

### Product Change Notification / SYST-29BFHE416

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30-Mar-2022

### **Product Category:**

**General Purpose LED Drivers** 

### **PCN Type:**

**Document Change** 

## **Notification Subject:**

Data Sheet - HV9961 LED Driver with Average-Current Mode Constant-Current Control Data Sheet Document Revision

### **Affected CPNs:**

SYST-29BFHE416\_Affected\_CPN\_03302022.pdf SYST-29BFHE416\_Affected\_CPN\_03302022.csv

### **Notification Text:**

SYST-29BFHE416

Microchip has released a new Product Documents for the HV9961 LED Driver with Average-Current Mode Constant-Current Control Data Sheet of devices. If you are using one of these devices please read the document located at HV9961 LED Driver with Average-Current Mode Constant-Current Control Data Sheet.

**Notification Status: Final** 

#### **Description of Change:**

- 1) Changed package drawings.
- 2) Updated Section "Electrical Characteristics".
- 3) Updated Section "Temperature Specifications".
- 4) Minor format changes throughout.

Impacts to Data Sheet: See above details.

Reason for Change: To Improve Productivity

Change Implementation Status: Complete

**Date Document Changes Effective:** 30 March 2022

NOTE: Please be advised that this is a change to the document only the product has not been changed.

Markings to Distinguish Revised from Unrevised Devices: N/A

#### **Attachments:**

HV9961 LED Driver with Average-Current Mode Constant-Current Control Data Sheet

Please contact your local Microchip sales office with questions or concerns regarding this notification.

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If you wish to <u>change your PCN profile</u>, <u>including opt out</u>, please go to the <u>PCN home page</u> select login and sign into your myMicrochip account. Select a profile option from the left navigation bar and make the applicable selections.

SYST-29BFHE416 - Data Sheet - HV9961 LED Driver with Average-Current Mode Constant-Current Control Data Sheet Document Revision

Affected Catalog Part Numbers (CPN)

HV9961LG-G HV9961NG-G HV9961NG-G-M901 HV9961NG-G-M934

Date: Tuesday, March 29, 2022

# **HV9961**

## LED Driver with Average-Current Mode Constant-Current Control

#### **Features**

- · Fast Average Current Control
- · Programmable Constant Off-time Switching
- · Linear Dimming Input
- · PWM Dimming Input
- · Output Short-circuit Protection with Skip Mode
- –40°C to +125°C Ambient Operating Temperature
- · Pin-compatible with HV9910B

#### **Applications**

- DC/DC or AC/DC LED Driver Applications
- · LED Backlight Driver for LCD Displays
- · General Purpose Constant-current Source
- · LED Signage and Displays
- · Architectural and Decorative LED Lighting
- · LED Street Lighting

#### **General Description**

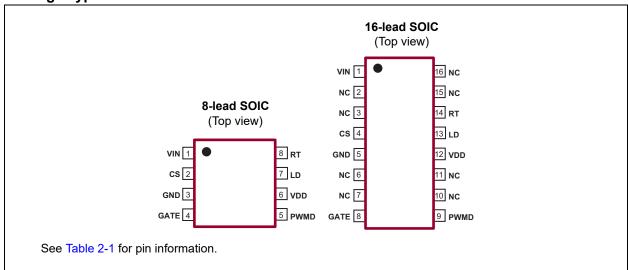
The HV9961 is an Average-Current mode constant-current control LED driver IC operating in a constant Off-time mode. Unlike the HV9910B, this control IC does not produce a peak-to-average error. This greatly improves accuracy as well as the line and load regulations of the LED current without any need for loop compensation or high-side current sensing. Its output LED current accuracy is ±3%.

The IC is equipped with a current limit comparator for Hiccup mode output short-circuit protection.

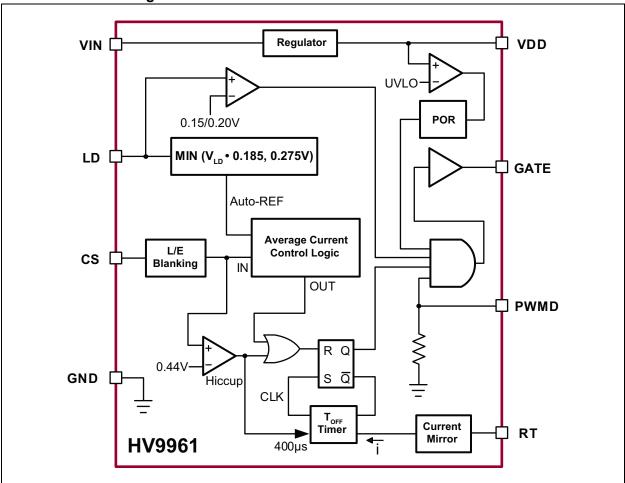
The HV9961 can be powered from an 8V–450V supply. It has a PWM dimming input that accepts an external control TTL-compatible signal. In addition, the output current can be programmed by an internal 275 mV reference or controlled externally through a 0V–1.5V linear dimming input.

The HV9961 is pin-to-pin compatible with HV9910B, and it can be used as a drop-in replacement for many applications to improve LED current accuracy and regulation.

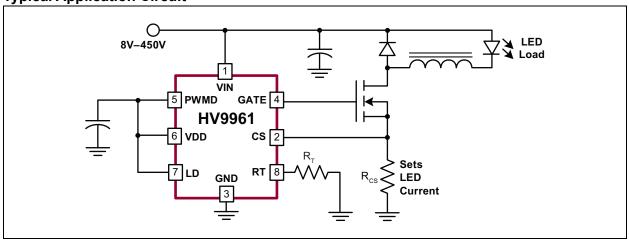
#### **Package Types**



### **Functional Block Diagram**



### **Typical Application Circuit**



#### 1.0 ELECTRICAL CHARACTERISTICS

#### **Absolute Maximum Ratings†**

V <sub>IN</sub> to GND	
V <sub>DD</sub> to GND	
CS, LD, PWMD, Gate, RT to GND	
Continuous Power Dissipation (T <sub>A</sub> = +25°C):	
8-lead SOIC	650 mW
16-lead SOIC	1000 mW

**† Notice:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only, and functional operation of the device at those or any other conditions above those indicated in the operational sections of this specification is not intended. Exposure to maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

<b>Electrical Specifications</b> : $T_A = 25$ °C, $V_{IN} = 12$ V, $V_{LD} = V_{DD}$ , and $V_{PWMD} = V_{DD}$ unless otherwise specified.							
Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions	
INPUT	•						
Input DC Supply Voltage Range	V <sub>INDC</sub>	8	_	450	V	DC input voltage (Note 1 and Note 2)	
Shutdown Mode Supply Current	I <sub>INSD</sub>	_	0.5	1	mA	Pin PWMD connected to GND (Note 2)	
INTERNAL REGULATOR							
Internally Regulated Voltage	V <sub>DD</sub>	7.25	7.5	7.75	>	$V_{IN}$ = 8V, $I_{DD(EXT)}$ = 0 mA, 500 pF at gate, $R_T$ = 226 kΩ	
Line Regulation of V <sub>DD</sub>	$\Delta V_{DD, \ line}$	0	_	1	>	$V_{IN} = 8V-450V,$ $I_{DD(EXT)} = 0 \text{ mA},$ $500 \text{ pF at gate, R}_{T} = 226 \text{ k}\Omega$	
Load Regulation of V <sub>DD</sub>	$\Delta V_{DD, load}$	0	_	100	mV	$I_{DD(EXT)} = 0 \text{ mA-1 mA},$ 500 pF at gate, $R_T = 226 \text{ k}\Omega$	
V <sub>DD</sub> Undervoltage Lockout Upper Threshold	V <sub>UVLO</sub>	6.45	6.7	6.95	٧	V <sub>IN</sub> rising (Note 2)	
V <sub>DD</sub> Undervoltage Lockout Hysteresis	ΔV <sub>UVLO</sub>	_	500	_	mV	V <sub>IN</sub> falling	
Maximum Input Current (Limited by UVLO)	I <sub>IN, MAX</sub>	3.5 1.5	_	_	mA	V <sub>IN</sub> = 8V, T <sub>A</sub> = 25°C ( <b>Note 3</b> ) V <sub>IN</sub> = 8V, T <sub>A</sub> = 125°C ( <b>Note 3</b> )	
PWM DIMMING		ı				,	
PWMD Input Low Voltage	V <sub>PWMD(LO)</sub>	_	_	8.0	V	V <sub>IN</sub> = 8V–450V ( <b>Note 2</b> )	
PWMD Input High Voltage	V <sub>PWMD(HI)</sub>	2.2	_	_	V	V <sub>IN</sub> = 8V–450V ( <b>Note 2</b> )	
PWMD Pull-down Resistance	R <sub>PWMD</sub>	50	100	150	kΩ	V <sub>PWMD</sub> = 5V	
AVERAGE-CURRENT SENSE LOG							
Current Sense Reference Voltage	V <sub>CST</sub>	264	275	286	mV	V <sub>LD</sub> = 1.5V	
LD-to-CS Voltage Ratio	A <sub>V(LD)</sub>	0.178	0.185	0.188	_	V <sub>LD</sub> = 1.2V	
LD-to-CS Voltage Offset	A <sub>V</sub> x V <sub>LD(OFFSET)</sub>	-15		15	mV	Offset = $V_{CS}$ - $A_{V(LD)} \times V_{LD}$ , $V_{LD}$ = 1.2V	
CS Threshold Temperature Regulation	ΔV <sub>CST(TEMP)</sub>	_	_	5	mV	(Note 2)	

- Note 1: Also limited by package power dissipation limit, whichever is lower
  - 2: Denotes specifications which apply over the full operating ambient temperature range of  $-40^{\circ}\text{C} < \text{T}_{\text{A}} < +125^{\circ}\text{C}$
  - **3:** Specification is obtained by characterization and is not 100% tested.

### **ELECTRICAL CHARACTERISTICS (CONTINUED)**

<b>Electrical Specifications</b> : $T_A = 25$ °C, $V_{IN} = 12$ V, $V_{LD} = V_{DD}$ , and $V_{PWMD} = V_{DD}$ unless otherwise specified.							
Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions	
LD Input Shutdown Threshold Voltage	V <sub>LD(OFF)</sub>	_	150	_	mV	V <sub>LD</sub> falling	
LD Input Enable Threshold Voltage	$V_{LD(EN)}$	_	200	_	mV	V <sub>LD</sub> rising	
Current Sense Blanking Interval	$T_{BLANK}$	150	l	340	ns	(Note 2)	
Minimum On-time	$T_{ON(MIN)}$		l	1000	ns	$V_{CS} = V_{CST} + 30 \text{ mV}$	
Maximum Steady-state Duty Cycle	D <sub>MAX</sub>	75	1		%	Reduction in output LED current may occur beyond this duty cycle	
SHORT-CIRCUIT PROTECTION							
Hiccup Threshold Voltage	$V_{CSH}$	410	440	470	mV		
Current Limit Delay CS-to-Gate	T <sub>DELAY</sub>	_	_	185	ns	$V_{CS} = V_{CSH} + 30 \text{ mV}$	
Short-circuit Hiccup Time	T <sub>HICCUP</sub>	350	400	550	μs		
Minimum On-time (Short-circuit)	T <sub>ON(MIN),SC</sub>	_	_	430	ns	$V_{CS} = V_{DD}$	
T <sub>OFF</sub> TIMER							
Off-time	T <sub>OFF</sub>	32	40	48	μs	$R_T = 1 M\Omega$	
	· OFF	8	10	12	μυ	$R_T = 226 \text{ k}\Omega$	
GATE DRIVER							
Gate Sourcing Current	I <sub>SOURCE</sub>	0.165	_	_	Α	$V_{GATE} = 0V, V_{DD} = 7.5V$	
Gate Sinking Current	I <sub>SINK</sub>	0.165	_	_	Α	$V_{GATE} = V_{DD}, V_{DD} = 7.5V$	
Gate Output Rise Time	t <sub>r</sub>	_	30	50	ns	$C_{GATE} = 500 \text{ pF}, V_{DD} = 7.5 \text{V}$	
Gate Output Fall Time	t <sub>f</sub>	_	30	50	ns	$C_{GATE} = 500 \text{ pF}, V_{DD} = 7.5 \text{V}$	

- Note 1: Also limited by package power dissipation limit, whichever is lower
  - 2: Denotes specifications which apply over the full operating ambient temperature range of  $-40^{\circ}\text{C} < \text{T}_{\text{A}} < +125^{\circ}\text{C}$
  - 3: Specification is obtained by characterization and is not 100% tested.

### **TEMPERATURE SPECIFICATIONS**

Parameter	Sym.	Min.	Тур.	Max.	Unit	Conditions
TEMPERATURE RANGES						
Operating Ambient Temperature	T <sub>A</sub>	-40	_	+125	°C	
Junction Temperature	TJ	-40	_	+150	°C	
Storage Temperature	T <sub>S</sub>	-65	_	+150	°C	
PACKAGE THERMAL RESISTANC	Ē					
8-lead SOIC	$\theta_{JA}$	_	101	_	°C/W	
16-lead SOIC	$\theta_{JA}$	_	83	_	°C/W	

### 2.0 PIN DESCRIPTION

The details on the pins of HV9961 are listed on Table 2-1. Refer to **Package Types** for the location of pins.

TABLE 2-1: PIN FUNCTION TABLE

Pin Number		Pin Name	Paravintian		
8-lead SOIC	16-lead SOIC	Pin Name	Description		
1	1	VIN	This pin is the input of an 8V–450V linear regulator.		
2	4	CS	This pin is the current sense pin used to sense the FET current with an external sense resistor.		
3	5	GND	Ground return for all internal circuitry. This pin must be electrically connected to the ground of the power train.		
4	8	Gate	This pin is the output of gate driver for driving an external N-channel power MOSFET.		
5	9	PWMD	This is the PWM dimming input of the IC. When this pin is pulled to GND, the gate driver is turned off. When the pin is pulled high, the gate driver operates normally.		
6	12	V <sub>DD</sub>	This is the power supply pin for all internal circuits. It must be bypassed with a low ESR capacitor to GND (at least 0.1 µF).		
7	13	LD	This pin is the linear dimming input, and it sets the current sense threshold as long as the voltage at this pin is less than 1.5V. If voltage at LD falls below 150 mV, the gate output is disabled. The gate signal recovers at 200 mV at LD.		
8	14	RT	A resistor connected between this pin and GND programs the gate off-time.		
_	2, 3, 6, 7, 10, 11, 15, 16	NC	No connection.		

#### 3.0 APPLICATION INFORMATION

#### 3.1 **General Description**

Peak current control (as in HV9910B) is the simplest and the most economical way to regulate a buck converter's output current. However, it suffers accuracy regulation problems that arise peak-to-average current error, contributed by the current ripple in the output inductor and the propagation delay in the current sense comparator. The full inductor current signal is unavailable for direct switch current sensing across the sense resistor at the ground path in this low-side switch buck converter when the control switch is at the ground potential because the switch is turned off. While it is very simple to detect the peak current in the switch, controlling the average inductor current is usually implemented by level translating the sense signal from +V<sub>IN</sub>. Although this is practical for a relatively low-input voltage, V<sub>IN</sub>, this type of average-current control may become excessively complex and expensive in the offline AC or other high-voltage DC applications.

The HV9961 uses a proprietary control scheme that allows fast and accurate control of the average current in the buck inductor by sensing the switch current only. No compensation of the current control loop is required. The output LED current's response to PWMD input is similar to that of the HV9910B. The effect of inductor current ripple amplitude on this control scheme is insignificant. Therefore, the LED current is independent of the variation in inductance, switching frequency or output voltage. Constant off-time control of the buck converter is used for stability and improving the LED current regulation over a wide range of input voltages. Unlike HV9910B, the HV9961 does not support Constant Frequency mode.

#### **Off Timer** 3.2

The timing resistor connected between RT and GND determines the off-time of the gate driver. Wiring this resistor between RT and Gate as with HV9910B is no longer supported. Refer to Equation 3-1 for the computation of the gate output's off-time.

#### **EQUATION 3-1:**

$$T_{OFF}(\mu s) = \frac{R_T(k\Omega)}{25} + 0.3$$

within the range of 30 k $\Omega \le R_T \le 1 M\Omega$ 

#### 3.3 **Average-Current Control** Feedback and Output Short-circuit **Protection**

The current through the switching Metal-oxide Semiconductor Field-effect Transistor (MOSFET) source is averaged and used to give constant-current feedback. This current is detected with a sense resistor at the CS pin. The feedback operates in a fast Open-loop mode. No compensation is required. Output current is programmed as seen in Equation 3-2.

#### **EQUATION 3-2:**

$$I_{LED} = \frac{0.275 V}{R_{CS}}$$

 $I_{LED} \,=\, \frac{0.275\,V}{R_{CS}}$  When the voltage at the LD input V<sub>LD</sub> ≥ 1.5V

If the voltage at the LD input is less than 1.5V, the output current is computed as specified in Equation 3-3.

#### **EQUATION 3-3:**

$$I_{LED} = \frac{V_{LD} \times 0.185}{R_{CS}}$$

When the voltage at the LD input  $0.2V \le V_{LD} < 1.5V$ 

The above equations are only valid for continuous conduction of the output inductor. It is good design practice to choose the inductance of the inductor such that the peak-to-peak inductor current is 30% to 40% of the average DC full-load current. Hence, the recommended inductance can be calculated as shown in Equation 3-4.

#### **EQUATION 3-4:**

$$L_O = \frac{V_{O(MAX)} \times T_{OFF}}{0.4 \times I_O}$$

The duty-cycle range of the current control feedback is limited to D  $\leq$  0.75. A reduction in the LED current may occur when the desired LED string voltage  $V_{\text{O}}$  is greater than 75% of the input voltage  $V_{\mbox{\scriptsize IN}}$  of the HV9961 LED driver.

Reducing the targeted output LED string voltage V<sub>O</sub> below  $V_{O(MIN)} = V_{IN} x D_{MIN}$ , where  $D_{MIN} = 1 \mu s/(T_{OFF})$ +1 µs), may also result in the loss of regulation of the LED current. This condition, however, causes an increase in the LED current and can potentially trip the short-circuit protection comparator.

The typical output characteristic of the HV9961 LED driver is shown in Figure 3-1. The corresponding HV9910B characteristic is given for the comparison.

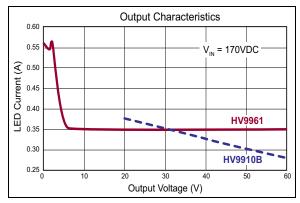


FIGURE 3-1: Typical Output Characteristic of an HV9961 LED Driver.

The short-circuit protection comparator trips when the voltage at CS exceeds 0.44V. When this occurs, the short-circuit gate off-time  $T_{HICCUP}=400~\mu s$  is generated to prevent the staircasing of the inductor current and, potentially, its saturation due to insufficient output voltage. The typical short-circuit inductor current is shown in the waveform of Figure 3-2.

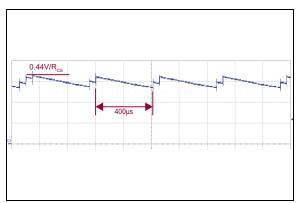


FIGURE 3-2: Short-circuit Inductor Current.

A leading-edge blanking delay is provided at CS to prevent false triggering of the current feedback and the short-circuit protection.

#### 3.4 Linear Dimming

When the voltage at LD falls below 1.5V, the internal 275 mV reference to the constant-current feedback becomes overridden by  $V_{LD} \ x$  0.185. As long as the current in the inductor remains continuous, the LED current is given by Equation 3-3. However, when  $V_{LD}$  falls below 150 mV, the gate output becomes disabled. The gate signal recovers when  $V_{LD}$  exceeds 200 mV. It is required in some applications to use the same brightness control signal input to shut off the lamp. The typical linear dimming response is shown in Figure 3-3.

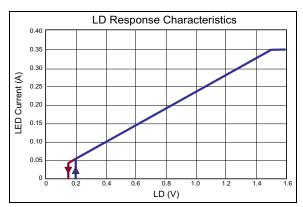


FIGURE 3-3: Typical Linear Dimming Response of an HV9961 LED Driver.

The linear dimming input could also be used for "mixed-mode" dimming to expand the dimming ratio. In such case, a pulse-width modulated signal with an amplitude below 1.5V should be applied to LD.

#### 3.5 Input Voltage Regulator

The HV9961 can be powered directly from an 8  $V_{DC}$ –450  $V_{DC}$  supply through its  $V_{IN}$  input. When this voltage is applied at the  $V_{IN}$  pin, the HV9961 maintains a constant 7.5V level at  $V_{DD}$ . This voltage can be used to power the IC and external circuitry connected to  $V_{DD}$  within the rated maximum current or within the thermal ratings of the package, whichever limit is lower. The  $V_{DD}$  pin must be bypassed by a low ESR capacitor to provide a low-impedance path for the high-frequency current of the gate output. The HV9961 can also be powered through the  $V_{DD}$  pin directly with a voltage greater than the internally regulated 7.5V, but less than 12V.

Despite the instantaneous voltage rating of 450V, continuous voltage at  $V_{\text{IN}}$  is limited by the power dissipation in the package. For example, when HV9961 draws  $I_{\text{IN}}$  = 2.5 mA from the  $V_{\text{IN}}$  input, and the 8-pin SOIC package is used, the maximum continuous voltage at  $V_{\text{IN}}$  is limited to the value shown in Equation 3-5.

#### **EQUATION 3-5:**

$$\begin{split} V_{IN(MAX)} &= \frac{T_{J(MAX)} - T_A}{R_{\theta,JA} \times I_{IN}} \\ &= 396 \, V \end{split}$$

Where:

Ambient temperature: T<sub>A</sub> = 25°C

Maximum working junction temperature:  $T_{J(MAX)} = 125$ °C Junction-to-ambient thermal resistance:

 $R_{\theta,JA} = 101^{\circ}C/W$ 

In such cases, when it is needed to operate the HV9961 from a higher voltage, a resistor or a Zener diode can be added in series with the  $V_{\text{IN}}$  input to divert some of the power loss from the HV9961. In the above example, using a 100V Zener diode will allow the circuit to work up to 490V. The input current drawn from the  $V_{\text{IN}}$  pin is represented by Equation 3-6.

#### **EQUATION 3-6:**

$$I_{IN} \approx 1 \, mA + Q_G \times f_S$$

Where:

f<sub>S</sub> = Switching frequency

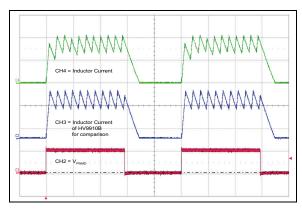
 ${
m Q}_{
m G} = {
m Gate \ charge \ of \ the \ external \ FET \ (obtained \ from \ the \ manufacturer's \ data \ sheet)}$ 

#### 3.6 Gate Output

The gate output of the HV9961 is used to drive an external MOSFET. It is recommended that the gate charge  $Q_G$  of the external MOSFET be less than 25 nC for switching frequencies  $\leq$ 100 kHz and less than 15 nC for switching frequencies >100 kHz.

#### 3.7 PWM Dimming

Due to the fast open-loop response of the average-current control loop of the HV9961, its PWM dimming performance nearly matches that of the HV9910B. The inductor current waveform comparison is shown in Figure 3-4.

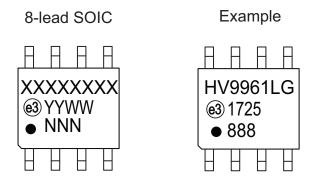


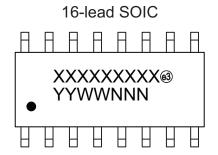
**FIGURE 3-4:** Typical PWM Dimming Response of an HV9961 LED Driver.

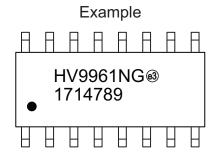
The rising and falling edges are limited by the current slew rate in the inductor. The first switching cycle is terminated upon reaching the 275 mV or  $V_{LD} \times 0.185$  level at CS. The circuit is further reaching its steady-state within 3–4 switching cycles regardless of the switching frequency.

#### 4.0 PACKAGING INFORMATION

#### 4.1 Package Marking Information





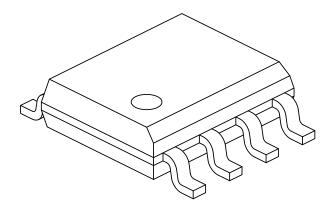


Legend: XX...X Product code or customer-specific information Year code (last digit of calendar year) ΥY Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') WW Alphanumeric traceability code NNN Pb-free JEDEC® designator for Matte Tin (Sn) (e3) This package is Pb-free. The Pb-free JEDEC designator (@3) can be found on the outer packaging for this package. •, ▲, ▼ Pin one index is identified by a dot, delta up, or delta down (triangle mark). In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information. Package may or may not include the corporate logo. Underbar (\_) and/or Overbar (¯) symbol may not be to scale.

## 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm (.150 ln.) Body [SOIC] For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging ○ 0.10 C A-B D NOTE 5 <u>E1</u> E1 0.10 C A-B ○ 0.10 C A-B NOTE 1 NX b ⊕ 0.25 M C A−B D В **TOP VIEW** // 0.10 C С A2 SEATING **PLANE** 8X 0.10 C SIDE VIEW A1 R0.13 R0.13 - 0.23 SEE VIEW C VIEW A-A VIEW C Microchip Technology Drawing No. C04-057-SN Rev F Sheet 1 of 2

### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm (.150 ln.) Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension	Limits	MIN	NOM	MAX	
Number of Pins	N		8		
Pitch	е		1.27 BSC		
Overall Height	Α	-	-	1.75	
Molded Package Thickness	A2	1.25	-	-	
Standoff §	A1	0.10	1	0.25	
Overall Width	Е	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D	4.90 BSC			
Chamfer (Optional)	h	0.25	1	0.50	
Foot Length	L	0.40	ı	1.27	
Footprint	L1	1.04 REF			
Foot Angle	φ	0° - 8°			
Lead Thickness	С	0.17 - 0.25			
Lead Width	b	0.31	-	0.51	
Mold Draft Angle Top	α	5° - 15°			
Mold Draft Angle Bottom	β	5°	-	15°	

#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- 3. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M  $\,$

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

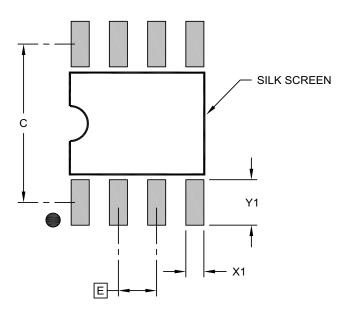
REF: Reference Dimension, usually without tolerance, for information purposes only.

5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-057-SN Rev F Sheet 2 of 2  $\,$ 

### 8-Lead Plastic Small Outline (SN) - Narrow, 3.90 mm (.150 ln.) Body [SOIC]

**bte:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

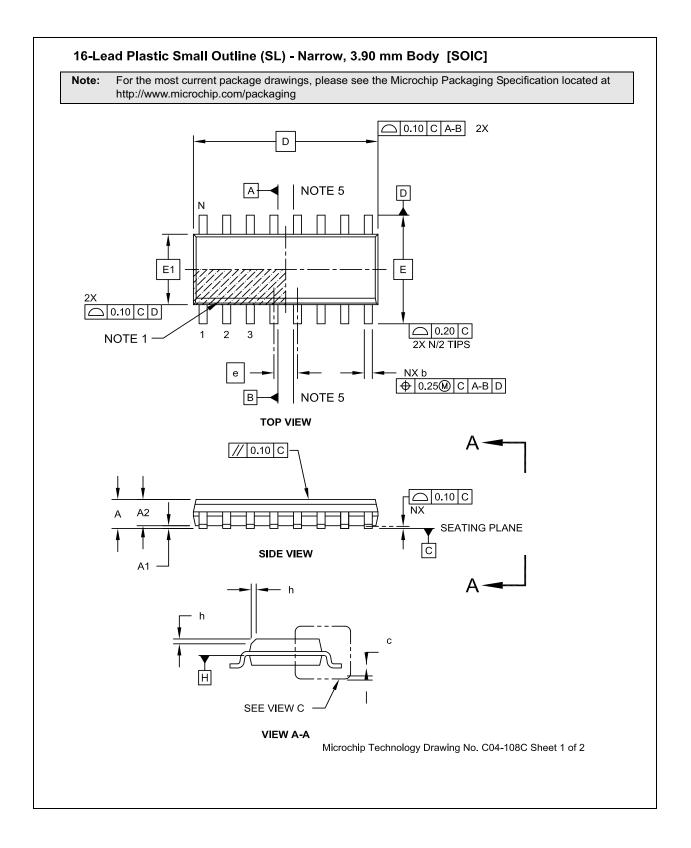
	Units			S
Dimension	Dimension Limits		NOM	MAX
Contact Pitch	E 1.27 BSC			
Contact Pad Spacing	С		5.40	
Contact Pad Width (X8)	X1			0.60
Contact Pad Length (X8)	Y1			1.55

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

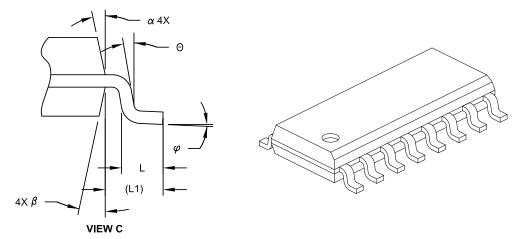
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-2057-SN Rev F



#### 16-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS				
Dimension Lir	mits	MIN	NOM	MAX	
Number of Pins	N		16	•	
Pitch	е		1.27 BSC		
Overall Height	Α	-	-	1.75	
Molded Package Thickness	A2	1.25	-	-	
Standoff §	A1	0.10	-	0.25	
Overall Width	E	6.00 BSC			
Molded Package Width	E1	3.90 BSC			
Overall Length	D	9.90 BSC			
Chamfer (Optional)	h	0.25 - 0.50			
Foot Length	L	0.40	-	1.27	
Footprint	L1		1.04 REF		
Lead Angle	Θ	0°			
Foot Angle	φ	0° - 8°			
Lead Thickness	С	0.10 - 0.25			
Lead Width	b	0.31 - 0.51			
Mold Draft Angle Top	α	5° - 15°			
Mold Draft Angle Bottom	β	5°	-	15°	

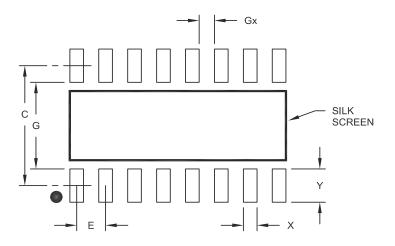
#### Notes:

- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. § Significant Characteristic
- Dimension D does not include mold flash, protrusions or gate burrs, which shall not exceed 0.15 mm per end. Dimension E1 does not include interlead flash or protrusion, which shall not exceed 0.25 mm per side.
- 4. Dimensioning and tolerancing per ASME Y14.5M
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.
- 5. Datums A & B to be determined at Datum H.

Microchip Technology Drawing No. C04-108C Sheet 2 of 2

16-Lead Plastic Small Outline (SL) - Narrow, 3.90 mm Body [SOIC]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



#### RECOMMENDED LAND PATTERN

	Units	N	11LLIMETER	S
Dimension	Limits	MIN	NOM	MAX
Contact Pitch		1.27 BSC		
Contact Pad Spacing	С		5.40	
Contact Pad Width	Х			0.60
Contact Pad Length	Υ			1.50
Distance Between Pads	Gx	0.67		
Distance Between Pads	G	3.90		

#### Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2108A



NOTES:

#### APPENDIX A: REVISION HISTORY

#### Revision B (March 2022)

- · Changed package drawings.
- Updated Section "Electrical Characteristics".
- Updated Section "Temperature Specifications".
- · Minor format changes throughout.

#### **Revision A (November 2017)**

- Converted Supertex Doc# DSFP-HV9961 to Microchip DS20005588A.
- · Changed the package marking format.
- Changed the packaging quantity of the LG package from 2500/Reel to 3300/Reel.
- Changed the packaging quantity of the NG M901 media type from 1000/Reel to 2600/Reel.
- Changed the packaging quantity of the NG M934 media type from 2500/Reel to 2600/Reel.
- Made minor text changes throughout the document.

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PART NO	o. XX		-X -X	Examples:	
Device	[		Environmental Media Type	, Cur Cur	D Driver with Average- rrent Mode Constant- rrent Control, 8-lead IC, 3300/Reel
Device:	HV9961	=	LED Driver with Average-Current Mode Constant-Current Control	Cur Cur	D Driver with Average- rrent Mode Constant- rrent Control, 16-lead IC, 45/Tube
Packages:	LG	=	8-lead SOIC		
	NG	=	16-lead SOIC	Cur	D Driver with Average- rrent Mode Constant- rrent Control. 16-lead
Environmenta	l: G	=	Lead (Pb)-free/RoHS-compliant Package	I I	IC, 2600/Reel
Media Types:	(blank)	=	3300/Reel for an LG Package		D Driver with Average-
	(blank)	=	45/Tube for an NG Package	11	rrent Mode Constant- rrent Control, 16-lead
	M901	=	2600/Reel for an NG Package		IC, 2600/Reel
	M934	=	2600/Reel for an NG Package	Note 1: Tape and Reel identi	ifier only appears in the
Note:	reel was star	ndard	M901 and M934, the base quantity for tape and ized to 2600/reel. Both options will result in the ne number of parts/reel.	used for ordering pur the device package.	description. This identifier is rposes and is not printed on Check with your Microchip ickage availability with the n.

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