

Product Change Notification: SYST-27AUGJ963

Date:

30-Sep-2024

Product Category:

Digital Potentiometers

Notification Subject:

Data Sheet - MCP4017/18/19 - 7-Bit Single I2C Digital POT with Volatile Memory in SC-70

Affected CPNs:

SYST-27AUGJ963_Affected_CPN_09302024.pdf SYST-27AUGJ963_Affected_CPN_09302024.csv

Notification Text:

SYST-27AUGJ963

Microchip has released a new Datasheet for the MCP4017/18/19 - 7-Bit Single I2C Digital POT with Volatile Memory in SC-70 of devices. If you are using one of these devices please read the document located at MCP4017/18/19 - 7-Bit Single I2C Digital POT with Volatile Memory in SC-70.

Notification Status: Final

Description of Change:

• Part number change from MCP4017T-XXXE/LT to MCP4017T-XXXE/LTY and from MCP4018T-XXXE/LT to MCP4018T-XXXE/LTY to improve manufacturability.

• Updated Section 10.0 "Packaging Information".

• Updated the Product Identification System section.

Impacts to Data Sheet: See above details.

Reason for Change: To improve productivity.

Change Implementation Status: Complete

Date Document Changes Effective: 30 Sep 2024

NOTE: Please be advised that this is a change to the document only the product has not been changed.

Markings to Distinguish Revised from Unrevised Devices: N/A

Attachments:

MCP4017/18/19 - 7-Bit Single I2C Digital POT with Volatile Memory in SC-70

Please contact your local **Microchip sales office** with questions or concerns regarding this notification.

Terms and Conditions:

If you wish to <u>receive Microchip PCNs via email</u> please register for our PCN email service at our **PCN home page** select register then fill in the required fields. You will find instructions about registering for Microchips PCN email service in the **PCN FAQ** section.

If you wish to <u>change your PCN profile</u>, <u>including opt out</u>, please go to the **PCN home page** select login and sign into your myMicrochip account. Select a profile option from

SYST-27AUGJ963 - Data Sheet - MCP4017/18/19 - 7-Bit Single I2C Digital POT with Volatile Memory in SC-70

Affected Catalog Part Numbers (CPN):

MCP4017T-103E/LT

MCP4017T-103E/LTVAO

MCP4017T-103E/LTY

MCP4017T-104E/LT

MCP4017T-104E/LTY

MCP4017T-502E/LT

MCP4017T-502E/LTY

MCP4017T-503E/LT

MCP4017T-503E/LTV01

MCP4017T-503E/LTY

MCP4018T-103E/LT

MCP4018T-103E/LTV01

MCP4018T-103E/LTVAO

MCP4018T-103E/LTY

MCP4018T-104E/LT

MCP4018T-104E/LTY

MCP4018T-502E/LT

MCP4018T-502E/LTY

MCP4018T-503E/LT

MCP4018T-503E/LTY

MCP4019T-103E/LT

MCP4019T-104E/LT

MCP4019T-502E/LT

MCP4019T-503E/LT



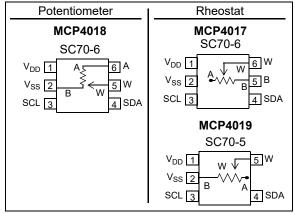
7-Bit Single I²CTM Digital POT with Volatile Memory in SC70

Features

- Potentiometer or Rheostat configuration options
- 7-bit: Resistor Network Resolution
- 127 Resistors (128 Steps)
- Zero Scale to Full Scale Wiper operation
- R_{AB} Resistances: 5 k Ω , 10 k Ω , 50 k Ω , or 100 k Ω
- Low Wiper Resistance: 100Ω (typical)
- Low Tempco:
 - Absolute (Rheostat): 50 ppm typical (0°C to 70°C)
- Ratiometric (Potentiometer): 10 ppm typical
- Simple I²C Protocol with read and write commands
- Brown-out Reset protection (1.5V typical)
- Power-on Default Wiper Setting (Mid-scale)
- Low-Power Operation:
- 2.5 µA Static Current (typical)
- Wide Operating Voltage Range:
 - 2.7V to 5.5V Device Characteristics Specified
 - 1.8V to 5.5V Device Operation

Device Features

Package Types

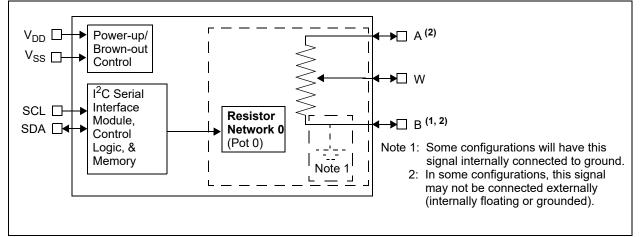


- Wide Bandwidth (-3 dB) Operation:
 2 MHz (typical) for 5.0 kΩ device
- Extended temperature range (-40°C to +125°C)
- Very small package (SC70)
- · Lead free (Pb-free) package

	e e	sdə		y	Resistance (typica	VDD		
Device	Control Interfac	# of Ste	Wiper Configuration	Memor Type	Options (k Ω)	Wiper (Ω)	Operating Range ⁽¹⁾	Package
MCP4017	l ² C	128	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	75	1.8V to 5.5V	SC70-6
MCP4018	l ² C	128	Potentiometer	RAM	5.0, 10.0, 50.0, 100.0	75	1.8V to 5.5V	SC70-6
MCP4019	l ² C	128	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	75	1.8V to 5.5V	SC70-5

Note 1: Analog characteristics only tested from 2.7V to 5.5V

Device Block Diagram



Comparison of Similar Microchip Devices (1)

		S			Resistance (typical)			ck ogy	
Device	Control Interface	# of Step:	Wiper Configuration	Memory Type	Options (k Ω)	V _{DD} Operating Range ⁽²⁾	HV Interface	WiperLock Technology	Package
MCP4017	l ² C	128	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	No	No	SC70-6
MCP4012	U/D	64	Rheostat	RAM	2.1, 5.0, 10.0, 50.0	1.8V to 5.5V	Yes	No	SOT-23-6
MCP4022	U/D	64	Rheostat	EE	2.1, 5.0, 10.0, 50.0	2.7V to 5.5V	Yes	Yes	SOT-23-6
MCP4132	SPI	129	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	Yes	No	PDIP-8,
MCP4142	SPI	129	Rheostat	EE	5.0, 10.0, 50.0, 100.0	2.7V to 5.5V	Yes	Yes	SOIC-8,
MCP4152	SPI	257	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	Yes	No	MSOP-8, DFN-8
MCP4162	SPI	257	Rheostat	EE	5.0, 10.0, 50.0, 100.0	2.7V to 5.5V	Yes	Yes	Binto
MCP4532	l ² C	129	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	Yes	No	MSOP-8,
MCP4542	l ² C	129	Rheostat	EE	5.0, 10.0, 50.0, 100.0	2.7V to 5.5V	Yes	Yes	DFN-8
MCP4552	l ² C	257	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	Yes	No	
MCP4562	l ² C	257	Rheostat	EE	5.0, 10.0, 50.0, 100.0	2.7V to 5.5V	Yes	Yes	
MCP4018	l ² C	128	Potentiometer	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	No	No	SC70-6
MCP4013	U/D	64	Potentiometer	RAM	2.1, 5.0, 10.0, 50.0	1.8V to 5.5V	Yes	No	SOT-23-6
MCP4023	U/D	64	Potentiometer	EE	2.1, 5.0, 10.0, 50.0	2.7V to 5.5V	Yes	Yes	SOT-23-6
MCP4019	l ² C	128	Rheostat	RAM	5.0, 10.0, 50.0, 100.0	1.8V to 5.5V	No	No	SC70-5
MCP4014	U/D	64	Rheostat	RAM	2.1, 5.0, 10.0, 50.0	1.8V to 5.5V	Yes	No	SOT-23-5
MCP4024	U/D	64	Rheostat	EE	2.1, 5.0, 10.0, 50.0	2.7V to 5.5V	Yes	Yes	SOT-23-5

Note 1: This table is broken into three groups by a thick line (and color coding). The unshaded devices in this table are the devices described in this data sheet, while the shaded devices offer a comparable resistor network configuration.

2: Analog characteristics only tested from 2.7V to 5.5V

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Voltage on V _{DD} with respect to V _{SS} 0.6V to +7.0V Voltage on SCL, and SDA with respect to V _{SS} $-0.6V$ to 12.5V
Voltage on all other pins (A, W, and B)
with respect to V_{SS} -0.3V to V_{DD} + 0.3V
Input clamp current, I _{IK}
$(V_1 < 0, V_1 > V_{DD}, V_1 > V_{PP} \text{ on HV pins}) \dots \pm 20 \text{ mA}$ Output clamp current, I _{OK}
$(V_{O} < 0 \text{ or } V_{O} > V_{DD}) \dots \pm 20 \text{ mA}$
Maximum output current sunk by any Output pin
Maximum output current sourced by any Output pin
Maximum current out of V _{SS} pin100 mA
Maximum current into V _{DD} pin 100 mA
Maximum current into A, W and B pins±2.5 mA
Package power dissipation ($T_A = +50^{\circ}C, T_J = +150^{\circ}C$)
SC70-5
SC70-6
Storage temperature65°C to +150°C
Ambient temperature with power applied
40°C to +125°C
ESD protection on all pins $\dots \ge 4 \text{ kV}$ (HBM) $\ge 400 \text{V}$ (MM)
Maximum Junction Temperature (T _J)+150°C

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

AC/DC CHARACTERISTICS

DC Characteristics	$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Parameters Sym		Min	Тур	Max	Units	Conditions	
Supply Voltage	V _{DD}	2.7		5.5	V	Analog Characteristics specified	
		1.8	_	5.5	V	Digital Characteristics specified	
V _{DD} Start Voltage to ensure Wiper Reset	V _{BOR}	_	—	1.65	V	RAM retention voltage (V _{RAM}) < V _{BOR}	
V _{DD} Rise Rate to ensure Power-on Reset	V _{DDRR}		(Note 7)		V/ms		
Delay after device exits the Reset state (V _{DD} > V _{BOR})	T _{BORD}	_	10	20	μS		
Supply Current (Note 8)	I _{DD}	_	45	80	μA	Serial Interface Active, Write all 0's to Volatile Wiper $V_{DD} = 5.5V$, $F_{SCL} = 400$ kHz	
		—	2.5	5	μA	Serial Interface Inactive, (Stop condition, SCL = SDA = V_{IH}), Wiper = 0, V_{DD} = 5.5V	

Note 1: Resistance is defined as the resistance between terminal A to terminal B.

2: INL and DNL are measured at V_W with $V_A = V_{DD}$ and $V_B = V_{SS}$.

3: MCP4018 device only, includes V_{WZSE} and V_{WFSE} .

4: Resistor terminals A, W and B's polarity with respect to each other is not restricted.

5: This specification by design.

6: Non-linearity is affected by wiper resistance (R_W), which changes significantly over voltage and temperature.

7: POR/BOR is not rate dependent.

8: Supply current is independent of current through the resistor network.

		Standard Operating				ess otherwise spe T _A ≤ +125°C (exter				
DC Characteristic	S	All parameters apply across the specified operating ranges unless noted. V_{DD} = +2.7V to 5.5V, 5 k Ω , 10 k Ω , 50 k Ω , 100 k Ω devices. Typical specifications represent values for V_{DD} = 5.5V, T _A = +25°C.								
Parameters	Sym	Min	Тур	Мах	Units	Co	Conditions			
Resistance	R _{AB}	4.0	5	6.0	kΩ	-502 devices (Not	e 1)			
(± 20%)		8.0	10	12.0	kΩ	-103 devices (Not	e 1)			
		40.0	50	60.0	kΩ	-503 devices (Not	e 1)			
		80.0	100	120.0	kΩ	-104 devices (Not	e 1)			
Resolution	N		128		Taps	No Missing Codes	3			
Step Resistance	R _S	—	R _{AB} / (127)	_	Ω	Note 5				
Wiper Resistance R _W		_	100	170	Ω	V_{DD} = 5.5 V, I _W = 2.0 mA, code = 00h				
			155	325	Ω	V _{DD} = 2.7 V, I _W =	2.0 mA, code = 00h			
Nominal	$\Delta R_{AB} / \Delta T$	_	50		ppm/°C					
Resistance			100	_	ppm/°C	T _A = -40°C to +85	T _A = -40°C to +85°C			
Тетрсо			150	_	ppm/°C	T _A = -40°C to +12	5°C			
Ratiometeric Tempco	$\Delta V_{WB} / \Delta T$	_	15	—	ppm/°C	Code = Midscale ((3Fh)			
Resistor Terminal Input Voltage Range (Terminals A, B and W)	V _A ,V _W ,V _B	Vss	—	V _{DD}	V	Note 4, Note 5				
Maximum current	Ι _Τ	_	_	2.5	mA	Terminal A	I _{AW} , W = Full Scale (FS)			
through Terminal			—	2.5	mA	Terminal B	I _{BW} , W = Zero Scale (ZS)			
(A, W or B) Note 5			—	2.5	mA	Terminal W	I_{AW} or I_{BW} , W = FS or ZS			
			—	1.38	mA		$I_{AB}, V_B = 0V, V_A = 5.5V,$ $R_{AB(MIN)} = 4000$			
				0.688	mA	Terminal A and	I_{AB} , $V_B = 0V$, $V_A = 5.5V$, $R_{AB(MIN)} = 8000$			
				0.138	mA	Terminal B	$I_{AB}, V_B = 0V, V_A = 5.5V,$ $R_{AB(MIN)} = 40000$			
		_	_	0.069	mA		$I_{AB}, V_B = 0V, V_A = 5.5V,$ $R_{AB(MIN)} = 80000$			

Note 1: Resistance is defined as the resistance between terminal A to terminal B.

- **2:** INL and DNL are measured at V_W with $V_A = V_{DD}$ and $V_B = V_{SS}$.
- 3: MCP4018 device only, includes V_{WZSE} and V_{WFSE} .
- 4: Resistor terminals A, W and B's polarity with respect to each other is not restricted.
- **5:** This specification by design.
- **6:** Non-linearity is affected by wiper resistance (R_W), which changes significantly over voltage and temperature.
- 7: POR/BOR is not rate dependent.
- 8: Supply current is independent of current through the resistor network.

DC Characteristics	$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Parameters	Min	Тур	Мах	Units		Conditions	
Full Scale Error (MCP4018 only) (code = 7Fh)	V _{WFSE}	-3.0	-0.1		LSb	5 kΩ	$2.7V \le V_{DD} \le 5.5V$
		-2.0	-0.1		LSb	10 k Ω	$2.7V \leq V_{DD} \leq 5.5V$
		-0.5	-0.1	_	LSb	50 k Ω	$2.7V \leq V_{DD} \leq 5.5V$
		-0.5	-0.1	_	LSb	$100 \ \text{k}\Omega$	$2.7V \leq V_{DD} \leq 5.5V$
Zero Scale Error	V _{WZSE}		+0.1	+3.0	LSb	$5 \text{ k}\Omega$	$2.7V \leq V_{DD} \leq 5.5V$
(MCP4018 only) (code = 00h)			+0.1	+2.0	LSb	10 k Ω	$2.7V \leq V_{DD} \leq 5.5V$
(code - 001)			+0.1	+0.5	LSb	50 k Ω	$2.7V \leq V_{DD} \leq 5.5V$
		—	+0.1	+0.5	LSb	100 k Ω	$2.7V \leq V_{DD} \leq 5.5V$
Potentiometer Integral Non-linearity	INL	-0.5	±0.25	+0.5	LSb		/ _{DD} ≤ 5.5V 18 device only (Note 2)
Potentiometer Differential Non- linearity	DNL	-0.25	±0.125	+0.25	LSb		$V_{DD} \le 5.5V$ 18 device only (Note 2)
Bandwidth -3 dB	BW	_	2	—	MHz	$5 k\Omega$	Code = 3Fh
(See Figure 2-83,		_	1	—	MHz	10 k Ω	Code = 3Fh
load = 30 pF)			260		kHz	50 k Ω	Code = 3Fh
			100	—	kHz	$100 \ \text{k}\Omega$	Code = 3Fh

Note 1: Resistance is defined as the resistance between terminal A to terminal B.

2: INL and DNL are measured at V_W with $V_A = V_{DD}$ and $V_B = V_{SS}$.

3: MCP4018 device only, includes V_{WZSE} and V_{WFSE} .

4: Resistor terminals A, W and B's polarity with respect to each other is not restricted.

5: This specification by design.

- **6:** Non-linearity is affected by wiper resistance (R_W), which changes significantly over voltage and temperature.
- 7: POR/BOR is not rate dependent.

8: Supply current is independent of current through the resistor network.

		Standard Operating					rwise specified) 5°C (extended)				
DC Characteristic	S	All parameters apply across the specified operating ranges unless noted. $V_{DD} = +2.7V$ to 5.5V, 5 k Ω , 10 k Ω , 50 k Ω , 100 k Ω devices. Typical specifications represent values for $V_{DD} = 5.5V$, $T_A = +25^{\circ}C$.									
Parameters	Sym	Min	Тур	Max	Units		Conditions				
Rheostat Integral	R-INL	-2.0	±0.5	+2.0	LSb	5 kΩ	5.5V, I _W = 900 μA				
Non-linearity		-5.0	+3.5	+5.0	LSb		2.7V, I _W = 430 μA (Note 6)				
MCP4018 (Note 3) MCP4017 and MCP4019 devices		See Section 2.0			LSb		1.8V (Note 6)				
		-2.0	±0.5	+2.0	LSb	10 kΩ	5.5V, Ι _W = 450 μA				
		-4.0	+2.5	+4.0	LSb		2.7V, I _W = 215 μA (Note 6)				
only (Note 3)		See	Section	2.0	LSb		1.8V (Note 6)				
		-1.125	±0.5	+1.125	LSb	50 k Ω	5.5V, Ι _W = 90 μA				
		-1.5	+1	+1.5	LSb		2.7V, I _W = 43 μA (Note 6)				
		See Section 2.0			LSb		1.8V (Note 6)				
		-0.8	±0.5	+0.8	LSb	100 kΩ	5.5V, Ι _W = 45 μA				
		-1.125	+0.25	+1.125	LSb		2.7V, I _W = 21.5 μA (Note 6)				
		See Section 2.0		LSb		1.8V (Note 6)					
Rheostat	R-DNL	-0.5	±0.25	+0.5	LSb	5 kΩ	5.5V, I _W = 900 mA				
Differential Non-		-0.75	+0.5	+0.75	LSb		2.7V, I _W = 430 µA (Note 6)				
linearity MCP4018		See	Section	2.0	LSb		1.8V (Note 6)				
(Note 3)		-0.5	±0.25	+0.5	LSb	10 k Ω	5.5V, Ι _W = 450 μA				
MCP4017 and		-0.75	+0.5	+0.75	LSb		2.7V, I _W = 215 μA (Note 6)				
MCP4019 devices only (Note 3)		See	Section	2.0	LSb		1.8V (Note 6)				
		-0.375	±0.25	+0.375	LSb	50 k Ω	5.5V, Ι _W = 90 μA				
		-0.375	±0.25	+0.375	LSb		2.7V, I _W = 43 μA (Note 6)				
		See	Section	2.0	LSb		1.8V (Note 6)				
		-0.375	±0.25	+0.375	LSb	100 kΩ	5.5V, Ι _W = 45 μA				
						-0.375	±0.25	+0.375	LSb		2.7V, I _W = 21.5 μA (Note 6)
		See Section 2.0			LSb		1.8V (Note 6)				
Capacitance (P _A)	C _{AW}	—	75	—	pF	f =1 MH	z, Code = Full Scale				
Capacitance (P _w)	C _W	—	120		pF		z, Code = Full Scale				
Capacitance (P _B)	C _{BW}	—	75	—	pF	f =1 MHz, Code = Full Scale					

Note 1: Resistance is defined as the resistance between terminal A to terminal B.

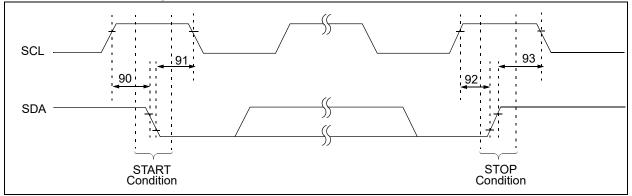
- **2:** INL and DNL are measured at V_W with $V_A = V_{DD}$ and $V_B = V_{SS}$.
- 3: MCP4018 device only, includes V_{WZSE} and V_{WFSE} .
- 4: Resistor terminals A, W and B's polarity with respect to each other is not restricted.
- 5: This specification by design.
- **6:** Non-linearity is affected by wiper resistance (R_W), which changes significantly over voltage and temperature.
- 7: POR/BOR is not rate dependent.
- 8: Supply current is independent of current through the resistor network.

		a	•	•				1 6 N			
		$\begin{array}{llllllllllllllllllllllllllllllllllll$									
DC Characteristic	S	All parameters apply across the specified operating ranges unless noted. V _{DD} = +2.7V to 5.5V, 5 k Ω , 10 k Ω , 50 k Ω , 100 k Ω devices. Typical specifications represent values for V _{DD} = 5.5V, T _A = +25°C.									
Parameters	Sym	Min	Тур	Max	Units	Conditions					
Digital Inputs/Out	puts (SDA, S	CK)									
Schmitt Trigger High Input Threshold	V _{IH}	0.7 V _{DD}	_	_	V	1.8V ≤ \	/ _{DD} ≤ 5.5V	,			
Schmitt Trigger Low Input Threshold	V _{IL}	-0.5		0.3V _{DD}	V						
Hysteresis of	V _{HYS}	—	$0.1V_{DD}$	—	V	All input	s except S	SDA and SCL			
Schmitt Trigger		N.A.	_	—	V		100 kHz	V _{DD} < 2.0V			
Inputs (Note 5)		N.A.	_	—	V	SDA and		$V_{DD} \geq 2.0 V$			
		0.1 V _{DD}	_	_	V	SCL	400 kHz	V _{DD} < 2.0V			
		0.05 V _{DD}	_	—	V		400 KHZ	$V_{DD} \ge 2.0V$			
Output Low	V _{OL}	V _{SS}	_	$0.2V_{DD}$	V	V _{DD} < 2.0V, I _{OL} = 1 mA					
Voltage (SDA)		V _{SS}	_	0.4	V	$V_{DD} \ge 2$.0V, I _{OL} = 3	3 mA			
Input Leakage Current	Ι _{ΙL}	-1	_	1	μA	$V_{IN} = V_I$	_{DD} and V _{IN}	= V _{SS}			
Pin Capacitance	C _{IN} , C _{OUT}	_	10		pF	f _C = 400) kHz				
RAM (Wiper) Value	9										
Value Range	N	0h	_	7Fh	hex						
Wiper POR/BOR Value	N _{POR/BOR}		3Fh		hex						
Power Requireme	nts										
Power Supply Sensitivity (MCP4018 only)	PSS	—	0.0005	0.0035	%/%		$V_{DD} = 2.7$ $V_A = 2.7$	7V to 5.5V, /, Code = 3Fh			

Note 1: Resistance is defined as the resistance between terminal A to terminal B.

- **2:** INL and DNL are measured at V_W with $V_A = V_{DD}$ and $V_B = V_{SS}$.
- 3: MCP4018 device only, includes V_{WZSE} and V_{WFSE} .
- 4: Resistor terminals A, W and B's polarity with respect to each other is not restricted.
- 5: This specification by design.
- **6:** Non-linearity is affected by wiper resistance (R_W), which changes significantly over voltage and temperature.
- 7: POR/BOR is not rate dependent.
- 8: Supply current is independent of current through the resistor network.

1.1 I²C Mode Timing Waveforms and Requirements



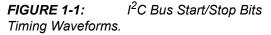


TABLE 1-1: I²C BUS START/STOP BITS REQUIREMENTS

I ² C AC Characteristics			Standard Operating Conditions (unless otherwise specified)Operating Temperature $-40^{\circ}C \le TA \le +125^{\circ}C$ (Extended)Operating Voltage VDD range is described in Section 2.0 "TypicalPerformance Curves"							
Param. No.	Symbol	Characte	ristic	Min	Мах	Units	Conditions			
	F _{SCL}		Standard Mode	0	100	kHz	C _b = 400 pF, 1.8V - 5.5V			
			Fast Mode	0	400	kHz	C _b = 400 pF, 2.7V - 5.5V			
D102	Cb	Bus capacitive	100 kHz mode		400	pF				
		loading	400 kHz mode		400	pF				
90	TSU:STA	START condition	100 kHz mode	4700		ns	Only relevant for repeated			
		Setup time	400 kHz mode	600	_	ns	START condition			
91	THD:STA	START condition	100 kHz mode	4000		ns	After this period the first			
		Hold time	400 kHz mode	600		ns	clock pulse is generated			
92	Tsu:sto	STOP condition	100 kHz mode	4000	_	ns				
		Setup time	400 kHz mode	600	_	ns				
93	THD:STO	STOP condition	100 kHz mode	4000	_	ns				
		Hold time	400 kHz mode	600	_	ns				

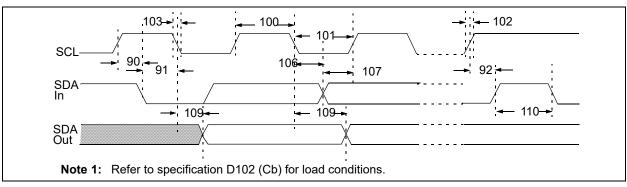


FIGURE 1-2: I²C Bus Data Timing.

TABLE 1-2 :	I ² C BUS DATA REQUIREMENTS (SLAVE MODE)
--------------------	---

I ² C AC Ch	aracterist	ics	$\begin{array}{llllllllllllllllllllllllllllllllllll$							
Parame- ter No.	Sym	Characteristic		Min	Max	Units	Conditions			
100	T _{HIGH}	Clock high time	100 kHz mode	4000	_	ns	1.8V-5.5V			
			400 kHz mode	600	—	ns	2.7V-5.5V			
101	T _{LOW}	Clock low time	100 kHz mode	4700	—	ns	1.8V-5.5V			
			400 kHz mode	1300	—	ns	2.7V-5.5V			
102A ⁽⁵⁾	T _{RSCL}	SCL rise time	100 kHz mode	—	1000	ns	Cb is specified to be from			
			400 kHz mode	20 + 0.1Cb	300	ns	10 to 400 pF			
102B ⁽⁵⁾	T _{RSDA}	SDA rise time	100 kHz mode	—	1000	ns	Cb is specified to be from			
			400 kHz mode	20 + 0.1Cb	300	ns	10 to 400 pF			
103A ⁽⁵⁾	03A (5) T _{FSCL} SCL fall time	SCL fall time	100 kHz mode	—	300	ns	Cb is specified to be from			
			400 kHz mode	20 + 0.1Cb	40	ns	10 to 400 pF			
103B ⁽⁵⁾	T _{FSDA}	SDA fall time	100 kHz mode	—	300	ns	Cb is specified to be from			
			400 kHz mode	20 + 0.1Cb (4)	300	ns	10 to 400 pF			
106	THD:DAT	Data input hold	100 kHz mode	0	—	ns	1.8V-5.5V, Note 6			
		time	400 kHz mode	0	_	ns	2.7V-5.5V, Note 6			
107	TSU:DAT	Data input	100 kHz mode	250	-	ns	(2)			
		setup time	400 kHz mode	100	_	ns				
109	T _{AA}	Output valid	100 kHz mode	—	3450	ns	(1)			
		from clock	400 kHz mode	—	900	ns				
110	110 T _{BUF}	Bus free time	100 kHz mode	4700	-	ns	Time the bus must be free			
			400 kHz mode	1300	—	ns	before a new transmission can start			
	T _{SP}	Input filter spike	100 kHz mode	—	50	ns	Philips Spec states N.A.			
		suppression (SDA and SCL)	400 kHz mode	—	50	ns				

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of START or STOP conditions.

2: A fast-mode (400 kHz) I²C-bus device can be used in a standard-mode (100 kHz) I²C-bus system, but the requirement Ts∪; DAT ≥ 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line

TR max.+Tsu;DAT = 1000 + 250 = 1250 ns (according to the standard-mode I^2C bus specification) before the SCL line is released.

- **3:** The MCP4018/MCP4019 device must provide a data hold time to bridge the undefined part between VIH and VIL of the falling edge of the SCL signal. This specification is not a part of the I²C specification, but must be tested in order to guarantee that the output data will meet the setup and hold specifications for the receiving device.
- **4:** Use Cb in pF for the calculations.
- 5: Not Tested.
- **6:** A Master Transmitter must provide a delay to ensure that difference between SDA and SCL fall times do not unintentionally create a Start or Stop condition.

TEMPERATURE CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, V _{DD} = +1.8V to +5.5V, V _{SS} = GND.										
Parameters	Sym	Min	Тур	Max	Units	Conditions				
Temperature Ranges										
Specified Temperature Range	Τ _Α	-40	—	+125	°C					
Operating Temperature Range	Τ _Α	-40	—	+125	°C					
Storage Temperature Range	Τ _Α	-65	—	+150	°C					
Thermal Package Resistances			•							
Thermal Resistance, 5L-SC70 (Note 1)	θ_{JA}	—	331	—	°C/W					
Thermal Resistance, 6L-SC70	θ_{JA}	—	TBD	—	°C/W					

Note 1: Package Power Dissipation (PDIS) is calculated as follows:

 $P_{DIS} = (T_J - T_A) / \theta_{JA},$ where: $T_J =$ Junction Temperature, $T_A =$ Ambient Temperature.

NOTES:

2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

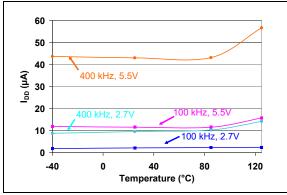


FIGURE 2-1: Interface Active Current (I_{DD}) vs. SCL Frequency (f_{SCL}) and Temperature $(V_{DD} = 1.8V, 2.7V \text{ and } 5.5V)$.

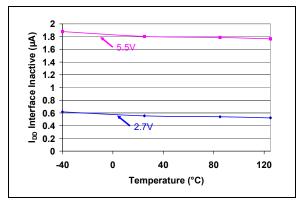


FIGURE 2-2: Interface Inactive Current (I_{SHDN}) vs. Temperature and V_{DD} . $(V_{DD} = 1.8V, 2.7V$ and 5.5V).

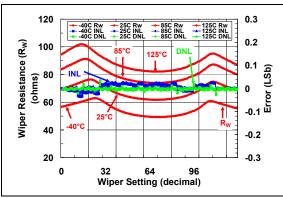


FIGURE 2-3: 5.0 k Ω : Pot Mode – $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 5.5V). (A = V_{DD}, B = V_{SS}).

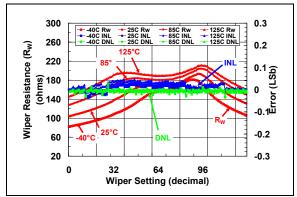


FIGURE 2-4: 5.0 k Ω : Pot Mode – $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V). (A = V_{DD} , B = V_{SS}).

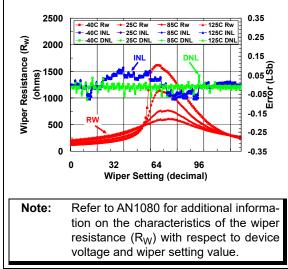


FIGURE 2-5: 5.0 k Ω : Pot Mode – $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V). (A = V_{DD} , B = V_{SS}).

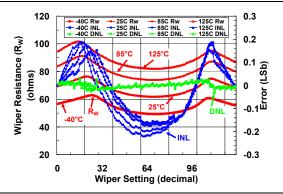


FIGURE 2-6: $5.0 \ k\Omega$: Rheo Mode – R_W (Ω) , INL (LSb), DNL (LSb) vs. Wiper Setting andTemperature (V_{DD} = 5.5V).(I_W = 1.4mA, $B = V_{SS}$).

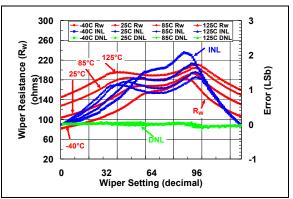


FIGURE 2-7: 5.0 k Ω : Rheo Mode – R_W (Ω), INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V).(I_W = 450uA, B = V_{SS}).

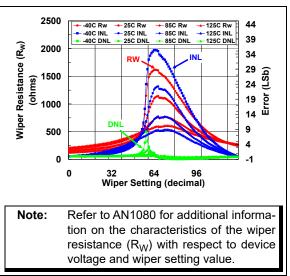


FIGURE 2-8: 5.0 k Ω : Rheo Mode – R_W (Ω), INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V). (I_W = TBD, B = V_{SS}).

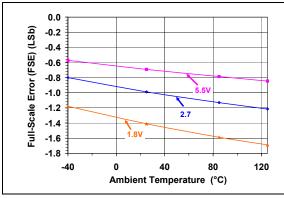


FIGURE 2-9: 5.0 $k\Omega$: Full Scale Error (FSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

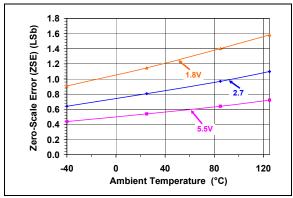


FIGURE 2-10: 5.0 $k\Omega$: Zero Scale Error (ZSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

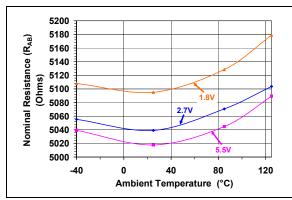


FIGURE 2-11: 5.0 k Ω : Nominal Resistance (Ω) vs. Temperature and V_{DD}.

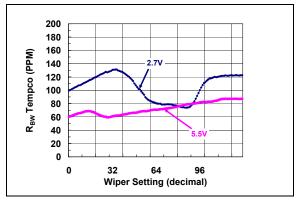


FIGURE 2-12: 5.0 k Ω : R_{BW} Tempco $\Delta R_{WB} / \Delta T$ vs. Code.

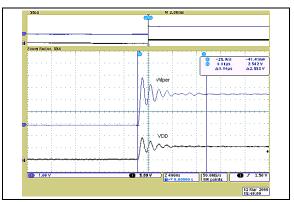


FIGURE 2-13: 5.0 $k\Omega$: Power-Up Wiper Response Time.

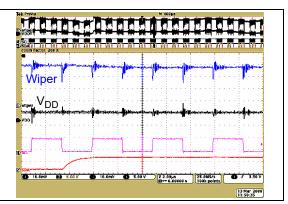


FIGURE 2-14: 5.0 k Ω : Digital Feedthrough (SCL signal coupling to Wiper pin).

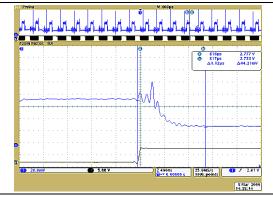


FIGURE 2-15: 5.0 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=5.5V).



FIGURE 2-16: 5.0 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=2.7V).

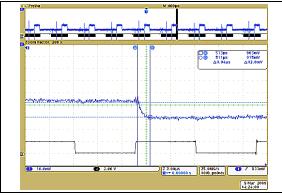


FIGURE 2-17: 5.0 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=1.8V).

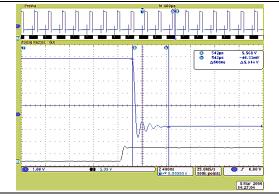


FIGURE 2-18: 5.0 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=5.5V).

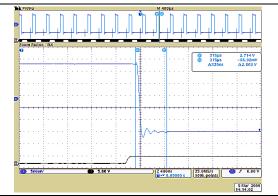


FIGURE 2-19: 5.0 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=2.7V).

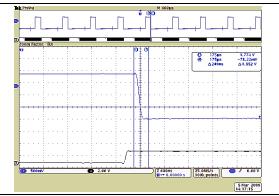


FIGURE 2-20: 5.0 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=1.8V).

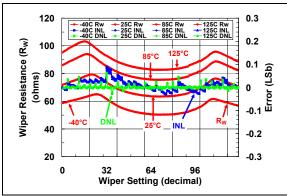


FIGURE 2-21: 10 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 5.5V$). ($A = V_{DD}$, $B = V_{SS}$).

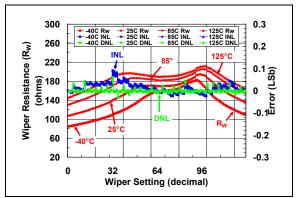


FIGURE 2-22: 10 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 2.7V$). ($A = V_{DD}$, $B = V_{SS}$).

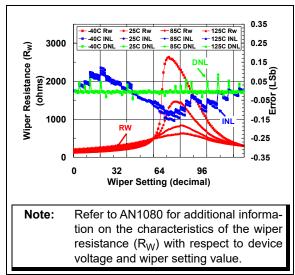


FIGURE 2-23: 10 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 1.8V$). ($A = V_{DD}$, $B = V_{SS}$).

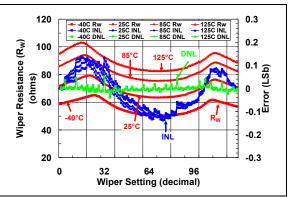


FIGURE 2-24: 10 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 5.5V$).($I_W = 450uA$, $B = V_{SS}$).

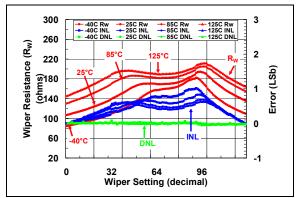


FIGURE 2-25: 10 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 2.7V$).($I_W = 210uA$, $B = V_{SS}$).

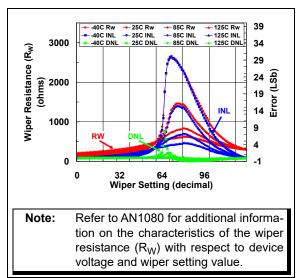


FIGURE 2-26: 10 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature ($V_{DD} = 1.8V$). ($I_W = TBD$, $B = V_{SS}$).

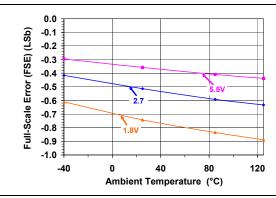


FIGURE 2-27: 10 k Ω : Full Scale Error (FSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

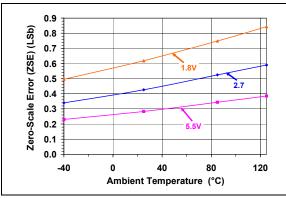


FIGURE 2-28: 10 k Ω : Zero Scale Error (ZSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

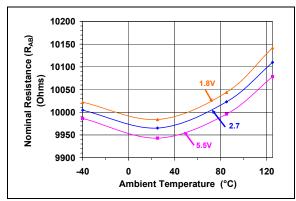


FIGURE 2-29: 10 $k\Omega$: Nominal Resistance (Ω) vs. Temperature and V_{DD}.

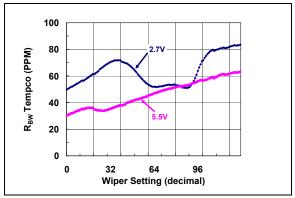


FIGURE 2-30: 10 k Ω : R_{BW} Tempco $\Delta R_{WB} / \Delta T$ vs. Code.

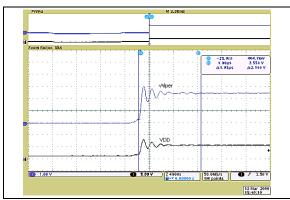


FIGURE 2-31: 10 $k\Omega$: Power-Up Wiper Response Time.

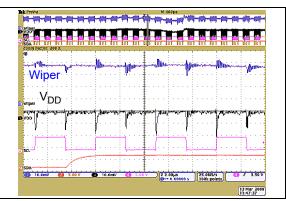


FIGURE 2-32: 10 $k\Omega$: Digital Feedthrough (SCL signal coupling to Wiper pin).

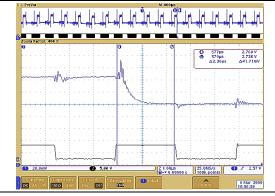


FIGURE 2-33: 10 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=5.5V).

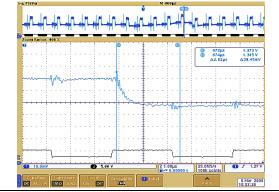


FIGURE 2-34: 10 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=2.7V).

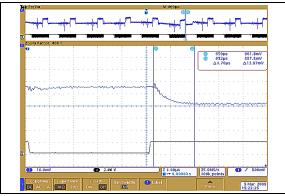


FIGURE 2-35: 10 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=1.8V).

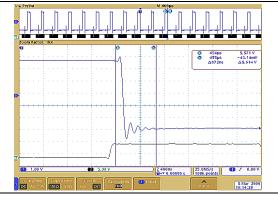


FIGURE 2-36: 10 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=5.5V).

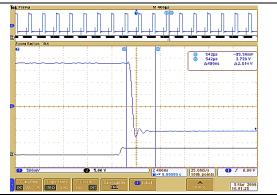


FIGURE 2-37: 10 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=2.7V).

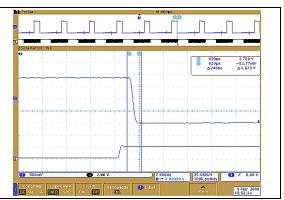


FIGURE 2-38: 10 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=1.8V).

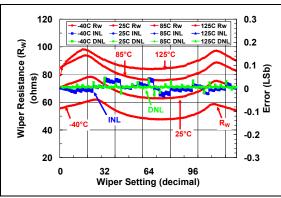


FIGURE 2-39: 50 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 5.5V).

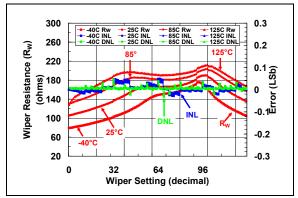


FIGURE 2-40: 50 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V).

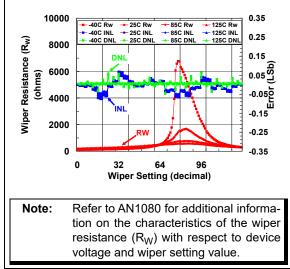


FIGURE 2-41: 50 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V).

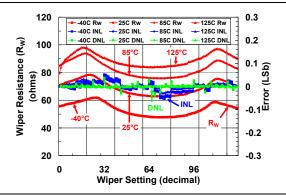


FIGURE 2-42: 50 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 5.5V).(I_W = 90uA, B = V_{SS}).

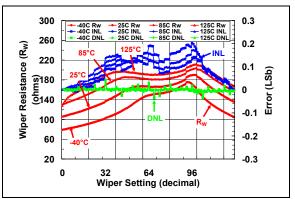


FIGURE 2-43: 50 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V).(I_W = 45uA, B = V_{SS}).

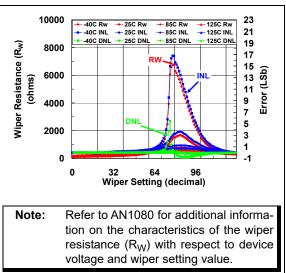


FIGURE 2-44: 50 k Ω Rheo Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V). (I_W = TBD, B = V_{SS}).

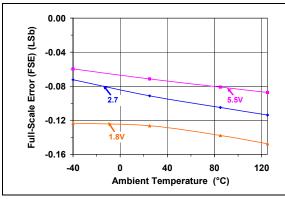


FIGURE 2-45: 50 k Ω : Full Scale Error (FSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

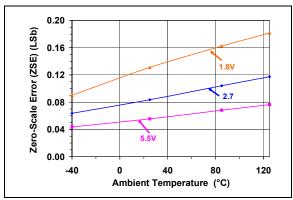


FIGURE 2-46: 50 k Ω : Zero Scale Error (ZSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

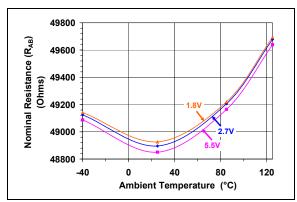


FIGURE 2-47: 50 $k\Omega$: Nominal Resistance (Ω) vs. Temperature and V_{DD} .

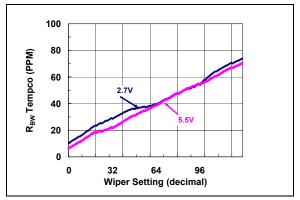


FIGURE 2-48: 50 k Ω : R_{BW} Tempco $\Delta R_{WB} / \Delta T$ vs. Code.

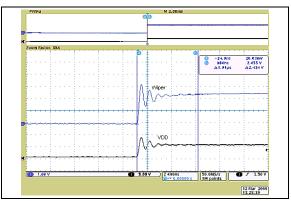


FIGURE 2-49: 50 k Ω : Power-Up Wiper Response Time.



FIGURE 2-50: 50 $k\Omega$: Digital Feedthrough (SCL signal coupling to Wiper pin).

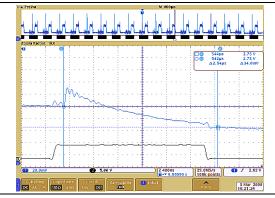


FIGURE 2-51: 50 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=5.5V).

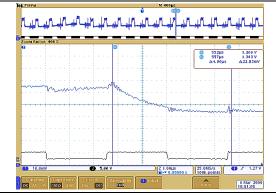


FIGURE 2-52: 50 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=2.7V).

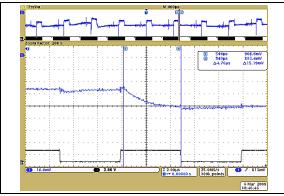


FIGURE 2-53: 50 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD}=1.8V).

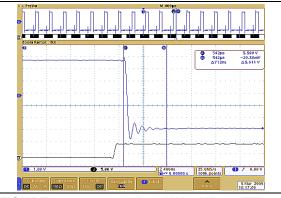


FIGURE 2-54: 50 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=5.5V).

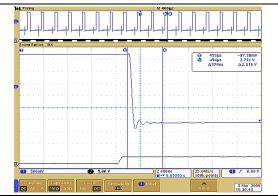


FIGURE 2-55: 50 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=2.7V).

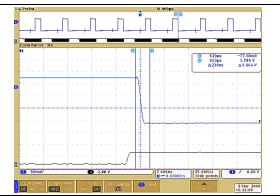


FIGURE 2-56: 50 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD}=1.8V).

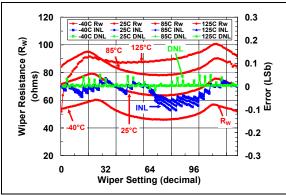


FIGURE 2-57: 100 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 5.5V).

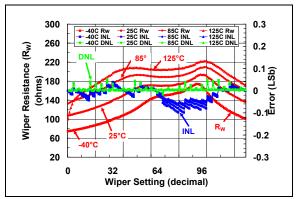


FIGURE 2-58: 100 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V).

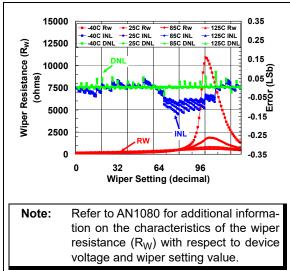


FIGURE 2-59: 100 k Ω Pot Mode : $R_W(\Omega)$, INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V).

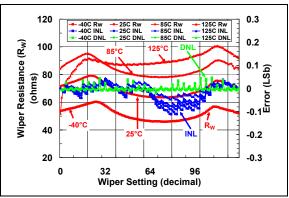


FIGURE 2-60: $100 \ k\Omega$ Rheo Mode : R_W (Ω) , INL (LSb), DNL (LSb) vs. Wiper Setting andTemperature (V_{DD} = 5.5V). (I_W = 45uA, B = V_{SS}).

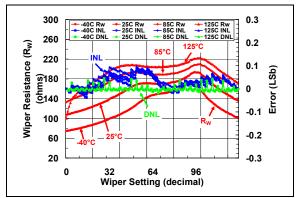


FIGURE 2-61: 100 k Ω Rheo Mode : R_W (Ω), INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 2.7V). (I_W = 21uA, B = V_{SS}).

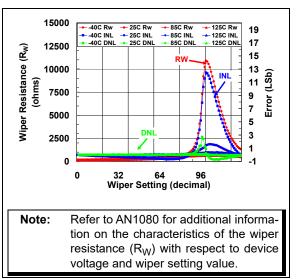


FIGURE 2-62: 100 k Ω Rheo Mode : R_W (Ω), INL (LSb), DNL (LSb) vs. Wiper Setting and Temperature (V_{DD} = 1.8V). (I_W = TBD, B = V_{SS}).

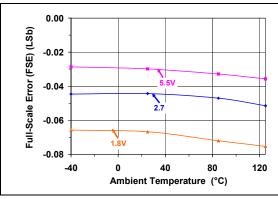


FIGURE 2-63: 100 $k\Omega$: Full Scale Error (FSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

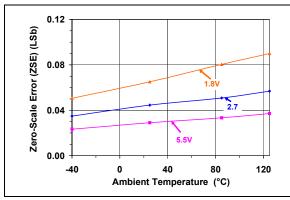


FIGURE 2-64: 100 k Ω : Zero Scale Error (ZSE) vs. Temperature (V_{DD} = 5.5V, 2.7V, 1.8V).

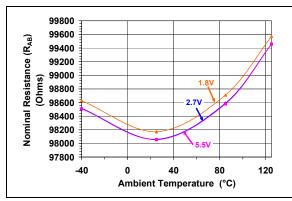


FIGURE 2-65: 100 k Ω : Nominal Resistance (Ω) vs. Temperature and V_{DD}.

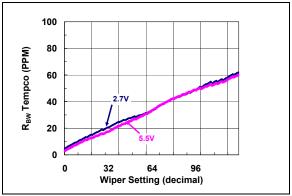


FIGURE 2-66: 100 k Ω : R_{BW} Tempco $\Delta R_{WB} / \Delta T$ vs. Code.

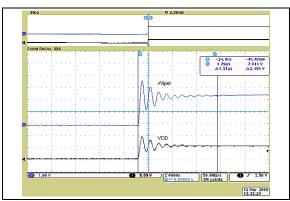


FIGURE 2-67: 100 $k\Omega$: Power-Up Wiper Response Time.

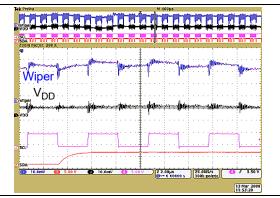


FIGURE 2-68: 100 $k\Omega$: Digital Feedthrough (SCL signal coupling to Wiper pin).

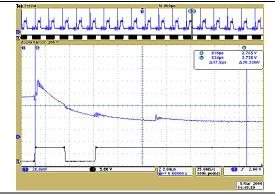


FIGURE 2-69: 100 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD} = 5.5V).

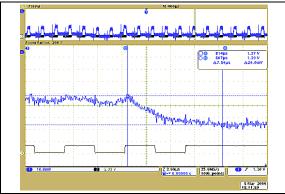


FIGURE 2-70: 100 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD} = 2.7V).

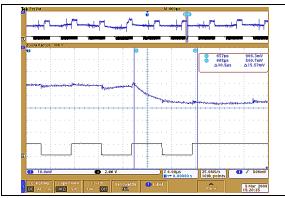


FIGURE 2-71: 100 k Ω : Write Wiper (40h \rightarrow 3Fh) Settling Time (V_{DD} = 1.8V).

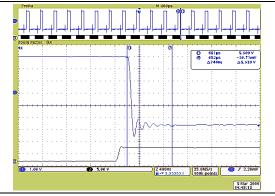


FIGURE 2-72: 100 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD} = 5.5V).

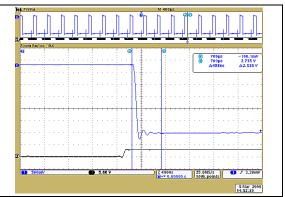


FIGURE 2-73: 100 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD} = 2.7V).

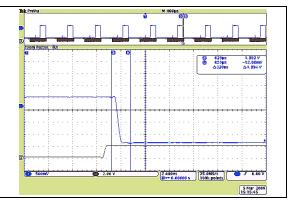


FIGURE 2-74: 100 k Ω : Write Wiper (FFh \rightarrow 00h) Settling Time (V_{DD} = 1.8V).

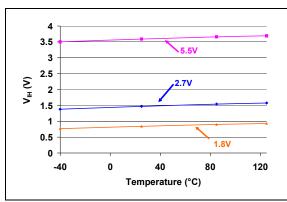


FIGURE 2-75: V_{IH} (SCL, SDA) vs. V_{DD} and Temperature.

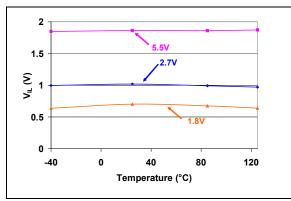


FIGURE 2-76: V_{IL} (SCL, SDA) vs. V_{DD} and Temperature.

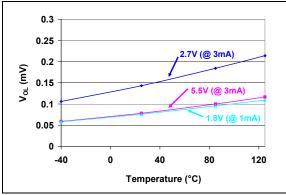


FIGURE 2-77: V_{OL} (SDA) vs. V_{DD} and Temperature.

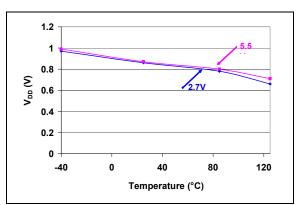


FIGURE 2-78: POR/BOR Trip point vs. V_{DD} and Temperature.

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = 5V, V_{SS} = 0V.

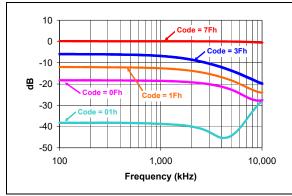


FIGURE 2-79: $5 k\Omega$ – Gain vs. Frequency (-3dB).

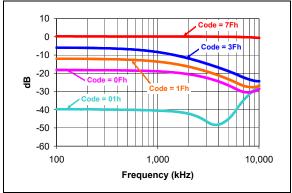


FIGURE 2-80: 10 $k\Omega$ – Gain vs. Frequency (-3dB).

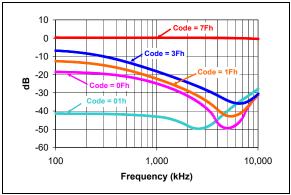


FIGURE 2-81: 50 $k\Omega$ – Gain vs. Frequency (-3dB).

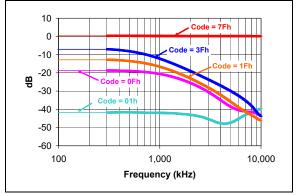


FIGURE 2-82: 100 $k\Omega$ – Gain vs. Frequency (-3dB).

2.1 Test Circuits

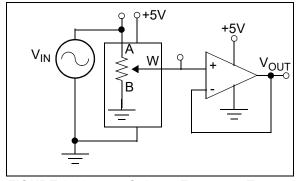


FIGURE 2-83: Gain vs. Frequency Test (-3dB).

NOTES:

3.0 **PIN DESCRIPTIONS**

The descriptions of the pins are listed in Table 3-1. Additional descriptions of the device pins follow.

Pin	Pin Number			Pin	Buffer		
Name	MCP4017 (SC70-6)	MCP4018 (SC70-6)	MCP4019 (SC70-5)	Туре	Туре	Function	
V _{DD}	1	1	1	Р	—	Positive Power Supply Input	
V _{SS}	2	2	2	Р	—	Ground	
SCL	3	3	3	I/O	ST (OD)	I ² C Serial Clock pin	
SDA	4	4	4	I/O	ST (OD)	I ² C Serial Data pin	
В	5	_	_	I/O	Α	Potentiometer Terminal B	
W	6	5	5	I/O	Α	Potentiometer Wiper Terminal	
А	—	6	—	I/O	А	Potentiometer Terminal A	

TABLE 3-1:PINOUT DESCRIPTION FOR THE MCP4017/18/19

Legend: A = Analog input ST (OD) = Schmitt Trigger with Open Drain

I = Input O = Output I/O = Input/Output P = Power

3.1 Positive Power Supply Input (V_{DD})

The V_{DD} pin is the device's positive power supply input. The input power supply is relative to V_{SS} and can range from 1.8V to 5.5V. A de-coupling capacitor on V_{DD} (to V_{SS}) is recommended to achieve maximum performance.

While the device's voltage is in the range of $1.8V \le V_{DD}$ < 2.7V, the Resistor Network's electrical performance of the device may not meet the data sheet specifications.

3.2 Ground (V_{SS})

The V_{SS} pin is the device ground reference.

3.3 I²C Serial Clock (SCL)

The SCL pin is the serial clock pin of the I²C interface. The MCP401X acts only as a slave and the SCL pin accepts only external serial clocks. The SCL pin is an open-drain output. Refer to **Section 5.0 "Serial Interface - I2C Module"** for more details of I²C Serial Interface communication.

3.4 I²C Serial Data (SDA)

The SDA pin is the serial data pin of the I^2C interface. The SDA pin has a Schmitt Trigger input and an open-drain output. Refer to **Section 5.0** "**Serial Interface - I2C Module**" for more details of I^2C Serial Interface communication.

3.5 Potentiometer Terminal B

The terminal B pin (available on some devices) is connected to the internal potentiometer's terminal B.

The potentiometer's terminal B is the fixed connection to the Zero Scale (0x00 tap) wiper value of the digital potentiometer.

The terminal B pin is available on the MCP4017 device. The terminal B pin does not have a polarity relative to the terminal W pin. The terminal B pin can support both positive and negative currents. The voltage on terminal B must be between V_{SS} and V_{DD} .

The terminal B pin is not available on the MCP4018 and MCP4019 devices. For these devices, the potentiometer's terminal B is internally connected to V_{SS} .

3.6 Potentiometer Wiper (W) Terminal

The terminal W pin is connected to the internal potentiometer's terminal W (the wiper). The wiper terminal is the adjustable terminal of the digital potentiometer. The terminal W pin does not have a polarity relative to terminals A or B pins. The terminal

W pin can support both positive and negative currents. The voltage on terminal W must be between V_{SS} and V_{DD}.

3.7 Potentiometer Terminal A

The terminal A pin (available on some devices) is connected to the internal potentiometer's terminal A.

The potentiometer's terminal A is the fixed connection to the Full Scale (0x7F tap) wiper value of the digital potentiometer.

The terminal A pin is available on the MCP4018 devices. The terminal A pin does not have a polarity relative to the terminal W pin. The terminal A pin can support both positive and negative current. The voltage on Terminal A must be between V_{SS} and V_{DD} .

The terminal A pin is not available on the MCP4017 and MCP4019 devices. For these devices, the potentiometer's terminal A is internally floating.

4.0 GENERAL OVERVIEW

The MCP4017/18/19 devices are general purpose digital potentiometers intended to be used in applications where a programmable resistance with moderate bandwidth is desired.

This data sheet covers a family of three digital potentiometer and rheostat devices. The MCP4018 device is configured as a potentiometer, while the MCP4017 and MCP4019 devices are configured as rheostats.

Applications generally suited for the MCP401X devices include:

- Set point or offset trimming
- Sensor calibration
- Selectable gain and offset amplifier designs
- · Cost-sensitive mechanical trim pot replacement

As the **Device Block Diagram** shows, there are four main functional blocks. These are:

- POR/BOR Operation
- Serial Interface I2C Module
- Resistor Network

The POR/BOR operation and the Memory Map are discussed in this section and the I^2C and Resistor Network operation are described in their own sections. The **Serial Commands** commands are discussed in **Section 5.4**.

4.1 POR/BOR Operation

The Power-on Reset occurs when the device has power applied to it from V_{SS} . The Brown-out Reset occurs when a device has power applied to it, and that power (voltage) drops below the specified range.

The device's RAM retention voltage (V_{RAM}) is lower than the POR/BOR voltage trip point (V_{POR}/V_{BOR}). The maximum V_{POR}/V_{BOR} voltage is less then 1.8V.

When $V_{POR}/V_{BOR} < V_{DD} < 2.7V$, the Resistor Network's electrical performance may not meet the data sheet specifications. In this region, the device is capable of reading and writing to its volatile memory if the proper serial command is executed.

Table 4-1 shows the digital pot's level of functionality across the entire V_{DD} range, while Figure 4-1 illustrates the Power-up and Brown-out functionality.

4.1.1 POWER-ON RESET

When the device powers up, the device V_{DD} will cross the V_{POR}/V_{BOR} voltage. Once the V_{DD} voltage crosses the V_{POR}/V_{BOR} voltage, the following happens:

- Volatile wiper register is loaded with the default wiper value (3Fh)
- The device is capable of digital operation

4.1.2 BROWN-OUT RESET

When the device powers down, the device V_{DD} will cross the V_{POR}/V_{BOR} voltage. Once the V_{DD} voltage decreases below the V_{POR}/V_{BOR} voltage the following happens:

• Serial Interface is disabled.

If the V_{DD} voltage decreases below the V_{RAM} voltage the following happens:

· Volatile wiper registers may become corrupted.

As the voltage recovers above the V_{POR}/V_{BOR} voltage see Section 4.1.1 "Power-on Reset".

Serial commands not completed due to a Brown-out condition may cause the memory location to become corrupted.

4.1.3 WIPER REGISTER (RAM)

The wiper register is volatile memory that starts functioning at the RAM retention voltage (V_{RAM}). The wiper register will be loaded with the default wiper value when V_{DD} will rise above the V_{POR}/V_{BOR} voltage.

4.1.4 DEVICE CURRENTS

The current of the device can be classified into two modes of the device operation. These are:

- Serial Interface Inactive (Static Operation)
- Serial Interface Active

Static Operation occurs when a Stop condition is received. Static Operation is exited when a Start condition is received.

V _{DD} Level	Serial Interface	Potentiometer Terminals	Wiper Setting	Comment
V _{DD} < V _{BOR} < 1.8V	Ignored	"unknown"	Unknown	
$V_{BOR} \le V_{DD} < 1.8V$	"Unknown"	Operational with reduced electrical specs	Wiper register loaded with POR/BOR value	
$1.8V \le V_{DD} < 2.7V$	Accepted	Operational with reduced electrical specs	Wiper register determines wiper set- ting	Electrical performance may not meet the data sheet specifications.
$2.7V \le V_{DD} \le 5.5V$	Accepted	Operational	Wiper register determines wiper set- ting	Meets the data sheet specifications

TABLE 4-1: DEVICE FUNCTIONALITY AT EACH V_{DD} REGION (Note 1)

Note 1: For system voltages below the minimum operating voltage, it is recommended that the customer use a voltage supervisor to hold the system in reset. This will ensure that MCP4017/18/19 commands are not attempted out of the operating range of the device.

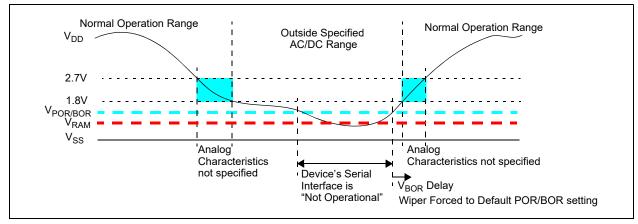


FIGURE 4-1:

Power-up and Brown-out.

5.0 SERIAL INTERFACE -I²C MODULE

A 2-wire I²C serial protocol is used to write or read the digital potentiometer's wiper register. The I²C protocol utilizes the SCL input pin and SDA input/output pin.

The I²C serial interface supports the following features.

- Slave mode of operation
- 7-bit addressing
- The following clock rate modes are supported:
 - Standard mode, bit rates up to 100 kb/s
- Fast mode, bit rates up to 400 kb/s
- Support Multi-Master Applications

The serial clock is generated by the Master.

The I^2C Module is compatible with the Phillips I^2C specification. Phillips only defines the field types, field lengths, timings, etc. of a frame. The frame *content* defines the behavior of the device. The frame content for the MCP4017, MCP4018, and MCP4019 devices are defined in this section of the data sheet.

Figure 5-1 shows a typical I²C bus configuration.

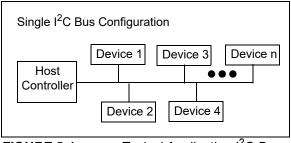


FIGURE 5-1: Configuration. *Typical Application I²C Bus*

Refer to **Section 2.0 "Typical Performance Curves"**, AC/DC Electrical Characteristics table for detailed input threshold and timing specifications.

5.1 I²C I/O Considerations

 $\rm I^2C$ specifications require active low, passive high functionality on devices interfacing to the bus. Since devices may be operating on separate power supply sources, ESD clamping diodes are not permitted. The specification recommends using open-drain transistors tied to V_{SS} (common) with a pull-up resistor. The specification makes some general recommendations on the size of this pull-up, but does not specify the exact value since bus speeds and bus capacitance impacts the pull-up value for optimum system performance.

Common pull-up values range from 1 k Ω to a max of ~10 k Ω . Power sensitive applications tend to choose higher values to minimize current losses during communication but these applications also typically utilize lower V_{DD}.

The SDA and SCL float (are not driving) when the device is powered down.

A "glitch" filter is on the SCL and SDA pins when the pin is an input. When these pins are an output, there is a slew rate control of the pin that is independent of device frequency.

5.1.1 SLOPE CONTROL

The device implements slope control on the SDA output. The slope control is defined by the Fast mode specifications.

For Fast (FS) mode, the device has spike suppression and Schmidt Trigger inputs on the SDA and SCL pins.

5.2 I²C Bit Definitions

I²C bit definitions include:

- Start Bit
- Data Bit
- Acknowledge (A) Bit
- Repeated Start Bit
- Stop Bit
- Clock Stretching

Figure 5-8 shows the waveform for these states.

5.2.1 START BIT

The Start bit (see Figure 5-2) indicates the beginning of a data transfer sequence. The Start bit is defined as the SDA signal falling when the SCL signal is "High".

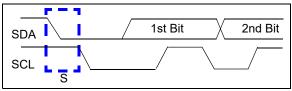
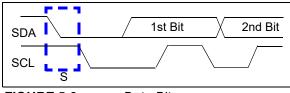


FIGURE 5-2: Start Bit.

5.2.2 DATA BIT

The SDA signal may change state while the SCL signal is Low. While the SCL signal is High, the SDA signal MUST be stable (see Figure 5-3).





5.2.3 ACKNOWLEDGE (A) BIT

The A bit (see Figure 5-4) is a response from the Slave device to the Master device. Depending on the context of the transfer sequence, the A bit may indicate different things. Typically the Slave device will supply an A response after the Start bit and 8 "data" bits have been received. The A bit will have the SDA signal low.

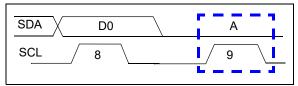


FIGURE 5-4:

Acknowledge Waveform.

If the Slave Address is not valid, the Slave Device will issue a Not A (\overline{A}). The \overline{A} bit will have the SDA signal high.

If an error condition occurs (such as an \overline{A} instead of A) then an START bit must be issued to reset the command state machine.

TABLE 5-1: MCP4017/18/19 A / A RESPONSES

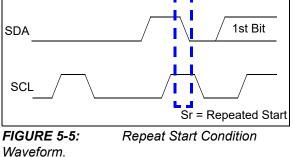
Event	Acknowledge Bit Response	Comment
General Call	Ā	
Slave Address valid	A	
Slave Address not valid	Ā	
Bus Collision	N.A.	I ² C Module Resets, or a "Don't Care" if the collision occurs on the Masters "Start bit".

5.2.4 REPEATED START BIT

The Repeated Start bit (see Figure 5-5) indicates the current Master Device wishes to continue communicating with the current Slave Device without releasing the l^2 C bus. The Repeated Start condition is the same as the Start condition, except that the Repeated Start bit follows a Start bit (with the Data bits + A bit) and not a Stop bit.

The Start bit is the beginning of a data transfer sequence and is defined as the SDA signal falling when the SCL signal is "High".

A bus collision during the Repeated Start condition occurs if:
•SDA is sampled low when SCL goes from low to high.
•SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data "1".
53



5.2.5 STOP BIT

The Stop bit (see Figure 5-6) indicates the end of the I^2C Data Transfer Sequence. The Stop bit is defined as the SDA signal rising when the SCL signal is "High".

A Stop bit resets the I²C interface of the other devices.

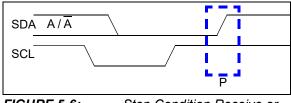


FIGURE 5-6:Stop Condition Receive orTransmit Mode.

5.2.6 CLOCK STRETCHING

"Clock Stretching" is something that the Secondary Device can do to allow additional time to "respond" to the "data" that has been received.

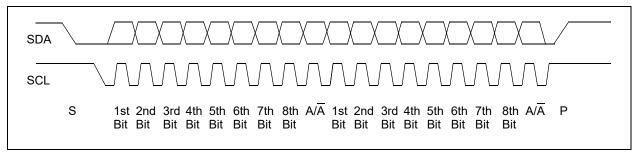
The MCP4017/18/19 will not stretch the clock signal (SCL) since memory read accesses occur fast enough.

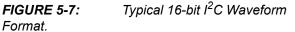
5.2.7 ABORTING A TRANSMISSION

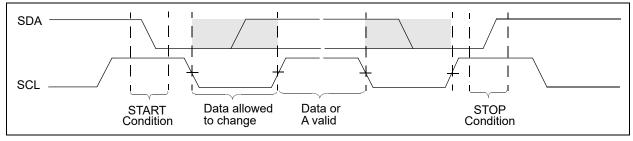
If any part of the l^2C transmission does not meet the command format, it is aborted. This can be intentionally accomplished with a START or STOP condition. This is done so that noisy transmissions (usually an extra START or STOP condition) are aborted before they corrupt the device.

5.2.8 IGNORING AN I²C TRANSMISSION AND "FALLING OFF" THE BUS

The MCP4017/18/19 expects to receive entire, valid I^2C commands and will assume any command not defined as a valid command is due to bus corruption and will enter a passive high condition on the SDA signal. All signals will be ignored until the next valid START condition and CONTROL BYTE are received.







*FIGURE 5-8: I*²*C* Data States and Bit Sequence.

5.2.9 I²C COMMAND PROTOCOL

The MCP4017/18/19 is a slave I^2C device which supports 7-bit slave addressing. The slave address contains seven fixed bits. Figure 5-9 shows the control byte format.

5.2.9.1 Control Byte (Slave Address)

The Control Byte is always preceded by a START condition. The Control Byte contains the slave address consisting of seven fixed bits and the R/\overline{W} bit. Figure 5-9 shows the control byte format and Table 5-2 shows the l²C address for the devices.

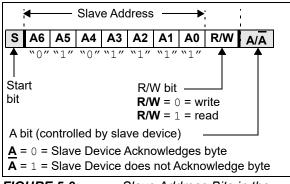


FIGURE 5-9: Slave Address Bits in the I²C Control Byte.

TABLE 5-2:DEVICE I²C ADDRESS

Device	I ² C Address	Comment
MCP4017	'0101111'	
MCP4018	'0101111'	
MCP4019	'0101111'	

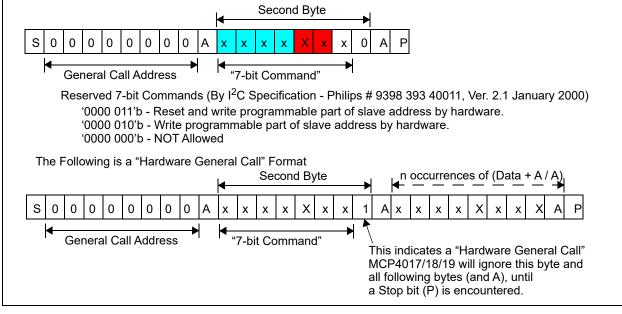
5.2.9.2 Hardware Address Pins

The MCP4017/MCP4018/MCP4019 does not support hardware address bits.

5.2.10 GENERAL CALL

The General Call is a method by which the Master device can communicate with all other Slave devices.

The MCP4017/18/19 devices do not respond to General Call addresses and commands, and therefore the communications are Not Acknowledged.





5.3 Software Reset Sequence

Note:	This technique should be supported by			
	any I ² C compliant device. The 24xxxx I ² C			
	Serial EEPROM devices support this tech-			
	nique, which is documented in AN1028.			

At times it may become necessary to perform a Software Reset Sequence to ensure the MCP4017/18/ 19 device is in a correct and known I^2C Interface state. This only resets the I^2C state machine.

This is useful if the MCP4017/18/19 device powers up in an incorrect state (due to excessive bus noise, etc), or if the Master Device is reset during communication. Figure 5-11 shows the communication sequence to software reset the device.

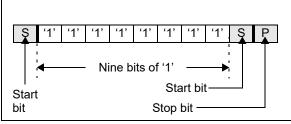


FIGURE 5-11: Software Reset Sequence Format.

The first Start bit will cause the device to reset from a state in which it is expecting to receive data from the Master Device. In this mode, the device monitors the data bus in Receive mode and can detect the Start bit, which forces an internal Reset.

The nine bits of '1' are used to force a Reset of those devices that could not be reset by the previous Start bit. This occurs only if the MCP4017/18/19 is driving an A on the l^2 C bus, or is in Output mode (from a Read command) and is driving a data bit of '0' onto the l^2 C bus. In both of these cases, the previous Start bit could not be generated due to the MCP4017/18/19 holding the bus low. By sending out nine '1' bits, it is ensured that the device will see a \overline{A} (the Master Device does not drive the l^2 C bus low to acknowledge the data sent by the MCP4017/18/19), which also forces the MCP4017/18/19 to reset.

The second Start bit is sent to address the rare possibility of an erroneous write. This could occur if the Master Device was reset while sending a Write command to the MCP4017/18/19, AND then as the Master Device returns to normal operation and issues a Start condition while the MCP4017/18/19 is issuing an A. In this case if the second Start bit is not sent (and the Stop bit was sent) the MCP4017/18/19 could initiate a write cycle.

Note: The potential for this erroneous write ONLY occurs if the Master Device is reset while sending a Write command to the MCP4017/18/19. The Stop bit terminates the current I^2C bus activity. The MCP4017/18/19 wait to detect the next Start condition.

This sequence does not effect any other I²C devices which may be on the bus, as they should disregard this as an invalid command.

5.4 Serial Commands

The MCP4017/18/19 devices support two serial commands. These commands are:

- Write Operation
- Read Operation

5.4.1 WRITE OPERATION

The write operation requires the START condition, Control Byte, Acknowledge, Data Byte, Acknowledge and STOP (or RESTART) condition. The Control (Slave Address) Byte requires the R/W bit equal to a logic zero (R/W = "0") to generate a write sequence. The MCP4017/18/19 is responsible for generating the Acknowledge (A) bits.

Data is written to the MCP4017/18/19 after every byte transfer (during the A bit). If a STOP or RESTART condition is generated during a data transfer (before the A bit), the data will not be written to MCP4017/18/19.

Data bytes may be written after each Acknowledge. The command is terminated once a Stop (P) condition occurs. Refer to Figure 5-12 for the write sequence. For a single byte write, the master sends a STOP or RESTART condition after the 1st data byte is sent.

The MSb of each Data Byte is a don't care, since the wiper register is only 7-bits wide.

Figure 5-14 shows the I^2C communication behavior of the Master Device and the MCP4017/18/19 device and the resultant I^2C bus values.

5.4.2 READ OPERATION

The read operation requires the START condition, Control Byte, Acknowledge, Data Byte, the master generating the \overline{A} and STOP condition. The Control Byte requires the R/ \overline{W} bit equal to a logic one (R/ \overline{W} = 1) to generate a read sequence. The MCP4017/18/19 will A the Slave Address Byte and \overline{A} all the Data Bytes. The I²C Master will \overline{A} the Slave Address Byte and the last Data Byte. If there are multiple Data Bytes, the I²C Master will A all Data Bytes except the last Data Byte (which it will \overline{A}).

The MCP4017/18/19 maintains control of the SDA signal until all data bits have been clocked out.

The command is terminated once a Stop (P) condition occurs. Refer to Figure 5-13 for the read command sequence. For a single read, the master sends a STOP or RESTART condition after the 1st data byte (and A bit) is sent from the slave.

Figure 5-14 shows the I^2C communication behavior of the Master Device and the MCP4017/18/19 device and the resultant I^2C bus values.

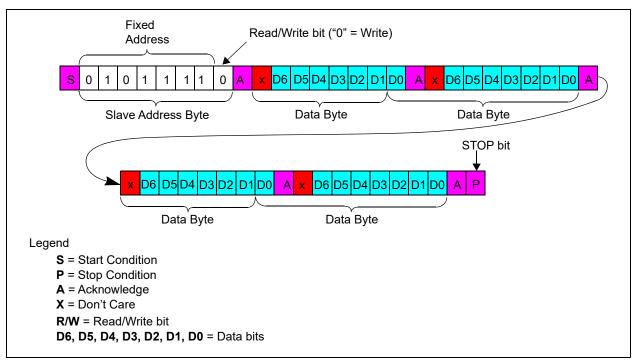
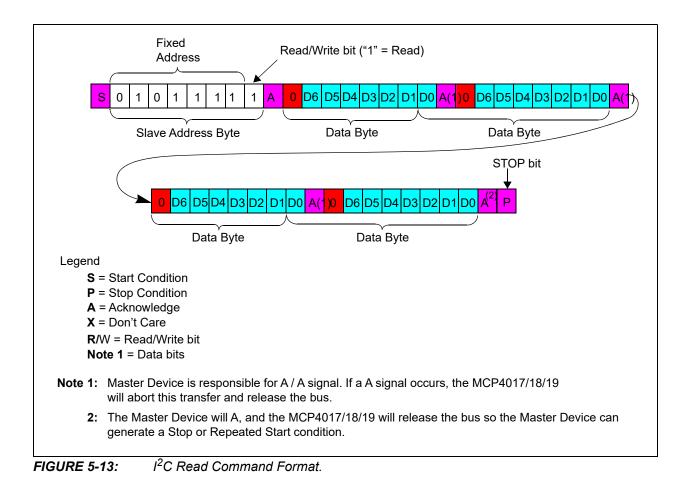


FIGURE 5-12: I²C Write Command Format.



MCP4017/18/19

Write 1 Byte	
write i byte	
	S Slave Address W A Data Byte (1) A P
Master	S 0 1 0 1 1 1 0 1 x d d d d d d 1 P
MCP4017/18/19	0
I ² C Bus	S 0 1 0 1 1 1 1 0 0 x d d d d d d 0 P
Write 2 Bytes	
	R
	S Slave Address W A Data Byte (1) A Data Byte (1) A P
Master	S 0 1 0 1 1 1 1 0 1 x d d d d d d d d d d d d d d d d d P
MCP4017/18/19	0 0 0
I ² C Bus	S 0 1 0 1 1 1 1 0 0 x d d d d d d d d d d d d d d d d d
Read 1 Byte	
	S Slave Address W A Data Byte A P
Master	S O I O I I I S O I I I I I
MCP4017/18/19	0 0 d d d d d d 1
l ² C Bus	S 0 1 0 1 1 1 1 1 0 0 d d d d d d 1 P
Read 2 Bytes	
	S Slave Address R / A Data Byte A Data Byte A P
Master	S 0 1 0 1 1 1 1
MCP4017/18/19	0 0 d d d d d d d 1 0 d d d d d d 1 1 0 d d d 1 1 1 0 0 0 0
I ² C Bus	S 0 1 0 1 1 1 1 1 0 0 d d d d d d d d d d d d d d d d d
Note 1: For W	/rite Commands, the MSb of the Data Byte is a don't care since the wiper register is only 7-bits wide.
	l^2



6.0 **RESISTOR NETWORK**

The Resistor Network is made up of two parts. These are:

- Resistor Ladder
- Wiper

Figure 6-1 shows a block diagram for the resistive network.

Digital potentiometer applications can be divided into two resistor network categories:

- Rheostat configuration
- Potentiometer (or voltage divider) configuration

The MCP4017 is a true rheostat, with terminal B and the wiper (W) of the variable resistor available on pins.

The MCP4018 device offers a voltage divider (potentiometer) with terminal B internally connected to ground.

The MCP4019 device is a rheostat device with terminal A of the resistor floating, terminal B internally connected to ground, and the wiper (W) available on pin.

6.1 Resistor Ladder Module

The resistor ladder is a series of equal value resistors (R_S) with a connection point (tap) between the two resistors. The total number of resistors in the series (ladder) determines the R_{AB} resistance (see Figure 6-1). The end points of the resistor ladder are connected to the device terminal A and terminal B pins. The R_{AB} (and R_S) resistance has small variations over voltage and temperature.

The Resistor Network has 127 resistors in a string between terminal A and terminal B. This gives 7-bits of resolution.

The wiper can be set to tap onto any of these 127 resistors, thus providing 128 possible settings (including terminal A and terminal B). This allows zero scale to full scale connections.

A wiper setting of 00h connects the terminal W (wiper) to terminal B (Zero Scale). A wiper setting of 3Fh is the mid-scale setting. A wiper setting of 7Fh connects the terminal W (wiper) to terminal A (Full Scale). Table 6-1 illustrates the full wiper setting map.

Terminal A and B, as well as the wiper W, do not have a polarity. These terminals can support both positive and negative current.

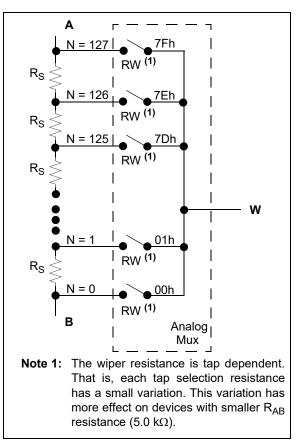


FIGURE 6-1: Resistor Network Block Diagram.

Wiper Setting	Properties
07Fh	Full Scale (W = A)
07Eh - 040h	W = N
03Fh	W = N (Mid-Scale)
03Eh - 001h	W = N
000h	Zero Scale (W = B)

Step resistance (R_S) is the resistance from one tap setting to the next. This value will be dependent on the R_{AB} value that has been selected. Equation 6-1 shows the calculation for the step resistance while Table 6-2 shows the typical step resistances for each device.

EQUATION 6-1: R_S CALCULATION

$$R_S = \frac{R_{AB}}{127}$$

Equation 6-2 illustrates the calculation used to determine the resistance between the wiper and terminal B.

EQUATION 6-2: R_{WB} CALCULATION $R_{WB} = \frac{R_{AB}N}{127} + R_{W}$

 $\mathbf{N} = 0 \text{ to } 127 + \mathbf{K}_W$ $\mathbf{N} = 0 \text{ to } 127 \text{ (decimal)}$

The digital potentiometer is available in four nominal resistances (R_{AB}) where the nominal resistance is defined as the resistance between terminal A and terminal B. The four nominal resistances are 5 kΩ, 10 kΩ, 50 kΩ, and 100 kΩ.

The total resistance of the device has minimal variation due to operating voltage (see Figure 2-11, Figure 2-29, Figure 2-47, or Figure 2-65).

	Resistance (Ω)				
Part Number	Case	Total (R _{AB})	Step (R _S)		
	Min.	4000	31.496		
MCP4017/18/19-502E	Typical	5000	39.370		
	Max.	6000	47.244		
	Min.	8000	62.992		
MCP4017/18/19-103E	Typical	10000	78.740		
	Max.	12000	94.488		
	Min.	40000	314.961		
MCP4017/18/19-503E	Typical	50000	393.701		
	Max.	60000	472.441		
	Min.	80000	629.921		
MCP4017/18/19-104E	Typical	100000	787.402		
	Max.	120000	944.882		

TABLE 6-2: STEP RESISTANCES

A POR/BOR event will load the volatile wiper register [memory] location with the default value. Table 6-3 shows the default values offered.

TABLE 6-3:	DEFAULT FACTORY
	SETTINGS SELECTION

Resistance	Typical	Default POR Wiper			
Code	R _{AB} Value	Setting	Code ⁽¹⁾		
-502	5.0 kΩ	Mid-scale	3Fh		
-103	10.0 kΩ	Mid-scale	3Fh		
-503	50.0 kΩ	Mid-scale	3Fh		
-104	100.0 kΩ	Mid-scale	3Fh		

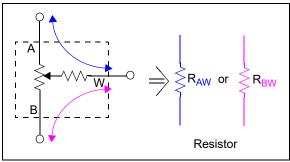
Note 1: Custom POR/BOR Wiper Setting options are available, contact the local Microchip Sales Office for additional information. Custom options have minimum volume requirements.

6.2 **Resistor Configurations**

6.2.1 RHEOSTAT CONFIGURATION

When used as a rheostat, two of the three digital potentiometer's terminals are used as a resistive element in the circuit. With terminal W (wiper) and either terminal A or terminal B, a variable resistor is created. The resistance will depend on the tap setting of the wiper (and the wiper's resistance). The resistance is controlled by changing the wiper setting

The unused terminal (B or A) should be left floating. Figure 6-2 shows the two possible resistors that can be used. Reversing the polarity of the A and B terminals will not affect operation.





Rheostat Configuration.

This allows the control of the total resistance between the two nodes. The total resistance depends on the "starting" terminal to the wiper terminal. So at the code 00h, the R_{BW} resistance is minimal (R_W), but the R_{AW} resistance in maximized (R_{AB} + R_W). Conversely, at the code 3Fh, the R_{AW} resistance is minimal (R_W), but the R_{BW} resistance in maximized (R_{AB} + R_W).

The resistance Step size $(\ensuremath{\mathsf{R}}_S)$ equates to one LSb of the resistor.

Note:	To avoid damage to the internal wiper						
	circuitry in this configuration, care should						
	be taken to ensure the current flow never						
	exceeds 2.5 mA.						

The pinout for the rheostat devices is such that as the wiper register is incremented, the resistance of the resistor will increase (as measured from terminal B to the terminal W).

6.2.2 POTENTIOMETER CONFIGURATION

When used as a potentiometer, all three terminals of the device are tied to different nodes in the circuit. This allows the potentiometer to output a voltage proportional to the input voltage. This configuration is sometimes called voltage divider mode. The potentiometer is used to provide a variable voltage by adjusting the wiper position between the two endpoints as shown in Figure 6-3. Reversing the polarity of the A and B terminals will not affect operation.

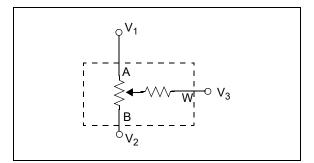
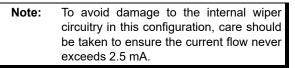


FIGURE 6-3: Configuration.

Potentiometer

The temperature coefficient of the R_{AB} resistors is minimal by design. In this configuration, the resistors all change uniformly, so minimal variation should be seen.

The wiper resistor temperature coefficient is different from the R_{AB} temperature coefficient. The voltage at node V3 (Figure 6-3) is not dependent on this wiper resistance, just the ratio of the R_{AB} resistors, so this temperature coefficient in most cases can be ignored.



6.3 Wiper Resistance

Wiper resistance is the series resistance of the analog switch that connects the selected resistor ladder node to the Wiper Terminal common signal (see Figure 6-1).

A value in the volatile wiper register selects which analog switch to close, connecting the W terminal to the selected node of the resistor ladder.

The resistance is dependent on the voltages on the analog switch source, gate, and drain nodes, as well as the device's wiper code, temperature, and the current through the switch. As the device voltage decreases, the wiper resistance increases (see Figure 6-4 and Table 6-4).

The wiper can connect directly to terminal B or to terminal A. A zero scale connection connects the terminal W (wiper) to terminal B (wiper setting of 000h). A full scale connection connects the terminal W (wiper) to terminal A (wiper setting of 7Fh). In these configurations, the only resistance between the terminal W and the other terminal (A or B) is that of the analog switches.

The wiper resistance is typically measured when the wiper is positioned at either zero scale (00h) or full scale (3Fh).

The wiper resistance in potentiometer-generated voltage divider applications is not a significant source of error.

The wiper resistance in rheostat applications can create significant nonlinearity as the wiper is moved toward zero scale (00h). The lower the nominal resistance, the greater the possible error.

In a rheostat configuration, this change in voltage needs to be taken into account. Particularly for the lower resistance devices. For the 5.0 k Ω device the maximum wiper resistance at 5.5V is approximately 3.2% of the total resistance, while at 2.7V it is approximately 6.5% of the total resistance.

In a potentiometer configuration, the wiper resistance variation does not effect the output voltage seen on the W pin.

The slope of the resistance has a linear area (at the higher voltages) and a non-linear area (at the lower voltages). In cases where resistance increases faster, then the voltage drops (at low voltages).

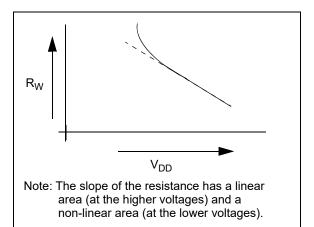


FIGURE 6-4: Relationship of Wiper Resistance (R_W) to Voltage.

Since there is minimal variation of the total device resistance over voltage, at a constant temperature (see Figure 2-11, Figure 2-29, Figure 2-47, or Figure 2-65), the change in wiper resistance over voltage can have a significant impact on the INL and DNL error.

Resistance (Ω)					R _W / R _S (%) ⁽ 1 ⁾			R _W / R _{AB} (%) ⁽ 2 ⁾		
Typical Wiper (R _W)		D -			в –					
Total (R _{AB})	Step (R _S)	Typical	Max @ 5.5V	Max @ 2.7V	R _W = Typical	R _W = Max @ 5.5V	R _W = Max @ 2.7V	R _W = Typical	R _W = Max @ 5.5V	R _W = Max @ 2.7V
5000	39.37	100	170	325	254.00%	431.80%	825.5%	2.00%	3.40%	6.50%
10000	78.74	100	170	325	127.00%	215.90%	412.75%	1.00%	1.70%	3.25%
50000	393.70	100	170	325	25.40%	43.18%	82.55%	0.20%	0.34%	0.65%
100000	787.40	100	170	325	12.70%	21.59%	41.28%	0.10%	0.17%	0.325%

TABLE 6-4: TYPICAL STEP RESISTANCES AND RELATIONSHIP TO WIPER RESISTANCE

Note 1: R_S is the typical value. The variation of this resistance is minimal over voltage.

2: R_{AB} is the typical value. The variation of this resistance is minimal over voltage.

6.4 **Operational Characteristics**

Understanding the operational characteristics of the device's resistor components is important to the system design.

6.4.1 ACCURACY

6.4.1.1 Integral Non-linearity (INL)

INL error for these devices is the maximum deviation between an actual code transition point and its corresponding ideal transition point after offset and gain errors have been removed. These endpoints are from 0x00 to 0x7F. Refer to Figure 6-5.

Positive INL means higher resistance than ideal. Negative INL means lower resistance than ideal.

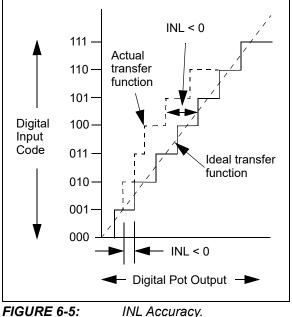
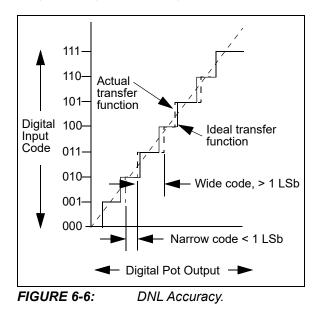


FIGURE 6-5:

6.4.1.2 Differential Non-linearity (DNL)

DNL error is the measure of variations in code widths from the ideal code width. A DNL error of zero would imply that every code is exactly 1 LSb wide.



6.4.1.3 Ratiometric temperature coefficient

The ratiometric temperature coefficient quantifies the error in the ratio $R_{AW}\!/R_{WB}$ due to temperature drift. This is typically the critical error when using a potentiometer device (MCP4018) in a voltage divider configuration.

6.4.1.4 Absolute temperature coefficient

The absolute temperature coefficient quantifies the error in the end-to-end resistance (Nominal resistance RAB) due to temperature drift. This is typically the critical error when using a rheostat device (MCP4017 and MCP4019) in an adjustable resistor configuration.

6.4.2 MONOTONIC OPERATION

Monotonic operation means that the device's resistance increases with every step change (from terminal A to terminal B or terminal B to terminal A).

The wiper resistance differs at each tap location. When changing from one tap position to the next (either increasing or decreasing), the ΔR_W is less then the ΔR_S . When this change occurs, the device voltage and temperature are "the same" for the two tap positions.

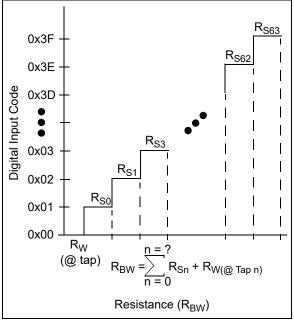


FIGURE 6-7: R_{BW}.

7.0 DESIGN CONSIDERATIONS

In the design of a system with the MCP4017/18/19 devices, the following considerations should be taken into account. These are:

- The Power Supply
- · The Layout

In the design of a system with the MCP4017/18/19 devices, the following considerations should be taken into account:

- Power Supply Considerations
- Layout Considerations

7.1 Power Supply Considerations

The typical application will require a bypass capacitor in order to filter high-frequency noise, which can be induced onto the power supply's traces. The bypass capacitor helps to minimize the effect of these noise sources on signal integrity. Figure 7-1 illustrates an appropriate bypass strategy.

In this example, the recommended bypass capacitor value is 0.1 μ F. This capacitor should be placed as close to the device power pin (V_{DD}) as possible (within 4 mm).

The power source supplying these devices should be as clean as possible. If the application circuit has separate digital and analog power supplies, V_{DD} and V_{SS} should reside on the analog plane.

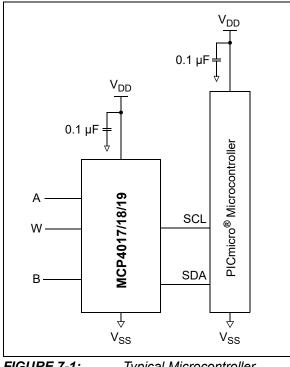


FIGURE 7-1: Typical Microcontroller Connections.

7.2 Layout Considerations

Inductively-coupled AC transients and digital switching noise can degrade the input and output signal integrity, potentially masking the MCP4017/18/19's performance. Careful board layout will minimize these effects and increase the Signal-to-Noise Ratio (SNR). Bench testing has shown that a multi-layer board utilizing a low-inductance ground plane, isolated inputs, isolated outputs and proper decoupling are critical to achieving the performance that the silicon is capable of providing. Particularly harsh environments may require shielding of critical signals.

If low noise is desired, breadboards and wire-wrapped boards are not recommended.

7.2.1 RESISTOR TEMPCO

Characterization curves of the resistor temperature coefficient (Tempco) are shown in Figure 2-11, Figure 2-29, Figure 2-47, and Figure 2-65.

These curves show that the resistor network is designed to correct for the change in resistance as temperature increases. This technique reduces the end-to-end change in R_{AB} resistance.

MCP4017/18/19

NOTES:

8.0 APPLICATIONS EXAMPLES

Digital potentiometers have a multitude of practical uses in modern electronic circuits. The most popular uses include precision calibration of set point thresholds, sensor trimming, LCD bias trimming, audio attenuation, adjustable power supplies, motor control overcurrent trip setting, adjustable gain amplifiers and offset trimming. The MCP4017/18/19 devices can be used to replace the common mechanical trim pot in applications where the operating and terminal voltages are within CMOS process limitations ($V_{DD} = 2.7V$ to 5.5V).

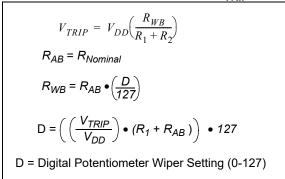
8.1 Set Point Threshold Trimming

Applications that need accurate detection of an input threshold event often need several sources of error eliminated. Use of comparators and operational amplifiers (op amps) with low offset and gain error can help achieve the desired accuracy, but in many applications, the input source variation is beyond the designer's control. If the entire system can be calibrated after assembly in a controlled environment (like factory test), these sources of error are minimized if not entirely eliminated.

Figure 8-1 illustrates a common digital potentiometer configuration. This configuration is often referred to as a "windowed voltage divider". Note that R₁ is not necessary to create the voltage divider, but its presence is useful when the desired threshold has limited range. It is "windowed" because R₁ can narrow the adjustable range of V_{TRIP} to a value much less than $V_{DD} - V_{SS}$. If the output range is reduced, the magnitude of each output step is reduced. This effectively increases the trimming resolution for a fixed digital potentiometer resolution. This technique may allow a lower-cost digital potentiometer to be utilized (64 steps instead of 256 steps).

The MCP4018's low DNL performance is critical to meeting calibration accuracy in production without having to use a higher precision digital potentiometer.

EQUATION 8-1: CALCULATING THE WIPER SETTING FROM THE DESIRED V_{TRIP}



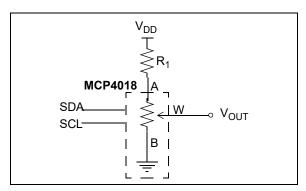


FIGURE 8-1:Using the DigitalPotentiometer to Set a Precise Output Voltage.

8.1.1 TRIMMING A THRESHOLD FOR AN OPTICAL SENSOR

If the application has to calibrate the threshold of a diode, transistor or resistor, a variation range of 0.1V is common. Often, a desired resolution of 2 mV or better is adequate to accurately detect the presence of a precise signal. A "windowed" voltage divider, utilizing the MCP4018, would be a potential solution. Figure 8-2 illustrates this example application.

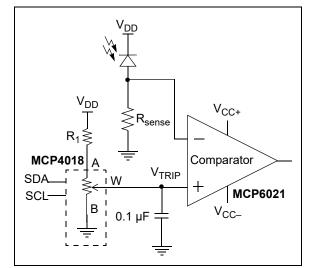


FIGURE 8-2: Set Point or Threshold Calibration.

8.2 Operational Amplifier Applications

Figure 8-3 and Figure 8-4 illustrate typical amplifier circuits that could replace fixed resistors with the MCP4017/18/19 to achieve digitally-adjustable analog solutions.

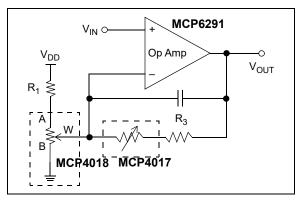


FIGURE 8-3: Trimming Offset and Gain in a Non-Inverting Amplifier.

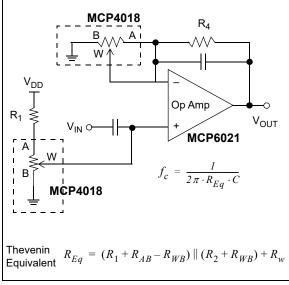


FIGURE 8-4: Programmable Filter.

8.3 Temperature Sensor Applications

Thermistors are resistors with very predictable variation in temperature. Thermistors are a popular sensor choice when a low-cost temperature-sensing solution is desired. Unfortunately, thermistors have non-linear characteristics that are undesirable, typically requiring trimming in an application to achieve greater accuracy. There are several common solutions to trim and linearize thermistors. Figure 8-5 and Figure 8-6 are simple methods for linearizing a 3-terminal NTC thermistor. Both are simple voltage dividers using a Positive Temperature Coefficient (PTC) resistor (R₁) with a transfer function capable of compensating for the linearity error in the Negative Temperature Coefficient (NTC) thermistor.

The circuit, illustrated by Figure 8-5, utilizes a digital rheostat for trimming the offset error caused by the thermistor's part-to-part variation. This solution puts the digital potentiometer's R_W into the voltage divider calculation. The MCP4017/18/19's R_{AB} temperature coefficient is a low 50 ppm (-20°C to +70°C). R_W 's error is substantially greater than R_{AB} 's error because R_W varies with V_{DD} , wiper setting and temperature. For the 50 k Ω devices, the error introduced by R_W is, in most cases, insignificant as long as the wiper setting is > 6. For the 2 k Ω devices, the error introduced by R_W is significant because it is a higher percentage of R_{WB} . For these reasons, the circuit illustrated in Figure 8-5 is not the most optimum method for "exciting" and linearizing a thermistor.

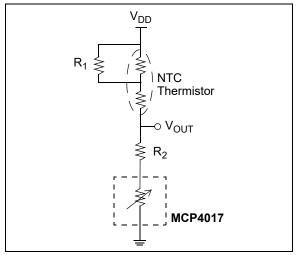


FIGURE 8-5: Thermistor Calibration using a Digital Potentiometer in a Rheostat Configuration.

The circuit illustrated by Figure 8-6 utilizes a digital potentiometer for trimming the offset error. This solution removes R_W from the trimming equation along with the error associated with R_W . R_2 is not required, but can be utilized to reduce the trimming "window" and reduce variation due to the digital pot's R_{AB} part-to-part variability.

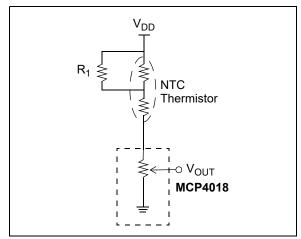


FIGURE 8-6: Thermistor Calibration using a Digital Potentiometer in a Potentiometer Configuration.

8.4 Wheatstone Bridge Trimming

Another common configuration to "excite" a sensor (such as a strain gauge, pressure sensor or thermistor) is the wheatstone bridge configuration. The wheatstone bridge provides a differential output instead of a single-ended output. Figure 8-7 illustrates a wheatstone bridge utilizing one to three digital potentiometers. The digital potentiometers in this example are used to trim the offset and gain of the wheatstone bridge.

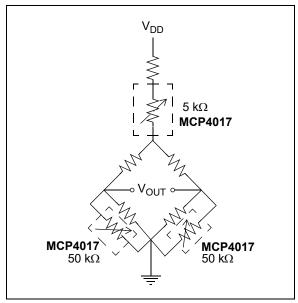


FIGURE 8-7: Wheatstone Bridge Trimming.

9.0 DEVELOPMENT SUPPORT

9.1 Development Tools

The MCP4017/18/19 devices can be evaluated with the MCP401X Evaluation Board. Please check the Microchip website for more information about this board.

TABLE 9-1: TECHNICAL DOCUMENTATION

9.2 Technical Documentation

Several additional technical documents are available to assist you in your design and development. These technical documents include Application Notes, Technical Briefs, and Design Guides. Table 9-1 shows some of these documents.

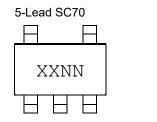
Application Note Number	Title	Literature #
AN1080	Understanding Digital Potentiometers Resistor Variations	DS01080
AN737	Using Digital Potentiometers to Design Low Pass Adjustable Filters	DS00737
AN692	Using a Digital Potentiometer to Optimize a Precision Single Supply Photo Detect	DS00692
AN691	Optimizing the Digital Potentiometer in Precision Circuits	DS00691
AN219	Comparing Digital Potentiometers to Mechanical Potentiometers	DS00219
_	Digital Potentiometer Design Guide	DS22017
_	Signal Chain Design Guide	DS21825

MCP4017/18/19

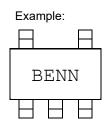
NOTES:

10.0 PACKAGING INFORMATION

10.1 Package Marking Information



Part Number	Code
MCP4019T-502E/LT	BENN
MCP4019T-103E/LT	BFNN
MCP4019T-503E/LT	BGNN
MCP4019T-104E/LT	BHNN



6-Lead SC70

Part Number	Code	Part Number	Code
MCP4017T-502E/LTY	AENN	MCP4018T-502E/LTY	AANN
MCP4017T-103E/LTY	AFNN	MCP4018T-103E/LTY	ABNN
MCP4017T-503E/LTY	AGNN	MCP4018T-503E/LTY	ACNN
MCP4017T-104E/LTY	AHNN	MCP4018T-104E/LTY	ADNN

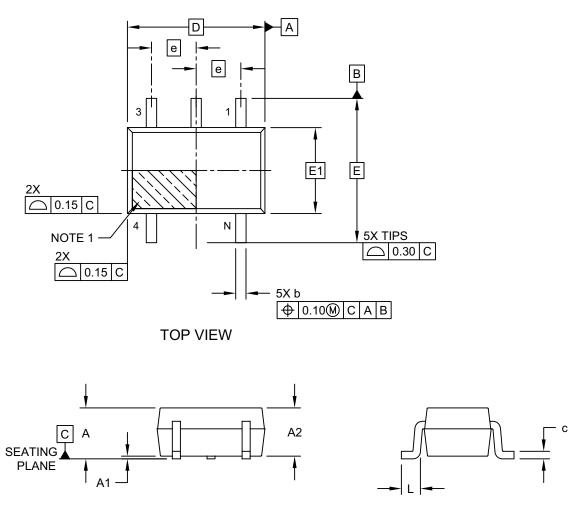




Legend	: XXX Y YY WW NNN @3 *	Customer-specific information Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code Pb-free JEDEC designator for Matte Tin (Sn) This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.			
Note:	In the event the full Microchip part number cannot be marked on one line, it wil be carried over to the next line, thus limiting the number of available characters for customer-specific information.				

5-Lead Plastic Small Outline Transistor (LTY) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



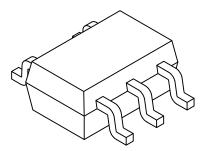
SIDE VIEW

END VIEW

Microchip Technology Drawing C04-061-LTY Rev E Sheet 1 of 2

5-Lead Plastic Small Outline Transistor (LTY) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Number of Pins	Ν	5		
Pitch	е	0.65 BSC		
Overall Height	Α	0.80 - 1.10		
Standoff	A1	0.00	-	0.10
Molded Package Thickness	A2	0.80	-	1.00
Overall Length	D	2.00 BSC		
Overall Width	Е	2.10 BSC		
Molded Package Width	E1	1.25 BSC		
Terminal Width	b	0.15	-	0.40
Terminal Length	L	0.10	0.20	0.46
Lead Thickness	С	0.08 - 0.26		

Notes:

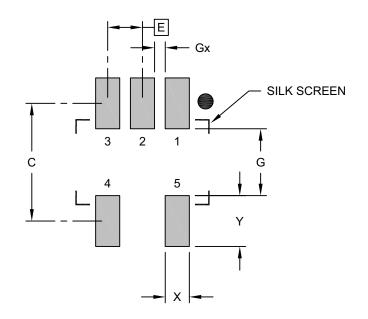
- 1. Pin 1 visual index feature may vary, but must be located within the hatched area.
- 2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- 3. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-061-LTY Rev E Sheet 2 of 2

5-Lead Plastic Small Outline Transistor (LTY) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	Units		MILLIMETERS		
Dimensio	Dimension Limits		NOM	MAX	
Contact Pitch E 0.65 BSC					
Contact Pad Spacing	С		2.20		
Contact Pad Width	Х			0.45	
Contact Pad Length	Y			0.95	
Distance Between Pads	G	1.25			
Distance Between Pads	Gx	0.20			

Notes:

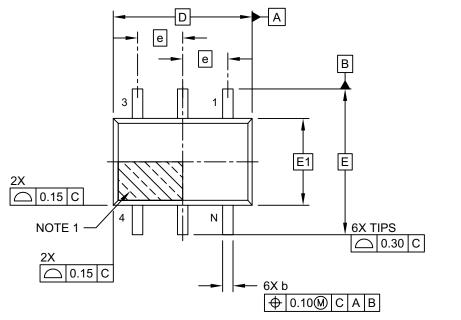
1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

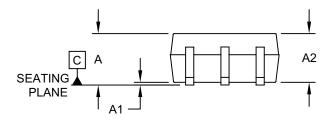
Microchip Technology Drawing No. C04-2061-LTY Rev E

6-Lead Plastic Small Outline Transistor (LT) [SC70]

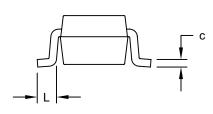
Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



TOP VIEW



SIDE VIEW

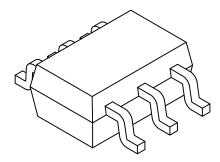


END VIEW

Microchip Technology Drawing C04-151B Sheet 1 of 2

6-Lead Plastic Small Outline Transistor (LT) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Number of Pins	Ν	6		
Pitch	е	0.65 BSC		
Overall Height	Α	0.80 - 1.10		
Standoff	A1	0.00	-	0.10
Molded Package Thickness	A2	0.80	0.90	1.00
Overall Length	D	2.00 BSC		
Exposed Pad Length	D2	2.50 2.60 2.70		2.70
Overall Width	Е	2.10 BSC		
Exposed Pad Width	E1	1.25 BSC		
Terminal Width	b	0.15	-	0.30
Terminal Length	L	0.10 0.20 0.46		0.46
Lead Thickness	c 0.08 - 0.22			0.22

Notes:

1. Pin 1 visual index feature may vary, but must be located within the hatched area.

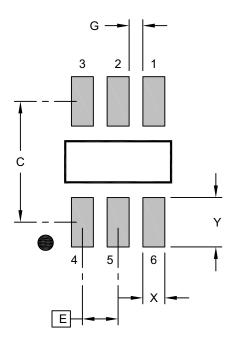
2. Dimensions D and E1 do not include mold flash or protrusions. Mold flash or

- protrusions shall not exceed 0.15mm per side.
- Dimensioning and tolerancing per ASME Y14.5M BSC: Basic Dimension. Theoretically exact value shown without tolerances. REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-151B Sheet 2 of 2

6-Lead Plastic Small Outline Transistor (LT) [SC70]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



RECOMMENDED LAND PATTERN

	MILLIMETERS			
Dimension Limits		MIN	NOM	MAX
Contact Pitch E		0.65 BSC		
Contact Pad Spacing	С		2.20	
Contact Pad Width (X6)	Х			0.40
Contact Pad Length (X6)	Y			0.90
Distance Between Pads	G	0.25		

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing C04-2151B

MCP4017/18/19

NOTES:

APPENDIX A: REVISION HISTORY

Revision B (September 2024)

- Part number change from MCP4017T-XXXE/LT to MCP4017T-XXXE/LTY and from MCP4018T-XXXE/LT to MCP4018T-XXXE/LTY to improve manufacturability;
- Updated Section 10.0 "Packaging Information";
- Updated the Product Identification System section.

Revision A (March 2009)

• Original Release of this Document.

MCP4017/18/19

NOTES:

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	<u>XXX X XX or XXX</u>	Exa	amples:	
	 stance Temperature Package ersion Range	a)	MCP4017T-502E/LTY:	5 kΩ, 6-LD SC-70.
ve		b)	MCP4017T-103E/LTY:	10 kΩ, 6-LD SC-70.
Device:	MCP4017: Single Rheostat with I ² C interface	c)	MCP4017T-503E/LTY:	50 kΩ, 6-LD SC-70.
	MCP4017T: Single Rheostat with I ² C interface (Tape and Reel) MCP4018: Single Potentiometer to GND	d)	MCP4017T-104E/LTY:	100 kΩ, 6-LD SC-70.
	MCP4018: Single Potentiometer to GND MCP4018T: Single Potentiometer to GND	a)	MCP4018T-502E/LTY:	5 kΩ, 6-LD SC-70.
	with I ² C Interface (Tape and Reel)	b)	MCP4018T-103E/LTY:	10 kΩ, 6-LD SC-70.
	MCP4019: Single Rheostat to GND with I ² C Interface	c)	MCP4018T-503E/LTY:	6-LD SC-70.
	MCP4019T: Single Rheostat to GND with I ² C Interface (Tape and Reel)	d)	MCP4018T-104E/LTY:	100 kΩ, 6-LD SC-70.
Resistance	502 = 5 kΩ	a)	MCP4019T-502E/LT:	5 kΩ, 5-LD SC-70.
Version:	103 = 10 kΩ 503 = 50 kΩ	b)	MCP4019T-103E/LT:	10 kΩ, 5-LD SC-70.
	104 = 100 kΩ	c)	MCP4019T-503E/LT:	50 kΩ, 5-LD SC-70.
Temperature Range:	E = -40°C to +125°C	d)	MCP4019T-104E/LT:	100 kΩ, 5-LD SC-70.
Package:	LT = Plastic Small Outline Transistor (SC70), 5-lead, 6-lead			
	LTY = Plastic Small Outline Transistor (SC70), 5-lead, 6-lead, NiPdAu-plated			

MCP4017/18/19

NOTES:

Note the following details of the code protection feature on Microchip products:

- Microchip products meet the specifications contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is secure when used in the intended manner, within operating specifications, and under normal conditions.
- Microchip values and aggressively protects its intellectual property rights. Attempts to breach the code protection features of Microchip product is strictly prohibited and may violate the Digital Millennium Copyright Act.
- Neither Microchip nor any other semiconductor manufacturer can guarantee the security of its code. Code protection does not
 mean that we are guaranteeing the product is "unbreakable" Code protection is constantly evolving. Microchip is committed to
 continuously improving the code protection features of our products.

This publication and the information herein may be used only with Microchip products, including to design, test, and integrate Microchip products with your application. Use of this information in any other manner violates these terms. Information regarding device applications is provided only for your convenience and may be superseded by updates. It is your responsibility to ensure that your application meets with your specifications. Contact your local Microchip sales office for additional support or, obtain additional support at https:// www.microchip.com/en-us/support/design-help/client-supportservices.

THIS INFORMATION IS PROVIDED BY MICROCHIP "AS IS". MICROCHIP MAKES NO REPRESENTATIONS OR WAR-RANTIES OF ANY KIND WHETHER EXPRESS OR IMPLIED, WRITTEN OR ORAL, STATUTORY OR OTHERWISE, RELATED TO THE INFORMATION INCLUDING BUT NOT LIMITED TO ANY IMPLIED WARRANTIES OF NON-INFRINGEMENT, MERCHANTABILITY, AND FITNESS FOR A PARTICULAR PURPOSE, OR WARRANTIES RELATED TO ITS CONDITION, QUALITY, OR PERFORMANCE.

IN NO EVENT WILL MICROCHIP BE LIABLE FOR ANY INDI-RECT, SPECIAL, PUNITIVE, INCIDENTAL, OR CONSE-QUENTIAL LOSS, DAMAGE, COST, OR EXPENSE OF ANY KIND WHATSOEVER RELATED TO THE INFORMATION OR ITS USE, HOWEVER CAUSED, EVEN IF MICROCHIP HAS BEEN ADVISED OF THE POSSIBILITY OR THE DAMAGES ARE FORESEEABLE. TO THE FULLEST EXTENT ALLOWED BY LAW, MICROCHIP'S TOTAL LIABILITY ON ALL CLAIMS IN ANY WAY RELATED TO THE INFORMATION OR ITS USE WILL NOT EXCEED THE AMOUNT OF FEES, IF ANY, THAT YOU HAVE PAID DIRECTLY TO MICROCHIP FOR THE INFORMATION.

Use of Microchip devices in life support and/or safety applications is entirely at the buyer's risk, and the buyer agrees to defend, indemnify and hold harmless Microchip from any and all damages, claims, suits, or expenses resulting from such use. No licenses are conveyed, implicitly or otherwise, under any Microchip intellectual property rights unless otherwise stated.

Trademarks

The Microchip name and logo, the Microchip logo, Adaptec, AVR, AVR logo, AVR Freaks, BesTime, BitCloud, CryptoMemory, CryptoRF, dsPIC, flexPWR, HELDO, IGLOO, JukeBlox, KeeLoq, Kleer, LANCheck, LinkMD, maXStylus, maXTouch, MediaLB, megaAVR, Microsemi, Microsemi logo, MOST, MOST logo, MPLAB, OptoLyzer, PIC, picoPower, PICSTART, PIC32 logo, PolarFire, Prochip Designer, QTouch, SAM-BA, SenGenuity, SpyNIC, SST, SST Logo, SuperFlash, Symmetricom, SyncServer, Tachyon, TimeSource, tinyAVR, UNI/O, Vectron, and XMEGA are registered trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

AgileSwitch, ClockWorks, The Embedded Control Solutions Company, EtherSynch, Flashtec, Hyper Speed Control, HyperLight Load, Libero, motorBench, mTouch, Powermite 3, Precision Edge, ProASIC, ProASIC Plus, ProASIC Plus logo, Quiet-Wire, SmartFusion, SyncWorld, TimeCesium, TimeHub, TimePictra, TimeProvider, and ZL are registered trademarks of Microchip Technology Incorporated in the U.S.A.

Adjacent Key Suppression, AKS, Analog-for-the-Digital Age, Any Capacitor, AnyIn, AnyOut, Augmented Switching, BlueSky, BodyCom, Clockstudio, CodeGuard, CryptoAuthentication, CryptoAutomotive, CryptoCompanion, CryptoController, dsPICDEM, dsPICDEM.net, Dynamic Average Matching, DAM, ECAN, Espresso T1S, EtherGREEN, EyeOpen, GridTime, IdealBridge, IGaT, In-Circuit Serial Programming, ICSP, INICnet, Intelligent Paralleling, IntelliMOS, Inter-Chip Connectivity, JitterBlocker, Knob-on-Display, MarginLink, maxCrypto, maxView, memBrain, Mindi, MiWi, MPASM, MPF, MPLAB Certified logo, MPLIB, MPLINK, mSiC, MultiTRAK, NetDetach, Omniscient Code Generation, PICDEM, PICDEM.net, PICkit, PICtail, Power MOS IV, Power MOS 7, PowerSmart, PureSilicon, QMatrix, REAL ICE, Ripple Blocker, RTAX, RTG4, SAM-ICE, Serial Quad I/O, simpleMAP, SimpliPHY, SmartBuffer, SmartHLS, SMART-I.S., storClad, SQI, SuperSwitcher, SuperSwitcher II, Switchtec, SynchroPHY, Total Endurance, Trusted Time, TSHARC, Turing, USBCheck, VariSense, VectorBlox, VeriPHY, ViewSpan, WiperLock, XpressConnect, and ZENA are trademarks of Microchip Technology Incorporated in the U.S.A. and other countries.

SQTP is a service mark of Microchip Technology Incorporated in the U.S.A.

The Adaptec logo, Frequency on Demand, Silicon Storage Technology, and Symmcom are registered trademarks of Microchip Technology Inc. in other countries.

GestIC is a registered trademark of Microchip Technology Germany II GmbH & Co. KG, a subsidiary of Microchip Technology Inc., in other countries.

All other trademarks mentioned herein are property of their respective companies.

 $\ensuremath{\mathbb{C}}$ 2024, Microchip Technology Incorporated and its subsidiaries.

All Rights Reserved.

For information regarding Microchip's Quality Management Systems, please visit www.microchip.com/quality. ISBN: 978-1-6683-0340-5



Worldwide Sales and Service

AMERICAS

Corporate Office 2355 West Chandler Blvd. Chandler, AZ 85224-6199 Tel: 480-792-7200 Fax: 480-792-7277 Technical Support: http://www.microchip.com/ support

Web Address: www.microchip.com

Atlanta Duluth, GA Tel: 678-957-9614 Fax: 678-957-1455

Austin, TX Tel: 512-257-3370

Boston Westborough, MA Tel: 774-760-0087 Fax: 774-760-0088

Chicago Itasca, IL Tel: 630-285-0071 Fax: 630-285-0075

Dallas Addison, TX Tel: 972-818-7423 Fax: 972-818-2924

Detroit Novi, MI Tel: 248-848-4000

Houston, TX Tel: 281-894-5983

Indianapolis Noblesville, IN Tel: 317-773-8323 Fax: 317-773-5453 Tel: 317-536-2380

Los Angeles Mission Viejo, CA Tel: 949-462-9523 Fax: 949-462-9608 Tel: 951-273-7800

Raleigh, NC Tel: 919-844-7510

New York, NY Tel: 631-435-6000

San Jose, CA Tel: 408-735-9110 Tel: 408-436-4270

Canada - Toronto Tel: 905-695-1980 Fax: 905-695-2078

ASIA/PACIFIC

Australia - Sydney Tel: 61-2-9868-6733

China - Beijing Tel: 86-10-8569-7000 China - Chengdu

Tel: 86-28-8665-5511 China - Chongqing Tel: 86-23-8980-9588

China - Dongguan Tel: 86-769-8702-9880

China - Guangzhou Tel: 86-20-8755-8029

China - Hangzhou Tel: 86-571-8792-8115

China - Hong Kong SAR Tel: 852-2943-5100

China - Nanjing Tel: 86-25-8473-2460

China - Qingdao Tel: 86-532-8502-7355

China - Shanghai Tel: 86-21-3326-8000

China - Shenyang Tel: 86-24-2334-2829

China - Shenzhen Tel: 86-755-8864-2200

China - Suzhou Tel: 86-186-6233-1526

China - Wuhan Tel: 86-27-5980-5300

China - Xian Tel: 86-29-8833-7252

China - Xiamen Tel: 86-592-2388138 China - Zhuhai

Tel: 86-756-3210040

ASIA/PACIFIC

India - Bangalore Tel: 91-80-3090-4444

India - New Delhi Tel: 91-11-4160-8631 India - Pune

Tel: 91-20-4121-0141 Japan - Osaka

Tel: 81-6-6152-7160

Japan - Tokyo Tel: 81-3-6880- 3770 Korea - Daegu

Tel: 82-53-744-4301 Korea - Seoul

Tel: 82-2-554-7200

Malaysia - Kuala Lumpur Tel: 60-3-7651-7906

Malaysia - Penang Tel: 60-4-227-8870

Philippines - Manila Tel: 63-2-634-9065

Singapore Tel: 65-6334-8870

Taiwan - Hsin Chu Tel: 886-3-577-8366

Taiwan - Kaohsiung Tel: 886-7-213-7830

Taiwan - Taipei Tel: 886-2-2508-8600

Thailand - Bangkok Tel: 66-2-694-1351

Vietnam - Ho Chi Minh Tel: 84-28-5448-2100

Tel: 972-9-775-5100

Italy - Milan Tel: 39-0331-742611 Fax: 39-0331-466781

EUROPE

Austria - Wels

Tel: 43-7242-2244-39

Tel: 45-4485-5910

Fax: 45-4485-2829

Tel: 358-9-4520-820

Tel: 33-1-69-53-63-20

Fax: 33-1-69-30-90-79

Germany - Garching

Tel: 49-2129-3766400

Germany - Heilbronn

Germany - Karlsruhe

Tel: 49-7131-72400

Tel: 49-721-625370

Germany - Munich

Tel: 49-89-627-144-0

Fax: 49-89-627-144-44

Germany - Rosenheim

Israel - Hod Hasharon

Tel: 49-8031-354-560

Tel: 49-8931-9700

Germany - Haan

Finland - Espoo

France - Paris

Fax: 43-7242-2244-393

Denmark - Copenhagen

Italy - Padova Tel: 39-049-7625286

Netherlands - Drunen Tel: 31-416-690399 Fax: 31-416-690340

Norway - Trondheim Tel: 47-7288-4388

Poland - Warsaw Tel: 48-22-3325737

Romania - Bucharest Tel: 40-21-407-87-50

Spain - Madrid Tel: 34-91-708-08-90 Fax: 34-91-708-08-91

Sweden - Gothenberg Tel: 46-31-704-60-40

Sweden - Stockholm Tel: 46-8-5090-4654

UK - Wokingham Tel: 44-118-921-5800 Fax: 44-118-921-5820