-circuit current	$T_J = -40 \div +125\text{ }^\circ\text{C}^{(1)}$	3		5	
$V_{GONH}$	GOUT / GON	Source output high-level voltage	$I_{GON} = 100\text{ mA}$	VH-0.15	VH-0.125		V
$R_{GON}$	GOUT / GON	Source $R_{DS\_ON}$	$I_{GON} = 100\text{ mA}$		1.25	1.5	$\Omega$
$I_{GOFF}$	GOUT / GOFF	Sink short-circuit current	$T_J = 25\text{ }^\circ\text{C}$		4		A
			$T_J = -40 \div +125\text{ }^\circ\text{C}^{(1)}$	3		5.5	
$V_{GOFFL}$	GOUT / GOFF	Sink output low-level voltage	$I_{GOFF} = 100\text{ mA}$		110	120	mV
$R_{GOFF}$	GOUT / GOFF	Sink $R_{DS\_ON}$	$I_{GOFF} = 100\text{ mA}$		1.1	1.2	$\Omega$
<b>Miller clamp function</b>							
$V_{CLAMPth}$	CLAMP	CLAMP voltage threshold	$V_{CLAMP}$ vs. GNDISO	1.3	2	2.6	V
$I_{CLAMP}$	CLAMP	CLAMP short-circuit current	$V_{CLAMP} = 15\text{ V}$				A
			$T_J = 25\text{ }^\circ\text{C}$		4		
			$T_J = -40 \div +125\text{ }^\circ\text{C}^{(1)}$	2		5	
$V_{CLAMP\_L}$	CLAMP	CLAMP low-level output voltage	$I_{CLAMP} = 100\text{ mA}$		96	115	mV
$R_{CLAMP}$	CLAMP	CLAMP $R_{DS\_ON}$	$I_{CLAMP} = 100\text{ mA}$		0.96	1.15	$\Omega$
<b>Overtemperature protection</b>							
$T_{SD}$		Shutdown temperature <sup>(1)</sup>		170			$^\circ\text{C}$
$T_{hys}$		Temperature hysteresis <sup>(1)</sup>			20		$^\circ\text{C}$
<b>Standby</b>							
$t_{STBY}$		Standby time	See Section 5.7	200	280	500	$\mu\text{s}$
$t_{WUP}$		Wake-up time	See Section 5.7	10	20	35	$\mu\text{s}$
$t_{awake}$		Wake-up delay	See Section 5.7	90	140	200	$\mu\text{s}$
$t_{stbyfilt}$		Standby filter	See Section 5.7	200	280	800	ns

1. Characterization data, not tested in production.

**Table 6. Isolation related package specifications**

Parameter	Symbol	Value	Unit	Conditions
Clearance (Minimum External Air Gap )	CLR	4	mm	Measured from input terminals to output terminals, shortest distance through air
Creepage (*) (Minimum External Tracking)	CPG	4	mm	Measured from input terminals to output terminals, shortest distance path along body
Comparative Tracking Index (Tracking Resistance)	CTI	$\geq 400$	V	DIN IEC 112/VDE 0303 Part 1
Isolation Group		II		Material Group (DIN VDE 0110, 1/89, Table 1)

**Table 7. Isolation characteristics**

Parameter	Symbol	Test conditions	Characteristic	Unit
Input to Output test voltage In accordance with VDE 0884-11	$V_{PR}$	Method a, Type test	2720	$V_{PEAK}$
		$V_{PR} = 2720, t_m = 10 \text{ s}$		
		Partial discharge < 5 pC		
		Method b1, 100 % Production test	3200	$V_{PEAK}$
		$V_{PR} = 3200, t_m = 1 \text{ s}$		
Partial discharge < 5 pC				
Transient Overvoltage (Highest Allowable Overvoltage)	$V_{IOTM}$	$t_{ini} = 60 \text{ s}$ , Type test	4800	$V_{PEAK}$
Maximum Surge Test Voltage	$V_{IOSM}$	Type test	4800	$V_{PEAK}$
Isolation Resistance	$R_{IO}$	$V_{IO} = 500 \text{ V}$ , Type test	$>10^9$	$\Omega$

**Table 8. UL 1577 Tests**

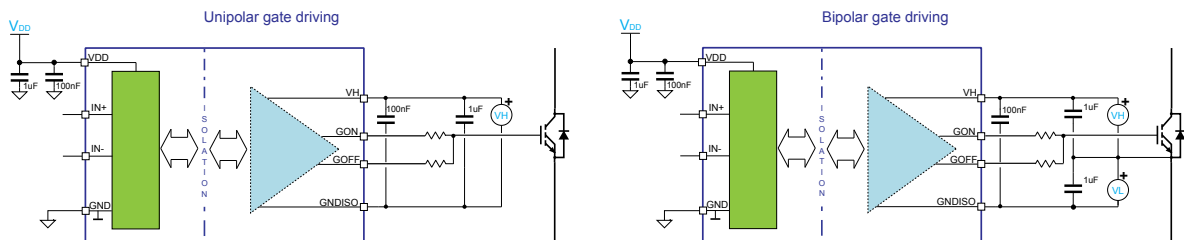
Description	Symbol	Characteristic	Unit
Isolation Withstand Voltage, 1min (Type test)	$V_{ISO}$	2828/4000	$V_{rms}/ P_{EAK}$
Isolation Voltage, 1sec (100% production)	$V_{ISOtest}$	3394/4800	$V_{rms}/ P_{EAK}$

## 5 Functional Description

### 5.1 Gate driving power supply and UVLO

The STGAP2SiCSN is a flexible and compact gate driver with 4 A output current and rail-to-rail outputs. The device allows implementation of either unipolar or bipolar gate driving.

**Figure 5. Power supply configuration for unipolar and bipolar gate driving**



Undervoltage protection is available on VH supply pin. A fixed hysteresis sets the turn-off threshold, thus avoiding intermittent operation.

When VH voltage goes below the  $VH_{off}$  threshold, the output buffer goes into “safe state”. When VH voltage reaches the  $VH_{on}$  threshold, the device returns to normal operation and sets the output according to actual input pins status.

The VDD and VH supply pins must be properly filtered with local bypass capacitors. The use of capacitors with different values in parallel provides both local storage for impulsive current supply and high-frequency filtering. The best filtering is obtained by using low-ESR SMT ceramic capacitors, which are therefore recommended. A 100 nF ceramic capacitor must be placed as close as possible to each supply pin, and a second bypass capacitor with a value in the range between 1  $\mu$ F and 10  $\mu$ F should be placed close to it.

### 5.2 Power-up, power-down and ‘safe state’

The following conditions define the “safe state”:

- GOFF = ON state;
- GON = High Impedance;
- CLAMP = ON state (for STGAP2SiCSNC);

Such conditions are maintained at power-up of the isolated side ( $VH < VH_{on}$ ) and during whole device power-down phase ( $VH < VH_{off}$ ), regardless of the value of the input pins.

The device integrates a structure which clamps the driver output to a voltage not higher than SafeClp when VH voltage is not high enough to actively turn the internal GOFF MOSFET on. If VH positive supply pin is floating or not supplied, the GOFF pin is therefore clamped to a voltage smaller than SafeClp.

If the supply voltage VDD of the control section of the device is not supplied, the output is put into *safe state*, and remains in such condition until the VDD voltage returns within operative conditions.

After power-up of both isolated and low voltage side, the device output state depends on the input pins’ status.



### 5.3 Control inputs

The device is controlled through the IN+ and IN- logic inputs, in accordance with the truth table described in Table 9.

**Table 9. Inputs truth table (applicable when device is not in UVLO or "safe state")**

Input pins		Output pins	
IN+	IN-	GON	GOFF
L	L	OFF	ON
H	L	ON	OFF
L	H	OFF	ON
H	H	OFF	ON

A deglitch filter allows input signals with duration shorter than  $t_{\text{deglitch}}$  to be ignored, thereby preventing noise spikes potentially present in the application from generating unwanted commutations.

### 5.4 Miller Clamp function

The Miller Clamp function allows the control of the Miller current during the power stage switching in half-bridge configurations. When the external power transistor is in the OFF state, the driver operates to avoid the induced turn-on phenomenon that may occur when the other switch in the same leg is being turned on, due to the  $C_{GD}$  capacitance.

During the turn-off period the gate of the external switch is monitored through the CLAMP pin. The CLAMP switch is activated when gate voltage goes below the voltage threshold,  $V_{\text{CLAMPth}}$ , thus creating a low impedance path between the switch gate and the GNDISO pin.

### 5.5 Watchdog

The isolated HV side has a watchdog function in order to identify when it is not able to communicate with LV side, for example because the VDD of the LV side is not supplied. In this case the output of the driver is forced in "safe state" until communication link is properly established again.

### 5.6 Thermal shutdown protection

The device provides a thermal shutdown protection. When junction temperature reaches the  $T_{SD}$  temperature threshold, the device is forced into "safe state". The device operation is restored as soon as the junction temperature is lower than  $T_{SD} - T_{\text{hys}}$ .

## 5.7 Standby function

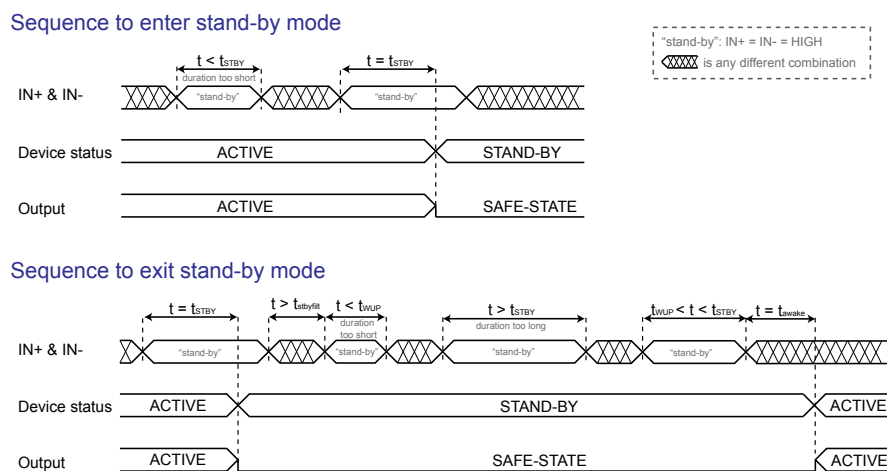
In order to reduce the power consumption of both control interface and gate driving sides the device can be put in standby mode. In standby mode the quiescent current from VDD and VH supply pins is reduced to  $I_{QDDBY}$  and  $I_{QHSDY}$  respectively, and the output remains in 'safe state' (the output is actively forced low).

The way to enter standby is to keep both IN+ and IN- high ("standby" value) for a time longer than  $t_{STBY}$ . During standby the inputs can change from the "standby" value.

To exit standby, IN+ and IN- must be put in any combination different from the "standby" value for a time longer than  $t_{stbyfilt}$ , and then in the "standby" value for a time  $t$  such that  $t_{WUP} < t < t_{STBY}$ .

When the input configuration is changed from the "standby" value the output is enabled and set according to inputs state after a time  $t_{awake}$ .

Figure 6. Standby state sequences



## 6 Typical application diagram

Figure 7. Typical application diagram - Separated outputs

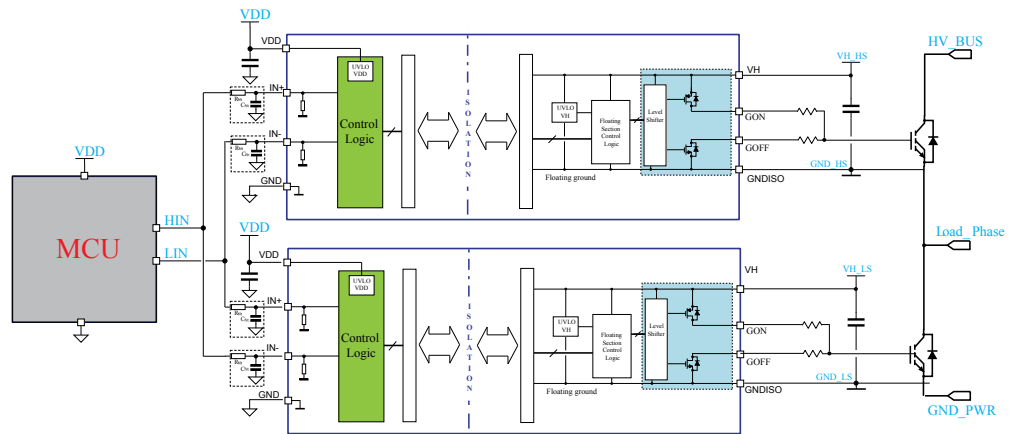
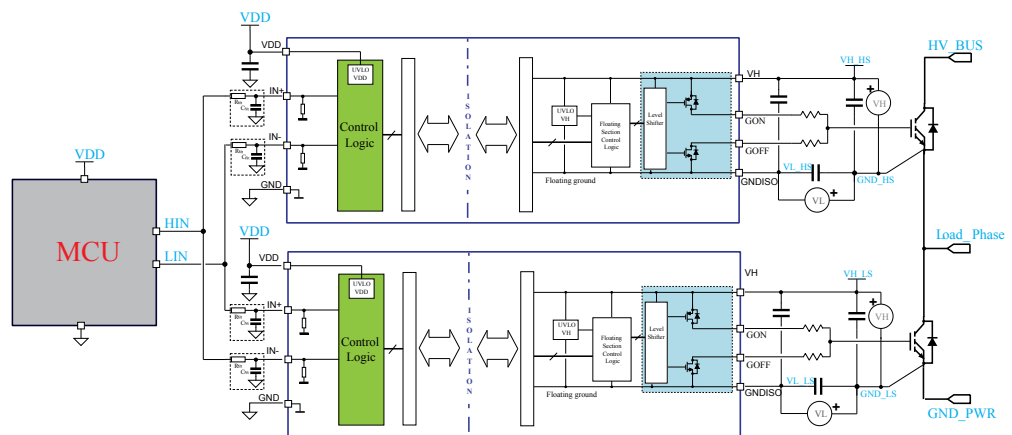
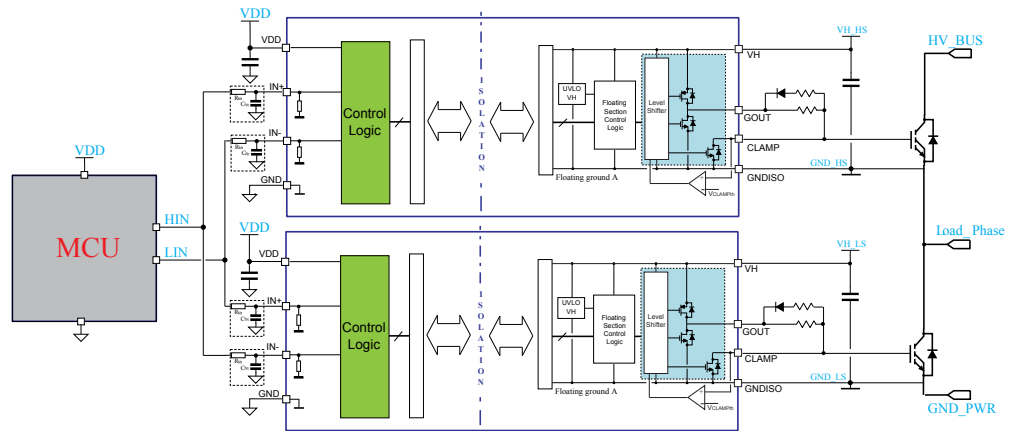


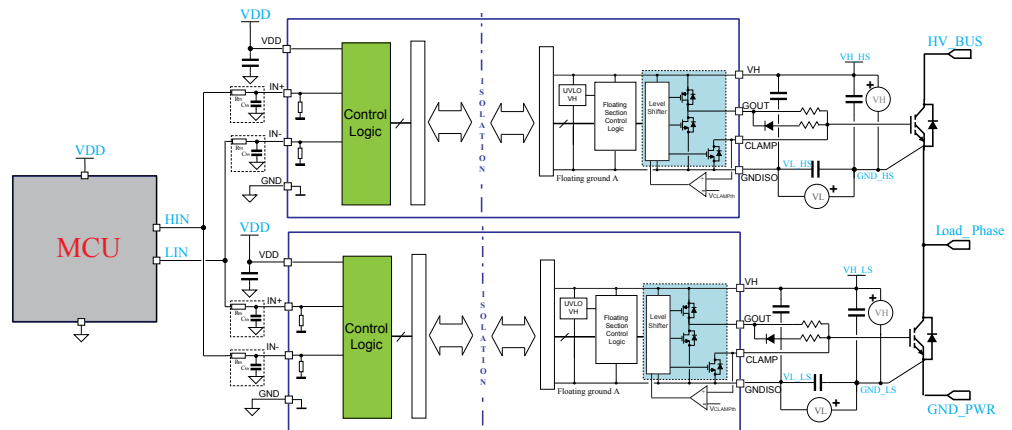
Figure 8. Typical application diagram - Separated outputs and negative gate driving



**Figure 9. Typical application diagram - Miller Clamp**



**Figure 10. Typical application diagram - Miller Clamp and negative gate driving**



## 7 Layout

### 7.1 Layout guidelines and considerations

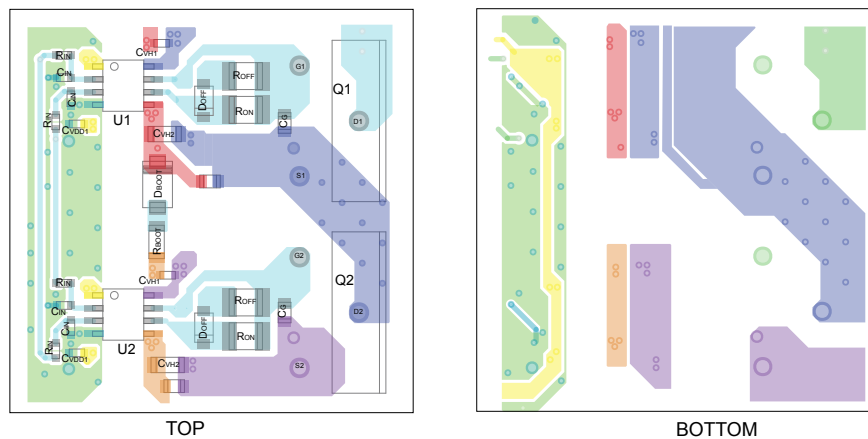
In order to optimize the PCB layout, the following considerations should be taken into account:

- SMT ceramic capacitors (or different types of low-ESR and low-ESL capacitors) must be placed close to each supply rail pin. A 100 nF capacitor must be placed between VDD and GND and between VH and GNDISO, as close as possible to device pins, in order to filter high-frequency noise and spikes. In order to provide local storage for pulsed current a second capacitor with value in the range between 1  $\mu$ F and 10  $\mu$ F should also be placed close to the supply pins.
- As a good practice, it is suggested to add filtering capacitors close to logic inputs of the device (IN+, IN-), in particular for fast switching or noisy applications.
- The power transistors must be placed as close as possible to the gate driver, so to minimize the gate loop area and inductance that might lead to noise or ringing.
- To avoid degradation of the isolation between the primary and secondary side of the driver, there should not be any trace or conductive area below the driver.
- If the system has multiple layers, it is recommended to connect the VH and GNDISO pins to internal ground or power planes through multiple vias of adequate size. These vias should be located close to the IC pins to maximize thermal conductivity.

### 7.2 Layout example

An example of the STGAP2SiCSN half-bridge suggested PCB layout with main signals highlighted by different colors is shown in Figure 11. It is recommended to follow this example for proper positioning and connection of filtering capacitors.

**Figure 11. Half-bridge suggested PCB layout**



## 8 Testing and characterization information

Figure 12. Timings definition

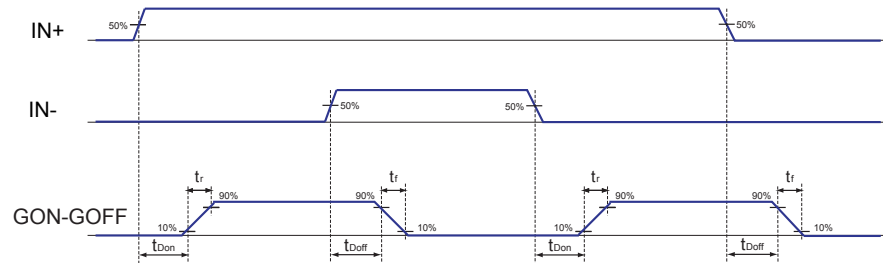
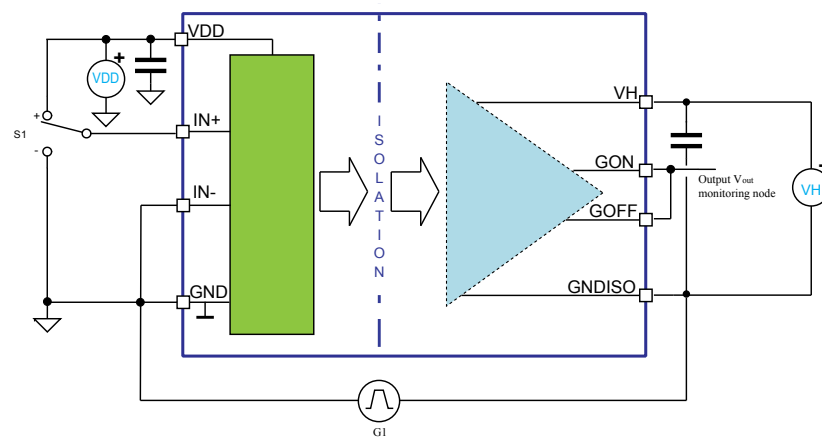


Figure 13. CMTI test circuit



## 9 Package information

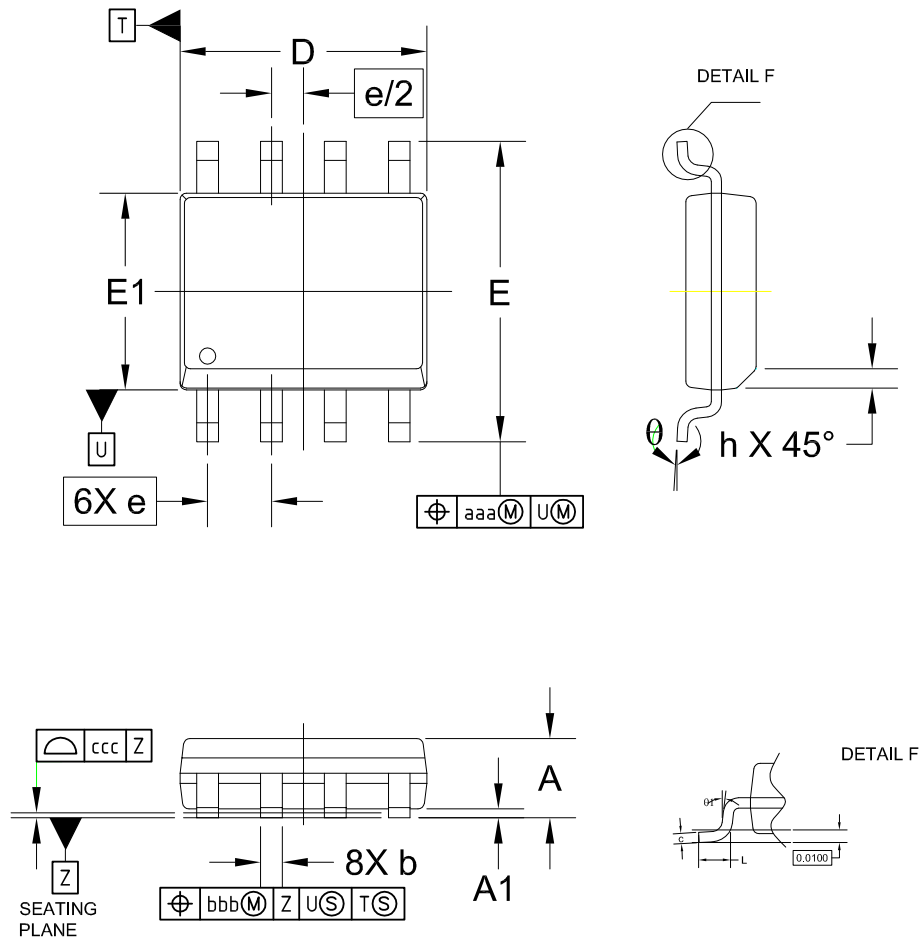
In order to meet environmental requirements, ST offers these devices in different grades of **ECOPACK** packages, depending on their level of environmental compliance. ECOPACK specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK is an ST trademark.

### 9.1 SO-8 package information

Dim.	mm			NOTES
	Min.	Typ.	Max.	
A	1.35	-	1.75	
A1	0.10	-	0.25	
b	0.35	-	0.49	
c	0.19	-	0.25	
D <sup>(1)</sup>	4.8	-	5	
E1	3.8	3.9	4	
E	5.8	6	6.2	
e	1.27 BSC			
L	0.4	-	1.25	
h	0.25	-	0.50	
θ	0°		7°	
Θ1	2°		12°	
aaa	0.25			
bbb	0.25			
ccc	0.1			

1. Dimension "D" does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions or gate burrs shall not exceed 0.15 mm per side.

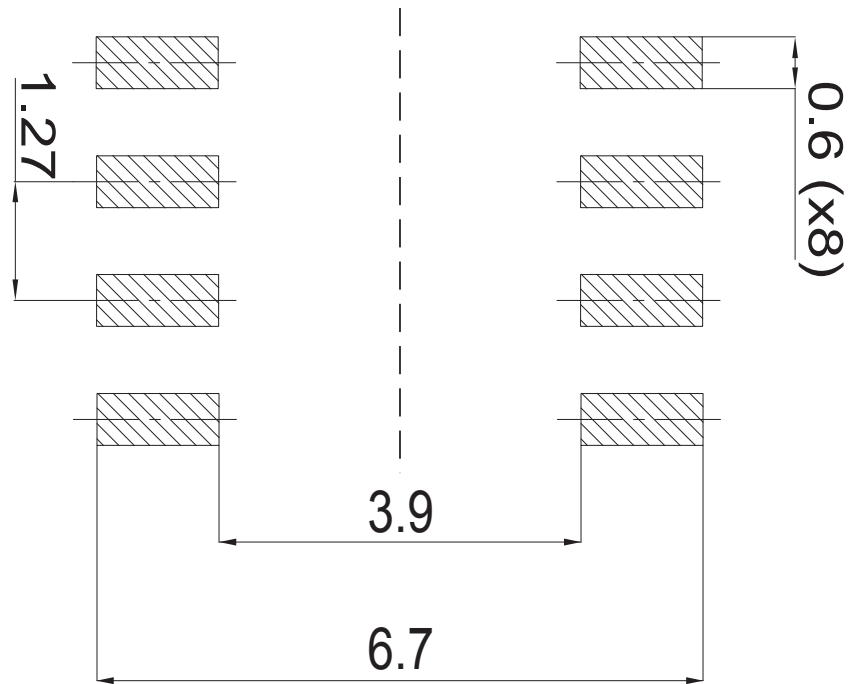
Figure 14. SO-8 mechanical data





## 9.2 SO-8 Suggested land pattern

Figure 15. SO-8 suggested land pattern



## 10 Ordering information

Table 10. Device summary

Order code	Output configuration	Package	Package marking	Packaging
STGAP2SiCSN	GON-GOFF	SO-8	GAP2ISN	Tape and Reel
STGAP2SiCSNC	GOUT-CLAMP	SO-8	GAP2ISCN	Tape and Reel

## Revision history

**Table 11. Document revision history**

Date	Version	Changes
13-Aug-2021	1	Initial release.

## Contents

<b>1</b>	<b>Block diagram</b>	<b>2</b>
<b>2</b>	<b>Pin description and connection diagram</b>	<b>3</b>
<b>3</b>	<b>Electrical data</b>	<b>4</b>
3.1	Absolute maximum ratings	4
3.2	Thermal data	4
3.3	Recommended operating conditions	4
<b>4</b>	<b>Electrical characteristics</b>	<b>5</b>
<b>5</b>	<b>Functional Description</b>	<b>8</b>
5.1	Gate driving power supply and UVLO	8
5.2	Power-up, power-down and 'safe state'	8
5.3	Control inputs	9
5.4	Miller Clamp function	9
5.5	Watchdog	9
5.6	Thermal shutdown protection	9
5.7	Standby function	10
<b>6</b>	<b>Typical application diagram</b>	<b>11</b>
<b>7</b>	<b>Layout</b>	<b>13</b>
7.1	Layout guidelines and considerations	13
7.2	Layout example	13
<b>8</b>	<b>Testing and characterization information</b>	<b>14</b>
<b>9</b>	<b>Package information</b>	<b>15</b>
9.1	[Package name] package information	15
9.2	SO-8 Suggested land pattern	17
<b>10</b>	<b>Ordering information</b>	<b>18</b>
	<b>Revision history</b>	<b>19</b>
	<b>Contents</b>	<b>20</b>
	<b>List of tables</b>	<b>21</b>
	<b>List of figures</b>	<b>22</b>

## List of tables

<b>Table 1.</b>	Pin Description . . . . .	3
<b>Table 2.</b>	Absolute maximum ratings . . . . .	4
<b>Table 3.</b>	Thermal data . . . . .	4
<b>Table 4.</b>	Recommended operating conditions . . . . .	4
<b>Table 5.</b>	Electrical characteristics . . . . .	5
<b>Table 6.</b>	Isolation related package specifications . . . . .	6
<b>Table 7.</b>	Isolation characteristics . . . . .	7
<b>Table 8.</b>	UL 1577 Tests . . . . .	7
<b>Table 9.</b>	Inputs truth table (applicable when device is not in UVLO or "safe state") . . . . .	9
<b>Table 10.</b>	Device summary . . . . .	18
<b>Table 11.</b>	Document revision history . . . . .	19

## List of figures

<b>Figure 1.</b>	Block diagram - Single output and Miller Clamp configuration . . . . .	2
<b>Figure 2.</b>	Block diagram - Separate outputs configuration . . . . .	2
<b>Figure 3.</b>	Pin connection (top view), Separated outputs option . . . . .	3
<b>Figure 4.</b>	Pin connection (top view), Single output and Miller CLAMP option . . . . .	3
<b>Figure 5.</b>	Power supply configuration for unipolar and bipolar gate driving . . . . .	8
<b>Figure 6.</b>	Standby state sequences . . . . .	10
<b>Figure 7.</b>	Typical application diagram - Separated outputs . . . . .	11
<b>Figure 8.</b>	Typical application diagram - Separated outputs and negative gate driving . . . . .	11
<b>Figure 9.</b>	Typical application diagram - Miller Clamp . . . . .	12
<b>Figure 10.</b>	Typical application diagram - Miller Clamp and negative gate driving . . . . .	12
<b>Figure 11.</b>	Half-bridge suggested PCB layout . . . . .	13
<b>Figure 12.</b>	Timings definition . . . . .	14
<b>Figure 13.</b>	CMTI test circuit . . . . .	14
<b>Figure 14.</b>	SO-8 mechanical data . . . . .	16
<b>Figure 15.</b>	SO-8 suggested land pattern . . . . .	17

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